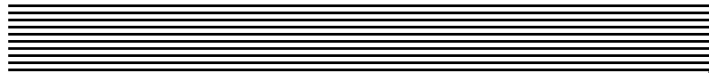


APPENDIX A



EXAMPLE *MATHEMATICA* NOTEBOOKS

Filter Design For Signal Processing

Using **MATLAB** and *Mathematica*

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A.1 Analysis by Transform Method of Analog LTI Circuits

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■ A.1.1 References

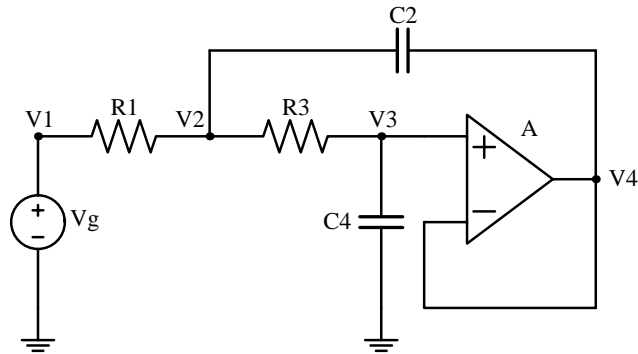
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For Use with Programmable Pocket Calculators and Mini Computers.
John Wiley and Sons, Ltd., New York, 1981, pp. 38–39
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“A practical method of designing RC active filters,”
IRE Trans. Circuit Theory, CT-2, Mar. 1955, pp. 74–85.
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“A tool for symbolic analysis and design of analog active filters,”
5th Int. Workshop, Symbolic Methods, Applications, Circuit Design,
SMACD’98, Kaiserslautern, Germany, pp. 71–74, Oct. 1998.
4. D. V. Tasic, M. D. Lutovac,
“Symbolic computation of impulse, step and sine responses of linear time-invariant systems,”
Proc. 10th Int. Symp. Theor. Electrical Engineering,
ISTET99, Magdeburg, Germany, Sep. 1999, pp. 653–657.
5. D. V. Tasic,
“SALECAS - a package for symbolic analysis of linear circuits and systems,”
4th Int. Workshop, Symbolic Methods, Applications, Circuit Design,
SMACD, Leuven, Belgium, pp. 227–230, Oct. 1996.
6. M. D. Lutovac, D. V. Tasic and B. L. Evans,
“Advanced Filter Design for Signal Processing using MATLAB and Mathematica,”
<http://iritel.iritel.bg.ac.yu/~lutovac/www/afdhhome.htm>
<http://galeb.etf.bg.ac.yu/~tasic/afdhhome.htm>
<http://www.ece.utexas.edu/~bevans/>

■ A.1.2 Initialization

```
SetDirectory[HomeDirectory[]];
<<Calculus`LaplaceTransform`
<<afd\math\m\clearall.m
<<afd\math\m\drawafil.m
<<afd\math\m\drawacc.m
```

■ A.1.3 Schematic

```
DrawLPSK[0, 0, 1/2, 1.25, 8];
```



■ A.1.4 Circuit Analysis

■ Reduced Modified Nodal Analysis (RMNA)

```
CircuitEquations = {
  V1 == Vg
  , (V2-V1)/R1 + (V2-V3)/R3 + (V2-V4)*(s*C2) == 0
  , (V3-V2)/R3 + V3*(s*C4) == 0
  , V3 == V4
};
NodeVoltages = {V1,V2,V3,V4};
CircuitResponse = Together[Flatten[
  Solve[CircuitEquations,NodeVoltages]
]];
```

■ A.1.5 Response

```
V1s = Simplify[V1 /. CircuitResponse];
Collect[Numerator[%],s]/Collect[Denominator[%],s]
```

Vg

```
V2s = Simplify[V2 /. CircuitResponse];
Collect[Numerator[%],Vg]/Collect[Denominator[%],s]
```

$$\frac{(1 + C4 R3 s) Vg}{1 + (C4 R1 + C4 R3) s + C2 C4 R1 R3 s^2}$$

```
V3s = Simplify[V3 /. CircuitResponse];
Collect[Numerator[%],s]/Collect[Denominator[%],s]
```

$$\frac{V_g}{1 + (C_4 R_1 + C_4 R_3) s + C_2 C_4 R_1 R_3 s^2}$$

```
V4s = Simplify[V4 /. CircuitResponse];
Collect[Numerator[%],s]/Collect[Denominator[%],s]
```

$$\frac{V_g}{1 + (C_4 R_1 + C_4 R_3) s + C_2 C_4 R_1 R_3 s^2}$$

■ A.1.6 Voltage Transfer Function

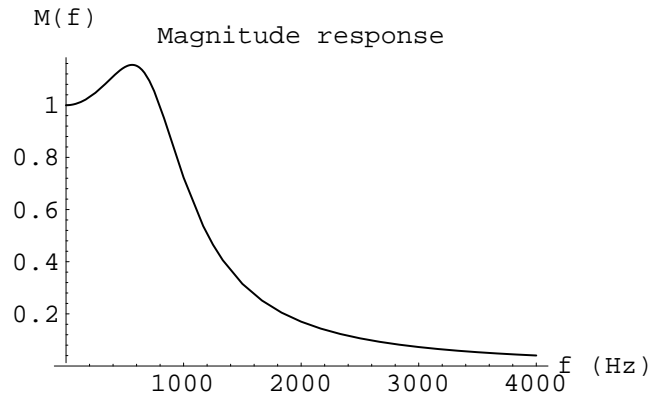
```
H = V4s/Vg;
Collect[Numerator[%],s]/Collect[Denominator[%],s]
```

$$\frac{1}{1 + (C_4 R_1 + C_4 R_3) s + C_2 C_4 R_1 R_3 s^2}$$

```
M = Abs[H] /. {C2->4*C, C4->C, R1->R, R3->R} ./ s -> I*2*Pi*f
```

$$\text{Abs}\left[\frac{1}{1 + 4 * C f \text{Pi} R - 16 C^2 F \text{Pi}^2 R^2}\right]$$

```
Plot[ {M ./ {C->10^(-8), R->10^4}}, {f,0,4000}
, PlotRange -> All
, AxesLabel -> {"f (Hz)", "M(f)"}
, AxesOrigin -> {0,0}
, PlotLabel -> "Magnitude response"];
```

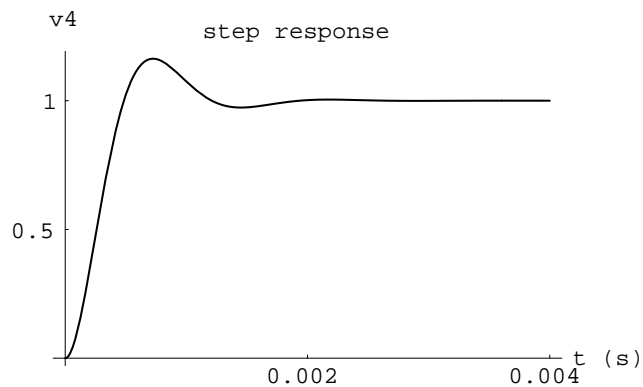


■ A.1.7 Step Response

```
v4t = InverseLaplaceTransform[
H/s /. {C2->4*C, C4->C, R1->R, R3->R}
, s, t];
v4t //Together //Simplify
```

$$1 - \frac{\cos\left[\frac{\sqrt{3}t}{4CR}\right] - \sin\left[\frac{\sqrt{3}t}{4CR}\right]}{e^{t/(4CR)}} - \frac{\sin\left[\frac{\sqrt{3}t}{4CR}\right]}{\sqrt{3}e^{t/(4CR)}}$$

```
Plot[v4t /. {C->10^(-8), R->10^4}
, {t,0,4*10^(-3)}
, PlotRange -> All
, AxesLabel -> {"t (s)", "v4"}
, Ticks -> {{0,0.002,0.004},{0,0.5,1}}
, PlotLabel -> "step response"];
```



A.2 Analysis by Transform Method of Discrete-Time LTI Systems

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■ A.2.1 References

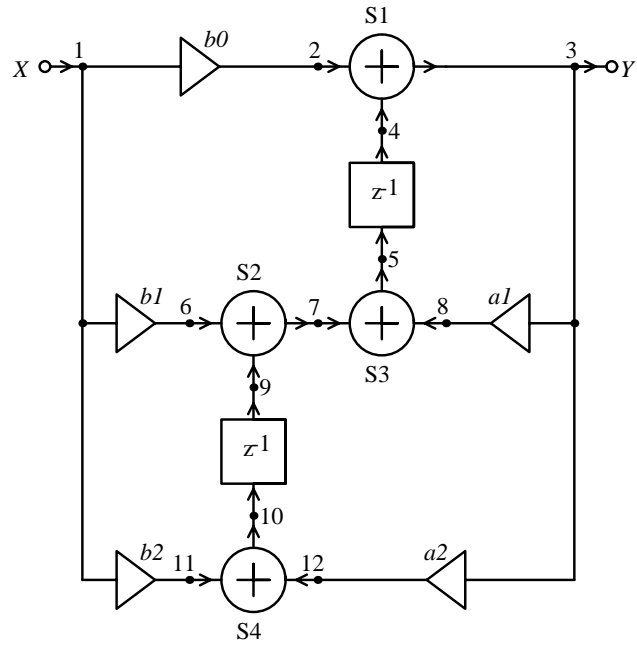
1. Alan Oppenheim, Ronald Schaffer, "Digital Signal Processing," Prentice-Hall, Englewood Cliffs, New Jersey, 1975.
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3. D. V. Tasic, M. D. Lutovac,
"Symbolic computation of impulse, step and sine responses of linear time-invariant systems,"
Proc. 10th Int. Symp. Theor. Electrical Engineering,
ISTET99, Magdeburg, Germany, Sep. 1999, pp. 653–657.
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"Mathematica, Signals and Systems,"
Georgia Tech Research Corp., Atlanta, Georgia, 1995.
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"Symbolic derivation of transfer functions of discrete-time systems,"
ISTET'97, Palermo, Italia, 9–11 June 1997, pp. 311–314.
6. D. V. Tasic,
"SALECAS - a package for symbolic analysis of linear circuits and systems,"
SMACD, Leuven, Belgium, pp. 227–230, Oct. 1996.
7. M. D. Lutovac, D. V. Tasic and B. L. Evans,
"Advanced Filter Design for Signal Processing using MATLAB and Mathematica,"
<http://iritel.iritel.bg.ac.yu/~lutovac/www/afdhome.htm>
<http://galeb.etf.bg.ac.yu/~tasic/afdhome.htm>
<http://www.ece.utexas.edu/~bevans/>

■ A.2.2 Initialization

```
SetDirectory[HomeDirectory[]];
<<afd\math\m\clearall.m
AppendTo[$Path, "APPS"];
Needs["SignalProcessing`Master`"]
<<afd\math\m\drawdfil.m
<<afd\math\m\drawiirf.m
```

■ A.2.3 Block Diagram

`DrawTDF2[0,0,1,1/0.8,10];`



■ A.2.4 Analysis ($v=1/z$)

```
ElementEquations = {
  Y1 == X
, Y2 == b0*Y1 + Xb0
, Y3 == Y2 + Y4
, Y4 == v*Y5
, Y5 == Y7 + Y8
, Y6 == b1*Y1 + Xb1
, Y7 == Y6 + Y9
, Y8 == a1*Y3 + Xa1
, Y9 == v*Y10
, Y10 == Y11 + Y12
, Y11 == b2*Y1 + Xb2
, Y12 == a2*Y3 + Xa2
};
```

```

NodeSignals = {Y1,Y2,Y3,Y4,Y5,Y6,Y7,Y8,Y9,Y10,Y11,Y12};
Response = Flatten[Solve[ElementEquations,NodeSignals]];
Y = Together[Y3/.Response]

```

$$\frac{-(b_0 X) - b_1 v X - b_2 v^2 X - v X a_1 - v^2 X a_2 - X b_0 - v X b_1 - v^2 X b_2}{-1 + a_1 v + a_2 v^2}$$

■ A.2.5 Transfer Function and Noise Transfer Functions

```

Hv = Y /. {X -> 1, Xa1 -> 0, Xa2 -> 0, Xb0 -> 0, Xb1 -> 0, Xb2 -> 0};
Ha1v = Y /. {X -> 0, Xa1 -> 1, Xa2 -> 0, Xb0 -> 0, Xb1 -> 0, Xb2 -> 0};
Ha2v = Y /. {X -> 0, Xa1 -> 0, Xa2 -> 1, Xb0 -> 0, Xb1 -> 0, Xb2 -> 0};
Hb0v = Y /. {X -> 0, Xa1 -> 0, Xa2 -> 0, Xb0 -> 1, Xb1 -> 0, Xb2 -> 0};
Hb1v = Y /. {X -> 0, Xa1 -> 0, Xa2 -> 0, Xb0 -> 0, Xb1 -> 1, Xb2 -> 0};
Hb2v = Y /. {X -> 0, Xa1 -> 0, Xa2 -> 0, Xb0 -> 0, Xb1 -> 0, Xb2 -> 1};
H = Collect[-Numerator[Hv],v] / (-Collect[Denominator[Hv],v]);
Ha1 = Collect[-Numerator[Ha1v],v]/(-Collect[Denominator[Ha1v],v]);
Ha2 = Collect[-Numerator[Ha2v],v]/(-Collect[Denominator[Ha2v],v]);
Hb0 = Collect[-Numerator[Hb0v],v]/(-Collect[Denominator[Hb0v],v]);
Hb1 = Collect[-Numerator[Hb1v],v]/(-Collect[Denominator[Hb1v],v]);
Hb2 = Collect[-Numerator[Hb2v],v]/(-Collect[Denominator[Hb2v],v]);
v2invz = {v->z^-1, v^2->z^-2};
Print["H(z) = ", H /. v2invz ]
Print["Ha1(z) = ", Ha1 /. v2invz ]
Print["Ha2(z) = ", Ha2 /. v2invz ]
Print["Hb0(z) = ", Hb0 /. v2invz ]
Print["Hb1(z) = ", Hb1 /. v2invz ]
Print["Hb2(z) = ", Hb2 /. v2invz ]

```

$$H(z) = \frac{b_0 + b_1 z^{-1} + b_2 z^{-2}}{1 - a_1 z^{-1} - a_2 z^{-2}}$$

$$Ha1(z) = \frac{z^{-1}}{1 - a_1 z^{-1} - a_2 z^{-2}}$$

$$Ha2(z) = \frac{z^{-2}}{1 - a_1 z^{-1} - a_2 z^{-2}}$$

$$Hb0(z) = \frac{1}{1 - a_1 z^{-1} - a_2 z^{-2}}$$

$$Hb1(z) = \frac{z^{-1}}{1 - a_1 z^{-1} - a_2 z^{-2}}$$

$$Hb2(z) = \frac{z^{-2}}{1 - a_1 z^{-1} - a_2 z^{-2}}$$

■ A.2.6 Complex Response in Terms of $v=z^{-1}$

`Y3z = Collect[-Numerator[Y],X]/(-Denominator[Y])`

$$\frac{(b_0 + b_1 v + b_2 v^2) X + v X a_1 + v^2 X a_2 + X b_0 + v X b_1 + v^2 X b_2}{1 - a_1 v - a_2 v^2}$$

■ A.2.7 Transfer Function

`H3z = H /. {b0->1, b1->2, b2->1, a1->0, a2->-1/2, v->z^{-1}}`

$$\frac{1 + z^{-2}}{1 + \frac{z^{-2}}{2}}$$

■ A.2.8 Impulse Response

`y3n = InverseZTransform[H3z,z,n]`

$$\frac{-\frac{1}{\sqrt{2}} \left(1 - 2 i \sqrt{2}\right) \text{DiscreteStep}[n]}{2} + \frac{\frac{1}{\sqrt{2}} \left(1 + 2 i \sqrt{2}\right) \text{DiscreteStep}[n]}{2} + 2 \text{KroneckerDelta}[n]$$

```
y3nDS = ComplexExpand[Coefficient[y3n,DiscreteStep[n]]]
```

$$\begin{aligned} & \cos\left[\frac{n\pi}{2}\right] \\ & -\left(\frac{1}{2} - \frac{n}{2}\right) \frac{\sin\left[\frac{n\pi}{2}\right]}{2} \end{aligned}$$

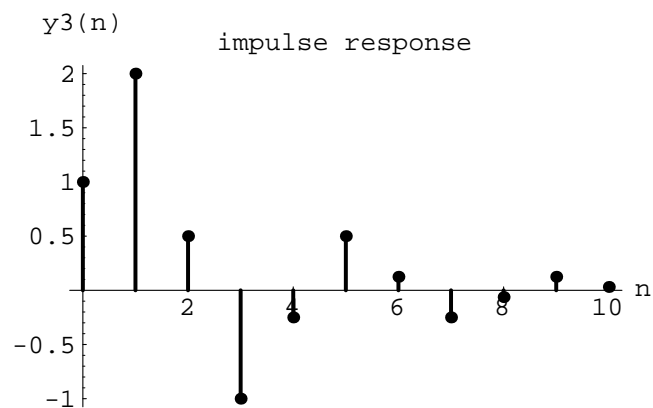
```
y3nKD = Coefficient[y3n,KroneckerDelta[n]]
```

```
2
```

```
h3n = y3nDS*DiscreteStep[n] + y3nKD*KroneckerDelta[n]
```

$$2 \text{ KroneckerDelta}[n] + \text{DiscreteStep}[n] \left(\cos\left[\frac{n\pi}{2}\right] - \left(\frac{1}{2} - \frac{n}{2}\right) \frac{\sin\left[\frac{n\pi}{2}\right]}{2} \right)$$

```
DiscreteSignalPlot[h3n
,{n,0,10}
,AxesLabel -> {"n","y3(n)"}
,PlotLabel -> "impulse response"
];
```



```
Table[{n,h3n}, {n,0,10}]
```

```
% //TableForm //N
```

```

      1      1      1      1      1
{0, 1}, {1, 2}, {2, -}, {3, -1}, {4, -(-)}, {5, -}, {6, -}, {7, -(-)},
      2      4      2      8      4

```

```

      1      1      1
{8, -(-)}, {9, -}, {10, --}
      16      8      32

```

```

0      1.
1.     2.
2.     0.5
3.     -1.
4.     -0.25
5.     0.5
6.     0.125
7.     -0.25
8.     -0.0625
9.     0.125
10.    0.03125

```

A.3 Switched Capacitor Filter - Mode 3a

Analysis and Design

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■ A.3.1 References

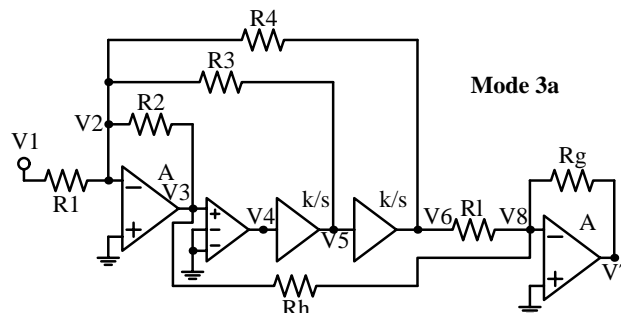
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Monolithic Filter Handbook, 1990.
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"Selective SC-filters with low passive sensitivity,"
Electronics Letters vol. 33, no. 8, pp. 674–675, Apr. 1997.
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"Programmable low-pass/high-pass SC-filters,"
Proc.9th Conf. MELECON '98, Tel-Aviv, Israel May 1998, pp.673–677
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"A tool for symbolic analysis and design of analog active filters,"
5th Int. Workshop, Symbolic Methods, Applications, Circuit Design,
SMACD'98, Kaiserslautern, Germany, pp. 71–74, Oct. 1998.
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"Advanced Filter Design for Signal Processing using MATLAB and Mathematica,"
<http://iritel.iritel.bg.ac.yu/~lutovac/www/afdhome.htm>
<http://galeb.etf.bg.ac.yu/~tasic/afdhome.htm>
<http://www.ece.utexas.edu/~bevans/>

■ A.3.2 Initialization

```
SetDirectory[HomeDirectory[]];
<<afd\math\m\clearall.m
<<afd\math\m\drawafil.m
<<afd\math\m\drawasc.m
```

■ A.3.3 Schematic

```
DrawMode3a[0,0,1/2,1/0.8,8];
```



■ A.3.4 Circuit Analysis

■ Reduced Modified Nodal Analysis

```

CircuitEquations = {
  V1 == Vg
, (V2-V1)/R1 + (V2-V3)/R2 + (V2-V5)/R3 + (V2-V6)/R4 == 0
, V3 == -A*V2
, V4 == V3
, V5 == V4*k/s
, V6 == V5*k/s
, (V8-V6)/R1 + (V8-V3)/Rh + (V8-V7)/Rg == 0
, V7 == -A*V8
};
NodeVoltages = {V1,V2,V3,V4,V5,V6,V7,V8};
CircuitResponse = Together[Flatten[
  Solve[CircuitEquations,NodeVoltages]
]];
Print["V1 = ", V1 /. CircuitResponse]
Print["V7 = ", V7 /. CircuitResponse]

V1 = Vg
V7 = -((A R2 R3 R4 s^2 (- (k Rg Rh) - Rg R1 s) Vg) /
  ((Rg Rh + Rg R1 + Rh R1 + A Rh R1)
  ((R1 R2 R3 + R1 R2 R4 + R1 R3 R4 + R2 R3 R4) s^3 -
  A (- (k R1 R2 R3 s) + s^2 (- (k R1 R2 R4) - R1 R3 R4 s))))))

```

■ A.3.5 Voltage Transfer Function

```

H = V7/V1 /. CircuitResponse //Together //Simplify;
H3a = Limit[H, A->Infinity];
Print["H(s) = ",
  Factor[Collect[Numerator[H3a],s]]/Factor[Collect[Denominator[H3a],s]]
]

```

$$H(s) = \frac{Rg R2 R3 R4 (k Rg Rh + R1 s)^2}{Rh R1 R1 (k R2 R3 + k R2 R4 s + R3 R4 s)^2}$$

■ A.3.6 Definitions and Procedures

```

PoleQpole[H_,s_] := Module[{den,fp,Qp},
  den = Denominator[H];
  fp = Sqrt[Coefficient[den,s,0]/Coefficient[den,s,2]]/(2*Pi);
  Qp = (Coefficient[den,s,2]/Coefficient[den,s,1])*(2*Pi*fp);
  Simplify[{fp, Qp}]];

```

```

ZeroQzero[H_,s_] := Module[{fz,num,Qz0},
  num = Numerator[H];
  Qz0 = (Coefficient[num,s,2]/Coefficient[num,s,1]);
  fz = Sqrt[Coefficient[num,s,0]/Coefficient[num,s,2]]/(2*Pi);
  Simplify[{fz, Qz0*fz}]];
Sensitivity[F_,x_] := (x/F)*D[F,x];
GSP[F_,A_] := Limit[A*Sensitivity[F,A],A->Infinity]//Simplify;
PrintLabeledList[expressions_List,labels_List] := Map[
  Print#[[1]], " = ",#[[2]]]&
,Transpose[{labels,expressions}]
];

```

■ A.3.7 Poles, Zeros, Q-Factors

```

{fp,Qp} = Simplify[PoleQpole[H,s]];
PrintLabeledList[{fp,Qp},{ "fp","Qp"}];

```

$$\begin{aligned}
 \text{fp} = & \frac{\sqrt{\frac{A^2 k^2 R_1 R_2 R_3}{R_1 R_2 R_3 + R_1 R_2 R_4 + R_1 R_3 R_4 + A R_1 R_3 R_4 + R_2 R_3 R_4}}}{2 \pi} \\
 \text{Qp} = & \frac{(\sqrt{\frac{A^2 k^2 R_1 R_2 R_3}{R_1 R_2 R_3 + R_1 R_2 R_4 + R_1 R_3 R_4 + A R_1 R_3 R_4 + R_2 R_3 R_4}})}{(R_1 R_2 R_3 + R_1 R_2 R_4 + R_1 R_3 R_4 + A R_1 R_3 R_4 + R_2 R_3 R_4)) / (A k R_1 R_2 R_4)}
 \end{aligned}$$

```

fp0 = Limit[fp, A -> Infinity];
Qp0 = Simplify[Limit[Qp, A -> Infinity]/.k->1];
PrintLabeledList[{fp0,Qp0},{ "fp","Qp"}];

```

$$\begin{aligned}
 \text{fp} = & \frac{\sqrt{\frac{k^2 R_2}{R_4}}}{2 \pi} \\
 \text{Qp} = & \frac{R_3 \sqrt{\frac{R_2}{R_4}}}{R_2}
 \end{aligned}$$

```
{fz,Qz} = Simplify[ZeroQzero[H,s]];
PrintLabeledList[{fz,Qz},{fz,"Qz"}];
```

```
Power::infy: Infinite expression  $\frac{1}{0}$  encountered.
```

```

      2
      k Rh
Sqrt[-----]
      R1
fz = -----
      2 Pi
Qz = ComplexInfinity
```

■ A.3.8 Gain-Sensitivity Product (GSP)

```
GSPfp = GSP[fp,A];
GSPQp = GSP[Qp,A];
PrintLabeledList[{GSPfp,GSPQp},{GSPfp,"GSPQp"}];
```

```

      1      R2      R2      R2
GSPfp = - + ---- + ---- + ----
      2      2 R1      2 R3      2 R4
      1      R2      R2      R2
GSPQp = -(-) - ---- - ---- - ----
      2      2 R1      2 R3      2 R4
```

■ A.3.9 Design

■ Find Element Values

```
DesignMode3a[K_,Qp_,wp_,wz_,fc1k_,P_:100,R1_:R1nom,R2_:R2nom,Rh_:Rhnom]:=
Module[{R3,R4,R1,Rg},
  R4 = R2*(2*Pi*fc1k/(P*wp))^2;
  R3 = Qp*Sqrt[R2*R4];
  R1 = Rh*(2*Pi*fc1k/(P*wz))^2;
  Rg = K*Rh*R1/R2;
  {R1,R2,R3,R4,R1,Rh,Rg}
];
{R1,R2,R3,R4,R1,Rh,Rg} = Together[
  DesignMode3a[K,Q,W,Z,Fc,P,R1n,R2n,Rhn]];
PrintLabeledList[{R1,R2,R3,R4,R1,Rh,Rg}
  ,{"R1","R2","R3","R4","R1","Rh","Rg"}];
```

```
R1 = R1n
R2 = R2n
```

```

      2      2      2
      Fc Pi R2n
R3 = 2 Q Sqrt[-----]
      2      2
      P W
```

$$R4 = \frac{4 Fc \sqrt{P} \sqrt{W}}{R2n}$$

$$R1 = \frac{4 Fc \sqrt{P} \sqrt{W}}{Rhn}$$

$$Rh = Rhn$$

$$K Rhn R1n$$

$$Rg = \frac{R2n}{Rhn}$$

■ Test

```
H3atest = Simplify[ExpandAll[Together[Limit[H , A -> Infinity]] /.
  {Sqrt[x^2*y^2/z^2] -> x*y/z, Sqrt[x^2*y^2*p^2/z^2] -> x*y*p/z
  , Sqrt[x^2*y^2*p^2/(z^2*n^2)] -> x*y*p/(z*n)} ]];
num = Numerator[H3atest];
den = Denominator[H3atest];
numlist = CoefficientList[num,s];
denlist = CoefficientList[den,s];
K3at =(numlist[[3]]/denlist[[3]]);
H3at = K3at *
Simplify[num/numlist[[3]]]/Simplify[den/denlist[[3]]] /. k-> 2*Pi*Fc/P
```

$$K (s^2 + Z^2)$$

$$s^2 + \frac{s W}{Q} + W^2$$

■ Design Examples

```

values = {K ->1, Q -> 1.0349, W -> 2*Pi*1710.9457, Z -> 2*Pi*5129.3034
, Fc -> 256.*10^3, P -> 100, R12 ->23.16*10^3
, R22 ->10.*10^3, Rh2 ->238.6*10^3} //N;
h1 = H /. k-> 2*Pi*Fc/P /.values //N;
H3atest = Limit[h1, A->Infinity];
H3atest =H3atest /. Sqrt[x.^2] -> x;
num = Numerator[H3atest];
den = Denominator[H3atest];
numlist = CoefficientList[num,s];
denlist = CoefficientList[den,s];
K3at =(numlist[[3]]/denlist[[3]]);
H3at = K3at * Simplify[num/numlist[[3]]]/Simplify[den/denlist[[3]]]

```

$$\frac{1. (1.03867 \cdot 10^9 + 1. s^2)}{1.15567 \cdot 10^8 + 10387.7 s + 1. s^2}$$

```

Rexample1 = N[{R1,R2,R3,R4,R1,Rh,Rg}/. values] /. Sqrt[R2n^2]->R2n;
PrintLabeledList[Rexample1,{"R1","R2","R3","R4","R1","Rh","Rg"}];

```

```

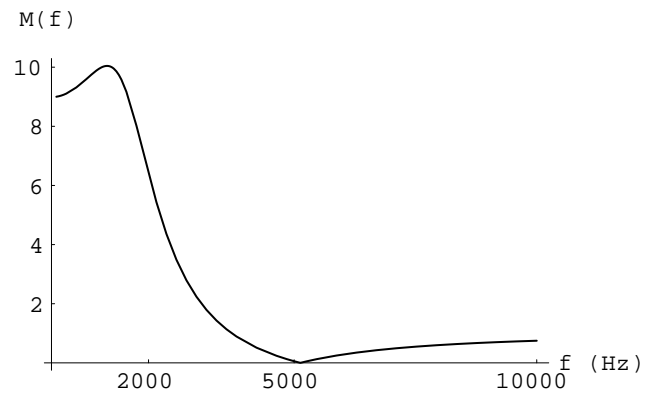
R1 = R1n
R2 = R2n
R3 = 1.54847 R2n
R4 = 2.23876 R2n
R1 = 0.249094 Rhn
Rh = Rhn
Rg = -----
      R2n

```

```

Plot[{Abs[H3at] /. s -> I*2*Pi*f}
, {f, 100, 10000}
,PlotRange -> All
,Ticks -> {{0,2000,5000,10000},{0,2,4,6,8,10}}
, AxesLabel -> {"f (Hz)","M(f)"}
];

```



■ A.3.10 Optimization

■ Find R_2/R_1 for Low Gain-Sensitivity Product

```
sf = (Simplify[N[Together[GSPfp /. values]]
      /. Sqrt[x_^2] -> x/. Sqrt[x_^2*y_^2] -> x*y ]
      /. Sqrt[R2n^2] -> R2n)
```

```
1.04624 +  $\frac{0.5 R_{2n}}{R_{1n}}$ 
```

■ Remark

We have to choose $R_1 > R_2$ to minimize GSP.

A.4 OTA-C General Biquad

Analysis and Design

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■ A.4.1 References

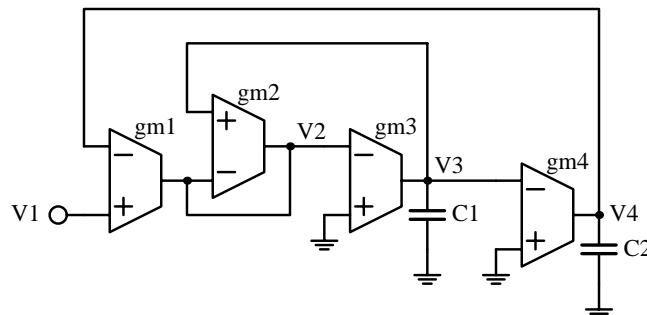
1. Wai-Kai-Chen, Ed.,
"The Circuits and Filters Handbook," p. 2479,
"CRC Press," "Boca Raton, Florida," "1995."
2. D.V. Tasic, M.D. Lutovac, B.L. Evans, I.M. Markoski,
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5th Int. Workshop, Symbolic Methods, Applications, Circuit Design,
SMACD'98, Kaiserslautern, Germany, pp. 71–74, Oct. 1998.
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<http://iritel.iritel.bg.ac.yu/~lutovac/www/afdhome.htm>
<http://galeb.etf.bg.ac.yu/~tasic/afdhome.htm>
<http://www.ece.utexas.edu/~bevans/>

■ A.4.2 Initialization

```
SetDirectory[HomeDirectory[]];
<<afd\math\m\clearall.m
<<afd\math\m\drawafil.m
<<afd\math\m\drawaota.m
```

■ A.4.3 Schematic

```
DrawOTAb[0, 0, 1/2, 1/08, 8]
```



■ A.4.4 Circuit Analysis

■ Reduced Modified Nodal Analysis

```
CircuitEquations = {
  V1 == Vg
, -(V1-V4)*gm1 - (V3-V2)*gm2 == 0
, -(-V2)*gm3 + V3/(1/(s*C1)) == 0
, -(-V3)*gm4 + V4/(1/(s*C2)) == 0
};

NodeVoltages = {V1,V2,V3,V4};
CircuitResponse = Together[Flatten[
  Solve[CircuitEquations,NodeVoltages]
]];
```

■ A.4.5 Voltage Transfer Function

```
H = V4/V1 /. CircuitResponse //Together //Simplify;
Print["H(s) = ", H]
Hhp = V2/V1 /. CircuitResponse //Together //Simplify;
Print["Hhp(s) = ", Hhp]
Hbp = V3/V1 /. CircuitResponse //Together //Simplify;
Print["Hbp(s) = ", Hbp]
```

$$H(s) = \frac{g_{m1} g_{m3} g_{m4}}{g_{m1} g_{m3} g_{m4} + C_2 g_{m2} g_{m3} s + C_1 C_2 g_{m2} s^2}$$

$$H_{hp}(s) = \frac{C_1 C_2 g_{m1} s}{g_{m1} g_{m3} g_{m4} + C_2 g_{m2} g_{m3} s + C_1 C_2 g_{m2} s^2}$$

$$H_{bp}(s) = -\left(\frac{C_2 g_{m1} g_{m3} s}{g_{m1} g_{m3} g_{m4} + C_2 g_{m2} g_{m3} s + C_1 C_2 g_{m2} s^2}\right)$$

■ A.4.6 Definitions and Procedures

```
PoleQpole[H_,s_] := Module[{den,fp,Qp},
  den = Denominator[H];
  fp = Sqrt[Coefficient[den,s,0]/Coefficient[den,s,2]]/(2*Pi);
  Qp = (Coefficient[den,s,2]/Coefficient[den,s,1])*(2*Pi*fp);
  Simplify[{fp, Qp}]];
ZeroQzero[H_,s_] := Module[{fz,num,Qz0},
  num = Numerator[H];
  Qz0 = (Coefficient[num,s,2]/Coefficient[num,s,1]);
  fz = Sqrt[Coefficient[num,s,0]/Coefficient[num,s,2]]/(2*Pi);
  Simplify[{fz, Qz0*fz}]];
```

```
PrintLabeledList[expressions_List, labels_List] := Map[
  Print[#[[1]], " = ", #[[2]]]&
, Transpose[{labels, expressions}]
];
```

■ A.4.7 Poles, Zeros, Q-Factors

```
{fp, Qp} = Simplify[PoleQpole[H, s]];
K1p = H /. s -> 0
PrintLabeledList[{fp, Qp}, {"fp", "Qp"}];
```

```
1
      gm1 gm3 gm4
      Sqrt[-----]
      C1 C2 gm2
fp = -----
      2 Pi
      gm1 gm3 gm4
      C1 Sqrt[-----]
      C1 C2 gm2
Qp = -----
      gm3
```

■ A.4.8 Design

■ Find Element Values

```
DesignOTA1[Qp_, wp_, C1_, C2_, gm1_, gm2_] :=
  Module[{gm3, gm4},
    gm3 = C1*wp/Qp;
    gm4 = C2*wp*Qp*gm2/gm1;
    {C1, C2, gm1, gm2, gm3, gm4}
  ];
{C1, C2, gm1, gm2, gm3, gm4} = Together[DesignOTA1[Q, W, c1, c2, g1, g2]];
PrintLabeledList[{C1, C2, gm1, gm2, gm3, gm4}, {"C1", "C2", "gm1", "gm2", "gm3", "gm4"}];
```

```
C1 = c1
C2 = c2
gm1 = g1
gm2 = g2
      c1 W
gm3 = ----
      Q
      c2 g2 Q W
gm4 = -----
      g1
```

■ A.4.9 Test

`Simplify[H]`

$$\frac{Q^2 W^2}{Q^2 s^2 + s^2 W^2 + Q^2 W^2}$$

■ Design Examples

```

values = {Q -> 4., W -> N[2*Pi*10^6] ,
  c1 -> 10.*10^(-12), c2 -> 10.*10^(-12), g1 -> g , g2 -> g} /N;
h1 = Together[H /. values] /. 1. -> 1;
Print["gm3 = ", 10^3*gm3 /. values, " mS "]
Print["gm4 = ", 10^3*gm4 /. values, " mS "]
h = (Numerator[h1]/g)/ (Simplify[Denominator[h1]/g])

gm3 = 0.015708 mS
gm4 = 0.251327 mS

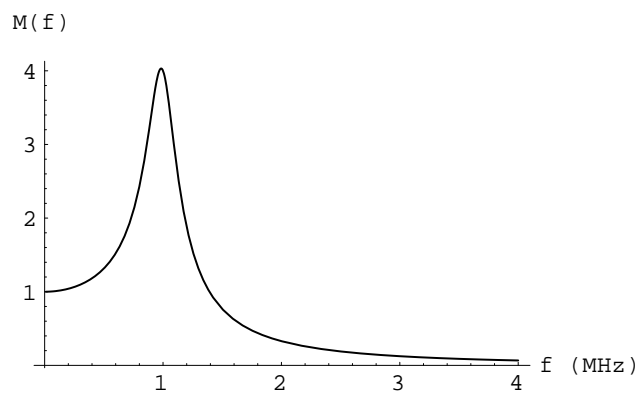
          13
    3.94784 10
-----
          13          6          2
    3.94784 10  + 1.5708 10  s + s

PrintLabeledList[N[{Q,W/(2*Pi)} /. values]
, {"Qp", "fp (Hz)"}];
Print["-----"]
Rexample1 = N[{C1*10^(12), C2*10^(12), gm1, gm2, gm3*10^3, gm4*10^3} /. values] ;
PrintLabeledList[Rexample1
, {"C1 (pF)", "C2 (pF)", "gm1", "gm2", "gm3 (mS)", "gm4 (mS)"}];

Qp = 4.
          6
fp (Hz) = 1. 10
-----
C1 (pF) = 10.
C2 (pF) = 10.
gm1 = g
gm2 = g
gm3 (mS) = 0.015708
gm4 (mS) = 0.251327

Plot[{Abs[h] /. s -> I*2*Pi*f*10^6}
, {f, 0.01, 4}
, PlotRange -> All
, AxesLabel -> {"f (MHz)", "M(f)"}
];

```



A.5 Lowpass-Medium-Q-Factor Active RC Filter Analysis and Design

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■ A.5.1 References

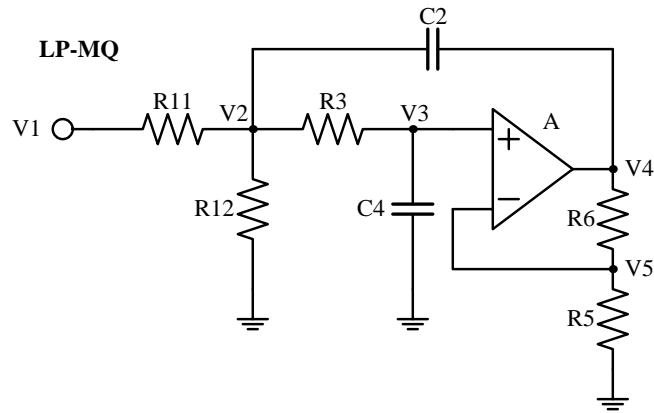
1. G. S. Moschytz and P. Horn,
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For Use with Programmable Pocket Calculators and Mini Computers
John Wiley and Sons, Ltd., New York, 1981, pp. 38–39.
2. R. P. Sallen and E. L. Key,
“A practical method of designing RC active filters,”
IRE Trans. Circuit Theory, CT-2, Mar. 1955, pp. 74–85.
3. D.V. Tasic, M.D. Lutovac, B.L. Evans, I.M. Markoski,
“A tool for symbolic analysis and design of analog active filters,”
5th Int. Workshop, Symbolic Methods, Applications, Circuit Design,
SMACD’98, Kaiserslautern, Germany, pp. 71–74, Oct. 1998.
4. M. D. Lutovac, D. V. Tasic and B. L. Evans,
“Advanced Filter Design for Signal Processing using MATLAB and Mathematica,”
<http://iritel.iritel.bg.ac.yu/~lutovac/www/afdhome.htm>
<http://galeb.etf.bg.ac.yu/~tasic/afdhome.htm>
<http://www.ece.utexas.edu/~bevans/>

■ A.5.2 Initialization

```
SetDirectory[HomeDirectory[]];  
<<afd\math\m\clearall.m  
<<afd\math\m\drawafil.m  
<<afd\math\m\drawarc.m
```


■ A.5.3 Schematic

DrawLPMQ[0,0,1/2,1/0.8];



■ A.5.4 Circuit Analysis

■ Reduced Modified Nodal Analysis

```
CircuitEquations = {V1 == Vg
, (V2-V1)/R11 + V2/R12 + (V2-V3)/R3 + (V2-V4)/(1/(s*C2)) == 0
, (V3-V2)/R3 + V3/(1/(s*C4)) == 0
, (V5-V4)/R6 + V5/R5 == 0
, (V3-V5)*A == V4};
NodeVoltages = {V1,V2,V3,V4,V5};
CircuitResponse = Together[Flatten[
  Solve[CircuitEquations,NodeVoltages]
]];
Print["V1 = ", V1 /. CircuitResponse]
Print["V4 = ", V4 /. CircuitResponse]
```

V1 = Vg

$$V4 = -\frac{(A^2 R_{12} R_3 R_5 V_g) / ((A^2 C_2 R_{11} R_{12} R_3 R_5 s - (-R_5 - A R_5 - R_6) (1 + C_4 R_3 s) (R_{11} R_{12} + R_{11} R_3 + R_{12} R_3 + C_2 R_{11} R_{12} R_3 s) + (-R_5 - A R_5 - R_6) (R_{11} R_{12} + A C_2 R_{11} R_{12} R_3 s))) - (A R_{12} R_3 (-R_5 - A R_5 - R_6) V_g)}{(A^2 C_2 R_{11} R_{12} R_3 R_5 s - (-R_5 - A R_5 - R_6) (1 + C_4 R_3 s) (R_{11} R_{12} + R_{11} R_3 + R_{12} R_3 + C_2 R_{11} R_{12} R_3 s) + (-R_5 - A R_5 - R_6) (R_{11} R_{12} + A C_2 R_{11} R_{12} R_3 s))}$$

■ A.5.5 Voltage Transfer Function

```
H = V4/V1 /. CircuitResponse //Together //Simplify;
Ha = Limit[H, A->Infinity];
Print["H(s) = ",
      Collect[Numerator[Ha],s]/Collect[Denominator[Ha],s]]

H(s) = (R12 (R5 + R6)) /
      (R11 R5 + R12 R5 + (C4 R11 R12 R5 + C4 R11 R3 R5 + C4 R12 R3 R5 - C2 R11 R12 R6) s +
       C2 C4 R11 R12 R3 R5 s )
```

■ A.5.6 Definitions and Procedures

```
PoleQpole[H_,s_] := Module[{den,fp,Qp},
  den = Denominator[H];
  fp = Sqrt[Coefficient[den,s,0]/Coefficient[den,s,2]]/(2*Pi);
  Qp = (Coefficient[den,s,2]/Coefficient[den,s,1])*(2*Pi*fp);
  Simplify[{fp, Qp}]];
ZeroQzero[H_,s_] := Module[{fz,num,Qz0},
  num = Numerator[H];
  Qz0 = (Coefficient[num,s,2]/Coefficient[num,s,1]);
  fz = Sqrt[Coefficient[num,s,0]/Coefficient[num,s,2]]/(2*Pi);
  Simplify[{fz, Qz0*fz}]];
Sensitivity[F_,x_] := (x/F)*D[F,x];
GSP[F_,A_] := Limit[A*Sensitivity[F,A],A->Infinity]//Simplify;
PrintLabeledList[expressions_List,labels_List] := Map[
  Print[#[[1]], " = ",#[[2]]]&
  ,Transpose[{labels,expressions}]
];
```

■ A.5.7 Poles, Zeros, Q-Factors

```
{fp,Qp} = Simplify[PoleQpole[H,s]];
PrintLabeledList[{fp,Qp},{ "fp","Qp"}];

      R11 + R12
      Sqrt[-----]
      C2 C4 R11 R12 R3
fp = -----
      2 Pi

      R11 + R12
Qp = (C2 C4 R11 R12 Sqrt[-----] R3 (R5 + A R5 + R6)) /
      C2 C4 R11 R12 R3

      (C2 R11 R12 R5 + C4 R11 R12 R5 + A C4 R11 R12 R5 + C4 R11 R3 R5 + A C4 R11 R3 R5 +
       C4 R12 R3 R5 + A C4 R12 R3 R5 + C2 R11 R12 R6 - A C2 R11 R12 R6 + C4 R11 R12 R6 +
       C4 R11 R3 R6 + C4 R12 R3 R6)
```

■ A.5.8 Gain-Sensitivity Product (GSP)

```
GSPfp = GSP[fp,A];
GSPQp = GSP[Qp,A];
PrintLabeledList[{GSPfp,GSPQp},{ "GSPfp", "GSPQp"}];

GSPfp = 0

GSPQp = 
$$\frac{C2 R11 R12 (R5 + R6)^2}{R5 (C4 R11 R12 R5 + C4 R11 R3 R5 + C4 R12 R3 R5 - C2 R11 R12 R6)}$$

```

■ A.5.9 Design

■ Find Element Values

```
DesignLPMQ[K_,Qp_,wp_,P_,C2x_,C4x_,R5x_] := Module[
{C2,C4,R1,R11,R12,R3,R5,R6,Ko},
C2 = C2x;
C4 = C4x;
R1 = 1/(wp*Sqrt[C2x*C4x*P]);
R3 = P*R1;
R5 = R5x;
R6 = R5*((1+P)*C4/C2-Sqrt[P*C4/C2]/Qp);
Ko = 1+R6/R5;
R11 = R1*Ko/K;
R12 = R1*Ko/(Ko-K);
{R11,R12,C2,R3,C4,R5,R6}
];
{R11,R12,C2,R3,C4,R5,R6} = Together[DesignLPMQ[K,Q,W,P,C2x,C4x,R5x]];
```

■ A.5.10 Design Example

```
values = {K -> 1., Q -> 7.5, W -> 2*Pi*2500., P -> 1.5333078
, C2x -> 33.*10^(-9), C4x -> 10.*10^(-9), R5x -> 6800.} //N;
PrintLabeledList[N[{K,Q,W/(2*Pi),P} /. values]
,{"K","Qp","fp (Hz)","P"} ];
Print["-----"]
PrintLabeledList[Together[{R11,R12,C2*10^9,R3,C4*10^9,R5,R6} /. values],
{"R11 (ohm)","R12 (ohm)","C2 (nF)","R3 (ohm)","C4 (nF)","R5 (ohm)","R6 (ohm)"}];

K = 1.
Qp = 7.5
fp (Hz) = 2500.
P = 1.53331
-----
R11 (ohm) = 4745.54
R12 (ohm) = 7011.9
C2 (nF) = 33.
R3 (ohm) = 4339.48
C4 (nF) = 10.
R5 (ohm) = 6800.
R6 (ohm) = 4602.13
```

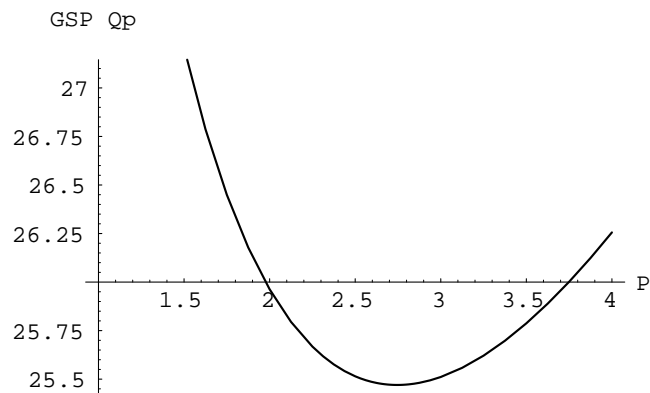
■ A.5.11 Optimization

■ Find P for Low Gain-Sensitivity Product

```
values = {K -> 1.476288, Q -> 7.5, W -> 2*Pi*2500.,
, C2x -> 100.*10^(-9), C4x -> 15.*10^(-9), R5x -> 10000.} //N;
gspQp = Together[Simplify[GSPQp] /. values]
```

$$\frac{0.435711 (7.66667 - 0.344265 \sqrt{P} + 1. P)^2}{\sqrt{P}}$$

```
P1 = 1.;
P2 = 4.;
Plot[gspQp /. values
, {P, P1, P2}
, AxesLabel -> {"P", "GSP Qp"}
];
```



```
{GSPmin,Pset} = FindMinimum[gspQp,{P,P1,P2}]
```

```
{25.4701, {P -> 2.74571}}
```

```
PrintLabeledList[N[{K,Q,W/(2*Pi),P} /. values /.Pset]
```

```
, {"K", "Qp", "fp (Hz)", "P"} ];
```

```
Print["-----"]
```

```
PrintLabeledList[Together[{R11,R12,C2*10^9,R3,C4*10^9,R5,R6} /. values /.Pset],
{"R11 (ohm)","R12 (ohm)","C2 (nF)","R3 (ohm)","C4 (nF)","R5 (ohm)","R6 (ohm)"}];
```

```
K = 1.47629
```

```
Qp = 7.5
```

```
fp (Hz) = 2500.
```

```
P = 2.74571
```

```
-----
```

```
R11 (ohm) = 991.99
```

```

          9
R12 (ohm) = 1.81368 10
C2 (nF) = 100.
R3 (ohm) = 2723.72
C4 (nF) = 15.
R5 (ohm) = 10000.
R6 (ohm) = 4762.89

Hhpmq = Together[Ha /. Pset /. values] ;
num = Numerator[Hhpmq];
den = Denominator[Hhpmq];
numlist = CoefficientList[num,s];
denlist = CoefficientList[den,s];
Hhpmq = (1/denlist[[3]]) * num/(den/denlist[[3]])//Factor

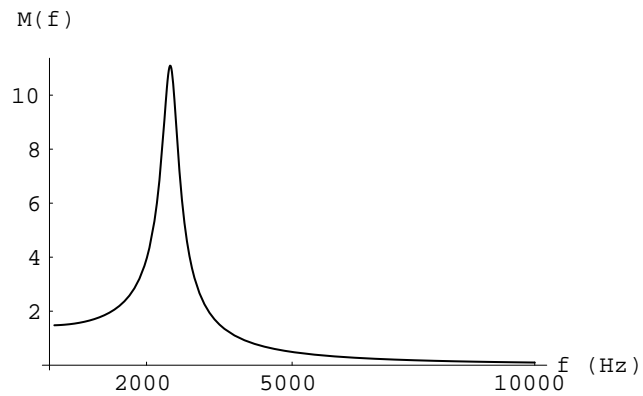
```

$$\frac{3.64259 \cdot 10^8}{2.4674 \cdot 10^8 s^2 + 2094.4 s + 1.}$$

```

Plot[{Abs[Hhpmq] /. s -> N[I*2*Pi*f]}
, {f, 100, 10000}
, PlotRange -> All
, Ticks -> {{0,2000,5000,10000},{0,2,4,6,8,10}}
, AxesLabel -> {"f (Hz)", "M(f)"}];

```



■ **Remark**

We choose $P=2.7457$ as a good choice because we obtain small GSP and $1/R_{12}=0$.

A.6 General Purpose High-Q-Factor RC Filter Analysis and Design

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■ A.6.1 References

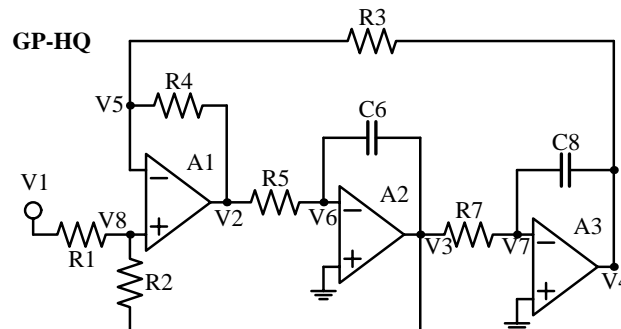
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For Use with Programmable Pocket Calculators and Mini Computers
John Wiley and Sons, Ltd., New York, 1981 pp. 64–65.
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State-variable synthesis for insensitive integrated circuit transfer functions,
IEEE J. Solid-State Circuits, SC-2, pp. 87–92, September, 1967.
3. D.V. Tasic, M.D. Lutovac, B.L. Evans, I.M. Markoski,
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5th Int. Workshop, Symbolic Methods, Applications, Circuit Design,
SMACD'98, Kaiserslautern, Germany, pp. 71–74, Oct. 1998.
4. M. D. Lutovac, D. V. Tasic and B. L. Evans,
“Advanced Filter Design for Signal Processing using MATLAB and Mathematica,”
<http://iritel.iritel.bg.ac.yu/~lutovac/www/afdhome.htm>
<http://galeb.etf.bg.ac.yu/~tasic/afdhome.htm>
<http://www.ece.utexas.edu/~bevans/>

■ A.6.2 Initialization

```
SetDirectory[HomeDirectory[]];
<<afd\math\m\cleara11.m
<<afd\math\m\drawafil.m
<<afd\math\m\drawarc.m
```

■ A.6.3 Schematic

```
DrawGPHQ[0,0,1/2,1/0.8];
```



■ A.6.4 Circuit Analysis

■ Reduced Modified Nodal Analysis

```
CircuitEquations = {V1 == Vg
, V2 == A*(V8 - V5)
, V3 == -A*V6
, V4 == -A*V7
, (V5-V2)/R4 + (V5-V4)/R3 == 0
, (V6-V2)/R5 + (V6-V3)/(1/(s*C6)) == 0
, (V7-V3)/R7 + (V7-V4)/(1/(s*C8)) == 0
, (V8-V1)/R1 + (V8-V3)/R2 == 0};
NodeVoltages = {V1,V2,V3,V4,V5,V6,V7,V8};
CircuitResponse = Together[Flatten[
  Solve[CircuitEquations,NodeVoltages]
]];

```

■ A.6.5 Voltage Transfer Function

```
H = V4/V1 /. CircuitResponse //Together //Simplify;
Ha = Limit[H, A->Infinity];
Print["H(s) = ",
  Collect[Numerator[Ha],s]/Collect[Denominator[Ha],s]]

```

$$H(s) = \frac{R_2 (R_3 + R_4)}{(R_1 R_4 + R_2 R_4 + (C_8 R_1 R_3 R_7 + C_8 R_1 R_4 R_7) s + (C_6 C_8 R_1 R_3 R_5 R_7 + C_6 C_8 R_2 R_3 R_5 R_7) s^2)}$$

```
K1p = Factor[Ha /. s->0]

```

$$\frac{R_2 (R_3 + R_4)}{(R_1 + R_2) R_4}$$

```
Hh = V2/V1 /. CircuitResponse //Together //Simplify;
Hha = Limit[Hh, A->Infinity];
Print["Hh(s) = ",
  Collect[Numerator[Hha],s]/Collect[Denominator[Hha],s]]

```

$$Hh(s) = \frac{(C_6 C_8 R_2 (R_3 + R_4) R_5 R_7 s^2)}{(R_1 R_4 + R_2 R_4 + (C_8 R_1 R_3 R_7 + C_8 R_1 R_4 R_7) s + (C_6 C_8 R_1 R_3 R_5 R_7 + C_6 C_8 R_2 R_3 R_5 R_7) s^2)}$$

```
Khp = Factor[Limit[Hha, s->Infinity]]

```

$$\frac{R_2 (R_3 + R_4)}{(R_1 + R_2) R_3}$$

```

Hb = V3/V1 /. CircuitResponse //Together //Simplify;
Hba = Limit[Hb, A->Infinity];
Print["Hb(s) = ",
      Collect[Numerator[Hba],s]/Collect[Denominator[Hba],s]]

Hb(s) = -((C8 R2 (R3 + R4) R7 s) /
          (R1 R4 + R2 R4 + (C8 R1 R3 R7 + C8 R1 R4 R7) s +
           (C6 C8 R1 R3 R5 R7 + C6 C8 R2 R3 R5 R7) s^2))

num = Numerator[Hba];
den = Denominator[Hba];
numlist = CoefficientList[num,s];
denlist = CoefficientList[den,s];
Hbp = Factor[numlist[[2]]/denlist[[2]]]

      R2
    -(-)
      R1

```

■ A.6.6 Definitions and Procedures

```

PoleQpole[H_,s_] := Module[{den,fp,Qp},
  den = Denominator[H];
  fp = Sqrt[Coefficient[den,s,0]/Coefficient[den,s,2]]/(2*Pi);
  Qp = (Coefficient[den,s,2]/Coefficient[den,s,1])*(2*Pi*fp);
  Simplify[{fp, Qp}]];
ZeroQzero[H_,s_] := Module[{fz,num,Qz0},
  num = Numerator[H];
  Qz0 = (Coefficient[num,s,2]/Coefficient[num,s,1]);
  fz = Sqrt[Coefficient[num,s,0]/Coefficient[num,s,2]]/(2*Pi);
  Simplify[{fz, Qz0*fz}]];
Sensitivity[F_,x_] := (x/F)*D[F,x];
GSP[F_,A_] := Limit[A*Sensitivity[F,A],A->Infinity]//Simplify;
GSPEpsA[F_,epsA_] := -(1/epsA)*Sensitivity[F,epsA]//Together;
PrintLabeledList[expressions_List,labels_List] := Map[
  Print[#[[1]], " = ",#[[2]]]&
  ,Transpose[{labels,expressions}]
];

```

■ A.6.7 Poles, Zeros, Q-Factors

```

{fp,Qp} = Simplify[PoleQpole[H,s]];
fp0 = Together[Limit[fp,A-> Infinity]];
Qp0 = Limit[Qp,A-> Infinity];
PrintLabeledList[{fp0,Qp0},{"fp","Qp"}];

```


$$fp = \frac{\sqrt{\frac{R4}{C6 C8 R3 R5 R7}}}{2 \pi}$$

$$Qp = \frac{C6 (R1 + R2) R3 R5 \sqrt{\frac{R4}{C6 C8 R3 R5 R7}}}{R1 (R3 + R4)}$$

■ A.6.8 Gain-Sensitivity Product (GSP)

```
fpepsA = Together[fp /. A->1/e];
QpepsA = Together[Qp /. A->1/e];
GSPQp = Simplify[GSPepsA[QpepsA,e] /. e->0];
GSPfp = Simplify[GSPepsA[fpepsA,e] /. e->0];
PrintLabeledList[{GSPfp,GSPQp},{ "GSPfp", "GSPQp"}];
```

$$GSPfp = \frac{-(R1 R3)^2 + 2 R1 R3 R4 + 3 R2 R3 R4 + R1 R4^2 + R2 R4^2}{2 R3 (R1 R4 + R2 R4)}$$

$$GSPQp = \frac{(2 C6 R1^2 R3^2 R4 R5 + 4 C6 R1 R2 R3^2 R4 R5 + 2 C6 R2^2 R3^2 R4 R5 - C8 R1^2 R3^2 R7 - C8 R1^2 R3 R4 R7 + 3 C8 R1 R2 R3^2 R4 R7 + 2 C8 R2^2 R3^2 R4 R7 - 3 C8 R1^2 R3 R4^2 R7 - 2 C8 R1 R2 R3 R4^2 R7 - C8 R1^2 R4^2 R7 - C8 R1 R2 R4^2 R7) / (2 R3 (R1 R4 + R2 R4) (C8 R1 R3 R7 + C8 R1 R4 R7))}{1}$$

■ A.6.9 Design

■ Find Element Values

```
DesignGPHQ[Qp_,wp_,Cx_,Rx_] :=
Module[{C6,C8,R1,R2,R3,R4,R5,R7,Ro},
C6 = Cx;
C8 = Cx;
Ro = 1/(wp*Cx);
R1 = Rx;
R3 = Rx;
R5 = Rx;
R7 = Rx;
R4 = Rx*(Rx/Ro)^2;
R2 = Rx*(Qp*(1+R4/Rx)/Sqrt[R4/Rx]-1);
{R1,R2,R3,R4,R5,C6,R7,C8};
{R1,R2,R3,R4,R5,C6,R7,C8} = DesignGPHQ[Q,Wp,Co,Rd];
```

■ Design Examples

```
values = {Q -> 6., Wp -> 2*Pi*1500.,
, Co -> 68.*10^(-9), Rd -> 1800.} //N;
gspQpfp = Together[Abs[GSPQp/2]+Abs[Q*GSPfp] /. values];
```

```
PrintLabeledList[N[{Q,Wp/(2*Pi),gspQpfp} /. values]
,{"Qp","fp (Hz)","GSP"} ];
Print["-----"]
PrintLabeledList[Together[{R1,R2,R3,R4,R5,C6*10^9,R7,C8*10^9} /. values]
,{"R1 (ohm)","R2 (ohm)","R3 (ohm)","R4 (ohm)","R5 (ohm)","C6 (nF)"
,"R7 (ohm)","C8 (nF)"}];
```

Qp = 6.

fp (Hz) = 1500.

GSP = 17.1412

R1 (ohm) = 1800.

R2 (ohm) = 20020.9

R3 (ohm) = 1800.

R4 (ohm) = 2395.4

R5 (ohm) = 1800.

C6 (nF) = 68.

R7 (ohm) = 1800.

C8 (nF) = 68.

values = {Q -> 25., Wp -> 2*Pi*1500.

, Co -> 22.*10^(-9), Rd -> 4700.} //N;

gspQpfp = Together[Abs[GSPQp/2]+Abs[Q*GSPfp] /. values];

PrintLabeledList[N[{Q,Wp/(2*Pi),gspQpfp} /. values]

Print["-----"]

PrintLabeledList[Together[{R1,R2,R3,R4,R5,C6*10^9,R7,C8*10^9} /. values]

,"R7 (ohm)","C8 (nF)"}];

Qp = 25.

fp (Hz) = 1500.

GSP = 74.014

R1 (ohm) = 4700.

R2 (ohm) = 230378.

R3 (ohm) = 4700.

R4 (ohm) = 4463.56

R5 (ohm) = 4700.

C6 (nF) = 22.

R7 (ohm) = 4700.

C8 (nF) = 22.

■ A.6.10 Optimization

■ Find Rd for Low Gain-Sensitivity Product

values = {Q -> 25., Wp -> 2*Pi*1500.

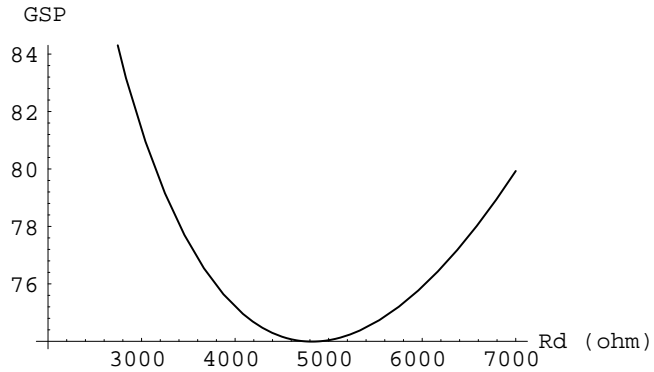
, Co -> 22.*10^(-9)} //N;

gspQpfp1 = Together[Abs[GSPQp/2.] + Abs[Q*GSPfp] /. values] //Simplify;

```

r1 = 2000;
r2 = 7000;
Plot[{gspQpfp1}
, {Rd, r1, r2}
, AxesLabel -> {"Rd (ohm)", "GSP"}
];

```



```

{Gspmin, Rset} = FindMinimum[gspQpfp1, {Rd, r1, r2}]

{73.99, {Rd -> 4822.22}}

values = {Q -> 25., Wp -> 2*Pi*1500.
, Co -> 22.*10^(-9)} //N;
gspQpfp = Together[Abs[GSPQp/2]+Abs[Q*GSPfp] /. values /. Rset];
PrintLabeledList[N[{Q,Wp/(2*Pi),gspQpfp} /. values/. Rset]
, {"Qp", "fp (Hz)", "GSP"} ];
Print["-----"]
PrintLabeledList[Together[{R1,R2,R3,R4,R5,C6*10^9,R7,C8*10^9} /. values/.Rset]
, {"R1 (ohm)", "R2 (ohm)", "R3 (ohm)", "R4 (ohm)", "R5 (ohm)", "C6 (nF)"
, "R7 (ohm)", "C8 (nF)"}];

Qp = 25.
fp (Hz) = 1500.
GSP = 73.99
-----
R1 (ohm) = 4822.22
R2 (ohm) = 236289.
R3 (ohm) = 4822.22
R4 (ohm) = 4820.9
R5 (ohm) = 4822.22
C6 (nF) = 22.
R7 (ohm) = 4822.22
C8 (nF) = 22.

```

```

H1p = Ha /. values /. Rset//Together;
num = Numerator[H1p];
den = Denominator[H1p];
numlist = CoefficientList[num,s];
denlist = CoefficientList[den,s];
H1p = 1/denlist[[3]] * num/(den/denlist[[3]])//Factor

```

$$\frac{1.74124 \cdot 10^8}{8.88264 \cdot 10^7 + 376.991 s + 1. s^2}$$

```

Hhp = Hha /. values /. Rset//Together;
num = Numerator[Hhp];
den = Denominator[Hhp];
numlist = CoefficientList[num,s];
denlist = CoefficientList[den,s];
Hhp = 1/denlist[[3]] * num/(den/denlist[[3]])//Factor

```

$$\frac{1.95973 s^2}{8.88264 \cdot 10^7 + 376.991 s + 1. s^2}$$

```

Hbp = Hba /. values /. Rset//Together;
num = Numerator[Hbp];
den = Denominator[Hbp];
numlist = CoefficientList[num,s];
denlist = CoefficientList[den,s];
Hbp = 1/denlist[[3]] * num/(den/denlist[[3]])//Factor

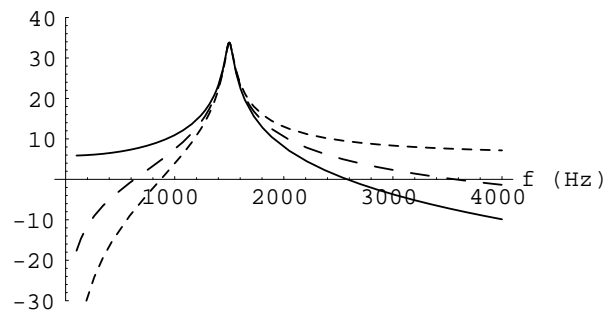
```

$$\frac{-18472.6 s}{8.88264 \cdot 10^7 + 376.991 s + 1. s^2}$$

```

AH1p = 20*Log[10,Abs[H1p]] /. s -> N[I*2*Pi*f];
AHhp = 20*Log[10,Abs[Hhp]] /. s -> N[I*2*Pi*f];
AHbp = 20*Log[10,Abs[Hbp]] /. s -> N[I*2*Pi*f];
Plot[{AH1p, AHhp, AHbp}
, {f, 100, 4000}
, PlotRange -> {-30,40}
, PlotStyle -> {Dashing[{}],
                Dashing[ {.02}],
                Dashing[ {.04}]}
, AxesLabel -> {"f (Hz)", "20*log M(f)"}
];

```

$20 \cdot \log M(f)$ 

A.7 Advanced Analog Filter Design Case Studies

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■ A.7.1 References

1. D.V. Tasic, M.D.Lutovac, B.L.Evans,
 "Advanced filter design,"
 Proc. IEEE Asilomar Conf. Signal, Systems, Computer,
 Nov. 1997, pp.710–715.
2. M. D. Lutovac, D. V. Tasic and B. L. Evans,
 "Advanced Filter Design for Signal Processing using MATLAB and Mathematica,"
<http://iritel.iritel.bg.ac.yu/~lutovac/www/afdhome.htm>
<http://galeb.etf.bg.ac.yu/~tasic/afdhome.htm>
<http://www.ece.utexas.edu/~bevans/>

■ A.7.2 Initialization

```
SetDirectory[HomeDirectory[]];
<<afd\math\m\clearall.m
<<graphics\graphics'
```

■ A.7.3 Notation

a – selectivity factor
 ap – maximum passband loss, dB, of realized filter
 Ap – maximum passband loss, dB, in specification
 as – minimum stopband loss, dB, of realized filter
 As – minimum stopband loss, dB, in specification
 A2a(n,Ap,As) – minimum selectivity factor from attenuation spec
 A2K(A) – characteristic function in terms of attenuation in dB
 e – ripple factor
 fp – passband edge [Hz] of realized filter
 Fp – passband edge [Hz] in specification
 fs – stopband edge [Hz] of realized filter
 Fs – stopband edge [Hz] in specification
 Ke(n,a,e,x) – characteristic function
 L(n,a) – discrimination factor
 n – order
 nbut(Fp,Fs,Ap,As) minimum Butterworth order from specification
 ncheb(Fp,Fs,Ap,As) minimum Chebyshev order from specification
 nellip(Fp,Fs,Ap,As) minimum order from specification
 q(k) – modular constant
 R(n,a,x) – elliptic rational function
 S(n,a,e) – list of transfer function poles
 S(n,a,e,i) – transfer function pole

SA – attenuation specification
 SK – characteristic function specification
 X(n,a) – list of zeros of elliptic rational function
 X(n,a,i) – zero of elliptic rational function
 Z(n,a,e) – zeta function

■ A.7.4 Definitions and Procedures

```
X[n_Integer, a_, i_Integer] := -JacobiCD[
  (2*i-1)*EllipticK[1/a^2]/n, 1/a^2
];
X[n_Integer, a_, i_Integer] := 0 /; And[i==(n+1)/2, OddQ[n]];
X[n_Integer, a_] := X[n, a, #]& /@ Range[n];
L[n_Integer, a_] := Block[{i, r},
  If[EvenQ[n],
    r = (1/a^n)*Product[(a^2 - X[n, a, i]^2)^2, {i, n/2}]/
      Product[(1 - X[n, a, i]^2)^2, {i, n/2}];,
    r = (1/a^(n-2))*Product[(a^2 - X[n, a, i]^2)^2, {i, (n-1)/2}]/
      Product[(1 - X[n, a, i]^2)^2, {i, (n-1)/2}];
  ];
  r
];
R[n_Integer, a_, x_] := Block[{i, r, r0},
  If[EvenQ[n],
    r = Product[x^2 - X[n, a, i]^2, {i, n/2}]/
      Product[x^2 - a^2/X[n, a, i]^2, {i, n/2}];
    r0 = Product[1 - X[n, a, i]^2, {i, n/2}]/
      Product[1 - a^2/X[n, a, i]^2, {i, n/2}];,
    r = x*Product[x^2 - X[n, a, i]^2, {i, (n-1)/2}]/
      Product[x^2 - a^2/X[n, a, i]^2, {i, (n-1)/2}];
    r0 = Product[1 - X[n, a, i]^2, {i, (n-1)/2}]/
      Product[1 - a^2/X[n, a, i]^2, {i, (n-1)/2}];
  ];
  r/r0
];
Z[n_Integer, a_, e_] := JacobiSN[
  InverseJacobiSN[1/Sqrt[1+e^2], 1-1/(L[n, a])^2]*
  EllipticK[1-1/a^2]/EllipticK[1-1/(L[n, a])^2], 1-1/a^2
];
S[n_Integer, a_, e_, i_Integer] := Block[
  {den, num, numim, numre, x, z},
  x = X[n, a, i];
  z = Z[n, a, e];
  numre = -z*Sqrt[1 - z^2]*Sqrt[1 - x^2]*Sqrt[1 - x^2/a^2];
  numim = x*Sqrt[1 - (1-1/a^2)*z^2];
  num = numre + I*numim;
```

```

den = 1 - (1 - x^2/a^2)*z^2;
num/den
];
S[n_Integer, a_, e_] := S[n,a,e,#]& /@ Range[n];
A2K[A_] := Sqrt[1 - 10^(-A/10)]/10^(-A/20);
Ke[n_Integer, a_, e_, x_] := e*Abs[R[n,a,x]];

```

■ A.7.5 Specification

```

SA = {3000., 3225., 0.2, 40.};
{Fp, Fs, Ap, As} = SA;
Kp = A2K[Ap];
Ks = A2K[As];
SK = {Fp, Fs, Kp, Ks}

{3000., 3225., 0.217091, 99.995}

```

■ A.7.6 Minimum Order

```

nellip[Fp_,Fs_,Ap_,As_] := Block[
  {num, den,
   k = Fp/Fs,
   L = Sqrt[(-1 + 10^(As/10))/(-1 + 10^(Ap/10))]},
  num = EllipticK[1-L^2]/EllipticK[1/L^2];
  den = EllipticK[1-k^2]/EllipticK[k^2];
  Ceiling[num/den//N]
];
ncheb[Fp_,Fs_,Ap_,As_] := Block[
  {L = Sqrt[(-1 + 10^(As/10))/(-1 + 10^(Ap/10))]},
  aspect = Fs/Fp,
  Ceiling[ArcCosh[L]/ArcCosh[aspect]//N]
];
nbut[Fp_,Fs_,Ap_,As_] := Block[
  {L = Sqrt[(-1 + 10^(As/10))/(-1 + 10^(Ap/10))]},
  aspect = Fs/Fp,
  Ceiling[Log[10,L]/Log[10,aspect]//N]
];
{nellip[Fp,Fs,Ap,As], ncheb[Fp,Fs,Ap,As], nbut[Fp,Fs,Ap,As]}
nmin = nellip[Fp,Fs,Ap,As];
nmax = 2*nmin;
nlist = Range[nmin,nmax]

{8, 18, 85}
{8, 9, 10, 11, 12, 13, 14, 15, 16}

```


■ A.7.7 Range of Selectivity Factor and Ripple Factor

```

q[k_] := Block[{c,e,r,s,t},
  If[k<=1/Sqrt[2.0],
    t = (1/2)*(1 - (1-k^2)^(1/4))/(1 + (1-k^2)^(1/4));,
    t = (1/2)*(1 - Sqrt[k])/(1 + Sqrt[k]);
  ];
  e = {1,5, 9, 13, 17, 21, 25, 29, 33, 37};
  c = {1,2,15,150,1707,20910,268616,3567400,48555069,673458874};
  s = Sum[c[[i]]*(t^e[[i]]),{i,Length[e]}];
  If[k<=1/Sqrt[2.0],
    r = s;,
    r = Exp[Pi^2/Log[s]];
  ];
  N[r]
];

A2a[n_,Ap_,As_] := Block[
  {m, num, den, terms=9, L, qL},
  L = Sqrt[(-1 + 10^(As/10))/(-1 + 10^(Ap/10))];
  qL = q[1/L]^(1/n);
  num = 1 + 2*Sum[(-1)^m*(qL)^(m^2), {m,1,terms}];
  den = 1 + 2*Sum[(qL)^(m^2), {m,1,terms}];
  1/Sqrt[1 - (num/den)^4]
];

amin8 = A2a[nmin,Ap,As];
amax8 = a/. FindRoot[R[nmin,a,Fs/Fp]==Ks/Kp, {a,Fs/Fp,1.1}];
amin9 = A2a[9,Ap,As];
amax9 = a/. FindRoot[R[9,a,Fs/Fp]==Ks/Kp, {a,amax8,1.1}];
amin10 = A2a[10,Ap,As];
amax10 = a/. FindRoot[R[10,a,Fs/Fp]==Ks/Kp, {a,amax9,1.2}];
amin11 = A2a[11,Ap,As];
amax11 = a/. FindRoot[R[11,a,Fs/Fp]==Ks/Kp, {a,amax10,1.2}];
amin12 = A2a[12,Ap,As];
amax12 = a/. FindRoot[R[12,a,Fs/Fp]==Ks/Kp, {a,amax11,1.2}];
amin13 = A2a[13,Ap,As];
amax13 = a/. FindRoot[R[13,a,Fs/Fp]==Ks/Kp, {a,amax12,1.3}];
amin14 = A2a[14,Ap,As];
amax14 = a/. FindRoot[R[14,a,Fs/Fp]==Ks/Kp, {a,amax13,1.4}];
amin15 = A2a[15,Ap,As];
amax15 = a/. FindRoot[R[15,a,Fs/Fp]==Ks/Kp, {a,amax14,1.6}];
amin16 = A2a[16,Ap,As];
amax16 = a/. FindRoot[R[16,a,Fs/Fp]==Ks/Kp, {a,amax15,1.9}];
aminlist = {amin8,amin9,amin10,amin11,amin12,amin13,amin14,amin15,amin16};
amaxlist = {amax8,amax9,amax10,amax11,amax12,amax13,amax14,amax15,amax16};
eminlist = Table[Ks/L[n,amaxlist[[n-8+1]]], {n,nmin,nmax}];
{emin8,emin9,emin10,emin11,emin12,emin13,emin14,emin15,emin16} = eminlist;
emax = Kp;
emaxlist = Table[Kp,{nmax-nmin+1}];

```

```
TableForm[Transpose[{nlist,aminlist,amaxlist,eminlist,emaxlist}]
, TableHeadings->{{},{}, {"n","amin","amax","emin","emax"}}]
```

n	amin	amax	emin	emax
8	1.04285	1.08323	0.0757872	0.217091
9	1.022	1.09807	0.0184689	0.217091
10	1.01135	1.12013	0.00368669	0.217091
11	1.00587	1.15172	0.000578651	0.217091
12	1.00304	1.19663	0.0000676012	0.217091
-6				
13	1.00158	1.26158	5.39161 10	0.217091
-7				
14	1.00082	1.35951	2.53319 10	0.217091
-9				
15	1.00042	1.51914	5.28479 10	0.217091
-11				
16	1.00022	1.82219	2.51893 10	0.217091

■ A.7.8 Range of Edge Frequencies

```
fpminlist = Fs / amaxlist;
{fpmin8,fpmin9,fpmin10,fpmin11,fpmin12,fpmin13,
 fpmin14,fpmin15,fpmin16} = fpminlist;
fpmaxlist = Fs / aminlist;
{fpmax8,fpmax9,fpmax10,fpmax11,fpmax12,fpmax13,
 fpmax14,fpmax15,fpmax16} = fpmaxlist;
fsminlist = Fp * aminlist;
{fsmin8,fsmin9,fsmin10,fsmin11,fsmin12,fsmin13,
 fsmin14,fsmin15,fsmin16} = fsminlist;
fsmaxlist = Fp * amaxlist;
{fsmax8,fsmax9,fsmax10,fsmax11,fsmax12,fsmax13,
 fsmax14,fsmax15,fsmax16} = fsmaxlist;
TableForm[Transpose[{nlist,fpminlist,fpmaxlist,fsminlist,fsmaxlist}]
, TableHeadings->{{},{}, {"n","fpmin","fpmax","fsmin","fsmax (Hz)"}}]
```

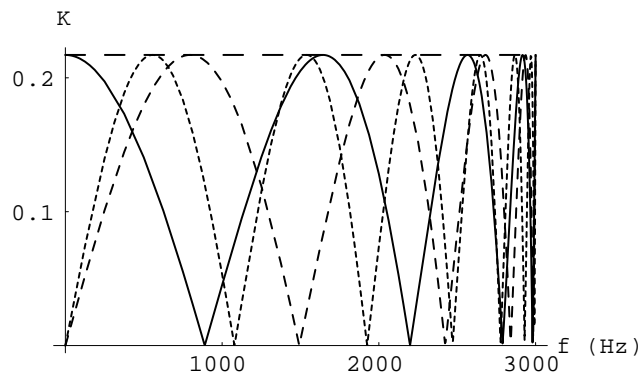
n	fpmin	fpmax	fsmin	fsmax (Hz)
8	2977.2	3092.49	3128.55	3249.7
9	2936.97	3155.57	3066.01	3294.21
10	2879.14	3188.79	3034.06	3360.38
11	2800.16	3206.17	3017.62	3455.16
12	2695.06	3215.22	3009.13	3589.89
13	2556.32	3219.92	3004.73	3784.74
14	2372.18	3222.36	3002.45	4078.53
15	2122.91	3223.63	3001.27	4557.42
16	1769.85	3224.29	3000.66	5466.56

■ A.7.9 Design D1

```

Plot[Evaluate[{Kp
    , Ke[ 8, Fs/Fp, emax, f/Fp]
    , Ke[ 9, Fs/Fp, emax, f/Fp]
    , Ke[13, Fs/Fp, emax, f/Fp]
}], {f, 0, Fp}
, AxesLabel -> {"f (Hz)", "K"}
, PlotStyle -> {Dashing[{0.04}],
    Dashing[{}],
    Dashing[{0.02}],
    Dashing[{0.01}]
}
, Ticks -> {{1000, 2000, 3000}, {0, 0.1, 0.2}}
, PlotRange -> All];

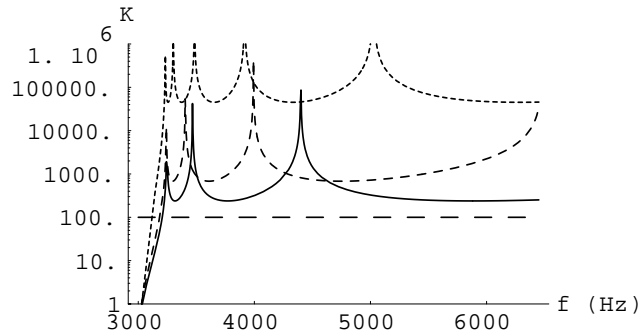
```



```

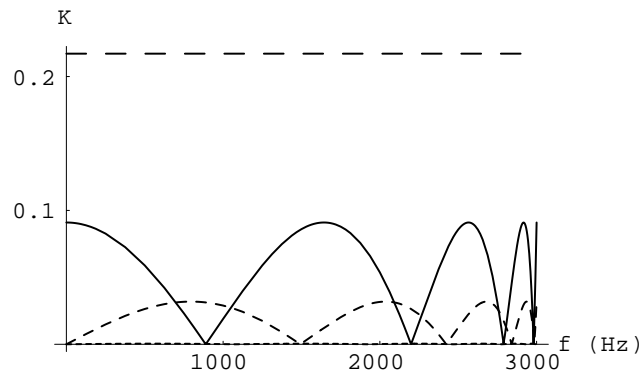
LogPlot[Evaluate[{Ks
    , Ke[ 8, Fs/Fp, emax, f/Fp]
    , Ke[ 9, Fs/Fp, emax, f/Fp]
    , Ke[13, Fs/Fp, emax, f/Fp]
}], {f, Fp, 2*Fs}
, AxesLabel -> {"f (Hz)", "K"}
, PlotStyle -> {Dashing[{0.04}],
    Dashing[{}],
    Dashing[{0.02}],
    Dashing[{0.01}]
}
, Ticks -> {{3000, 4000, 5000, 6000}, Automatic}
, PlotRange -> {1, 10^6}];

```



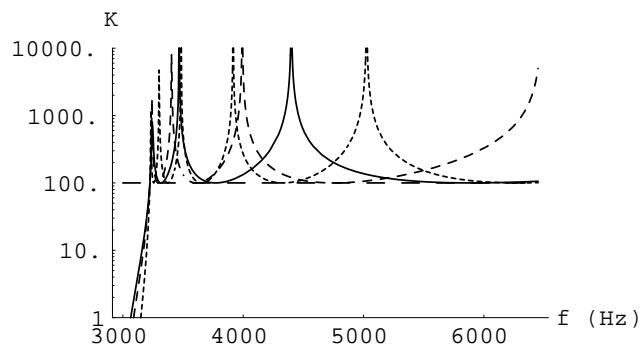
■ A.7.10 Design D2

```
Plot[Evaluate[{Kp
    , Ke[ 8, Fs/Fp, Ks/L[ 8,Fs/Fp], f/Fp]
    , Ke[ 9, Fs/Fp, Ks/L[ 9,Fs/Fp], f/Fp]
    , Ke[13, Fs/Fp, Ks/L[13,Fs/Fp], f/Fp]
}], {f,0,Fp}
, AxesLabel -> {"f (Hz)", "K"}
, PlotStyle -> {Dashing[{0.04}],
    Dashing[{}],
    Dashing[{0.02}],
    Dashing[{0.01}]
}
, Ticks -> {{1000, 2000, 3000}, {0, 0.1, 0.2}}
, PlotRange -> All];
```



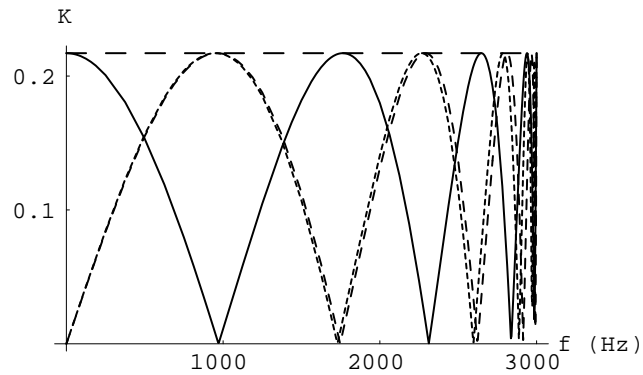
```
LogPlot[Evaluate[{Ks
    , Ke[ 8, Fs/Fp, Ks/L[ 8,Fs/Fp], f/Fp]
    , Ke[ 9, Fs/Fp, Ks/L[ 9,Fs/Fp], f/Fp]
    , Ke[13, Fs/Fp, Ks/L[13,Fs/Fp], f/Fp]
}], {f,Fp,2*Fs}
```

```
, AxesLabel -> {"f (Hz)", "K"}
, PlotStyle -> {Dashing[{0.04}],
               Dashing[{}],
               Dashing[{0.02}],
               Dashing[{0.01}]
              }
, Ticks -> {{3000, 4000, 5000, 6000}, Automatic}
, PlotRange -> {1, 10^4}];
```

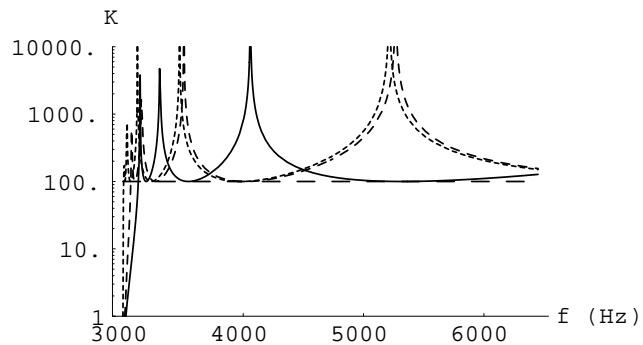


■ A.7.11 Design D3a

```
Plot[Evaluate[{Kp
               , Ke[ 8, amin8,  emax, f/Fp]
               , Ke[ 9, amin9,  emax, f/Fp]
               , Ke[13, amin13, emax, f/Fp]
              }],
      {f, 0, Fp}
, AxesLabel -> {"f (Hz)", "K"}
, PlotStyle -> {Dashing[{0.04}],
               Dashing[{}],
               Dashing[{0.02}],
               Dashing[{0.01}]
              }
, Ticks -> {{1000, 2000, 3000}, {0, 0.1, 0.2}}
, PlotRange -> All];
```

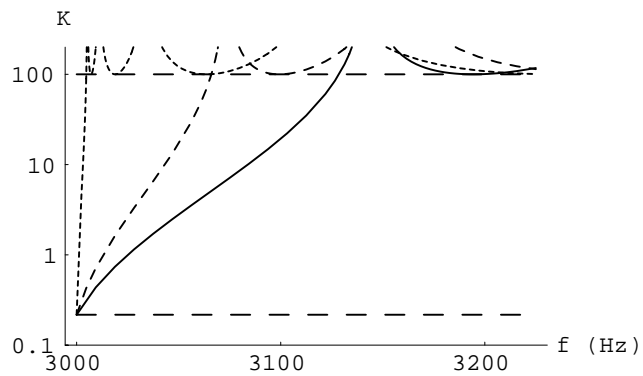


```
LogPlot[Evaluate[{Ks
  , Ke[ 8, amin8,  emax, f/Fp]
  , Ke[ 9, amin9,  emax, f/Fp]
  , Ke[13, amin13, emax, f/Fp]
}], {f, Fp, 2*Fs}
, AxesLabel -> {"f (Hz)", "K"}
, PlotStyle -> {Dashing[{0.04}],
  Dashing[{}],
  Dashing[{0.02}],
  Dashing[{0.01}]
}
, Ticks -> {{3000, 4000, 5000, 6000}, Automatic}
, PlotRange -> {1, 10^4}];
```



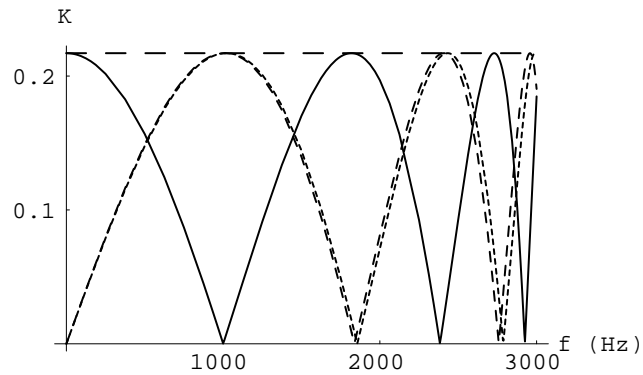
```
LogPlot[Evaluate[{Kp, Ks
  , Ke[ 8, amin8,  emax, f/Fp]
  , Ke[ 9, amin9,  emax, f/Fp]
  , Ke[13, amin13, emax, f/Fp]
}], {f, Fp, Fs}
, AxesLabel -> {"f (Hz)", "K"}]
```

```
, PlotStyle -> {Dashing[{0.04}],
                Dashing[{0.04}],
                Dashing[{}],
                Dashing[{0.02}],
                Dashing[{0.01]}
                }
, Ticks -> {{3000, 3100, 3200}, {0.1, 1, 10, 100}}
, PlotRange -> {0.1, 200}];
```

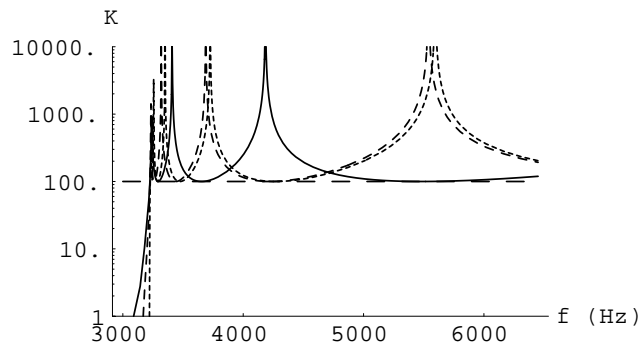


■ A.7.12 Design D3b

```
Plot[Evaluate[{Kp
                , Ke[ 8, amin8,  emax, f/fpmax8]
                , Ke[ 9, amin9,  emax, f/fpmax9]
                , Ke[13, amin13, emax, f/fpmax13]
                }],
, {f, 0, Fp}
, AxesLabel -> {"f (Hz)", "K"}
, PlotStyle -> {Dashing[{0.04}],
                Dashing[{}],
                Dashing[{0.02}],
                Dashing[{0.01]}
                }
, Ticks -> {{1000, 2000, 3000}, {0, 0.1, 0.2}}
, PlotRange -> All];
```



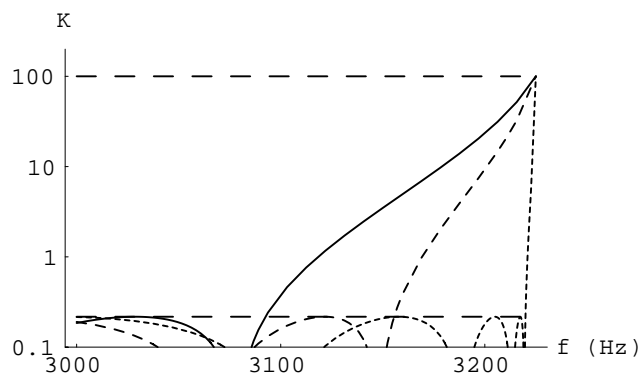
```
LogPlot[Evaluate[{Ks
  , Ke[ 8, amin8,  emax, f/fpmax8]
  , Ke[ 9, amin9,  emax, f/fpmax9]
  , Ke[13, amin13, emax, f/fpmax13]
}], {f, Fp, 2*Fs}
, AxesLabel -> {"f (Hz)", "K"}
, PlotStyle -> {Dashing[{0.04}],
  Dashing[{}],
  Dashing[{0.02}],
  Dashing[{0.01}]
}
, Ticks -> {{3000, 4000, 5000, 6000}, Automatic}
, PlotRange -> {1, 10^4}];
```



```
LogPlot[Evaluate[{Kp, Ks
  , Ke[ 8, amin8,  emax, f/fpmax8]
  , Ke[ 9, amin9,  emax, f/fpmax9]
  , Ke[13, amin13, emax, f/fpmax13]
}], {f, Fp, Fs}
, AxesLabel -> {"f (Hz)", "K"}]
```

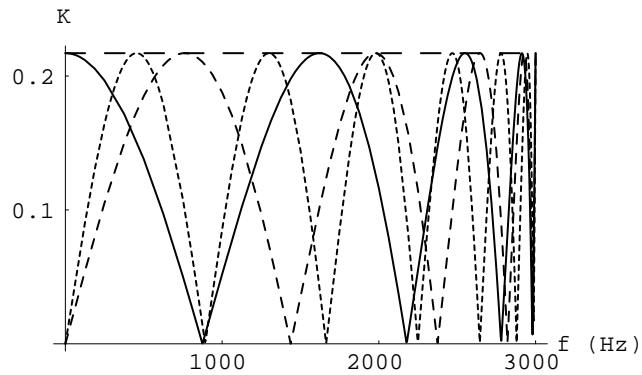


```
, PlotStyle -> {Dashing[{0.04}],
                Dashing[{0.04}],
                Dashing[{}],
                Dashing[{0.02}],
                Dashing[{0.01]}
                }
, Ticks -> {{3000, 3100, 3200}, {0.1, 1, 10, 100}}
, PlotRange -> {0.1, 200}];
```

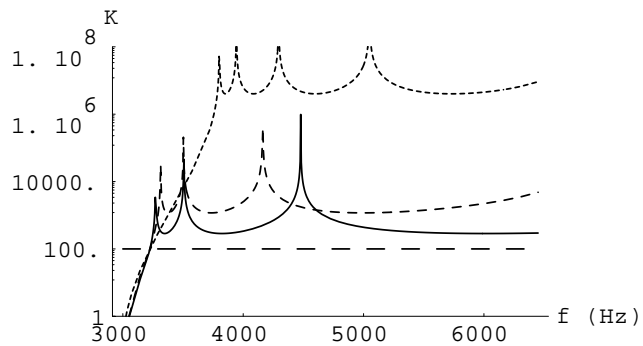


■ A.7.13 Design D4a

```
Plot[Evaluate[{Kp
                , Ke[ 8, amax8,  emax, f/Fp]
                , Ke[ 9, amax9,  emax, f/Fp]
                , Ke[13, amax13, emax, f/Fp]
                }],
{f, 0, Fp}
, AxesLabel -> {"f (Hz)", "K"}
, PlotStyle -> {Dashing[{0.04}],
                Dashing[{}],
                Dashing[{0.02}],
                Dashing[{0.01]}
                }
, Ticks -> {{1000, 2000, 3000}, {0, 0.1, 0.2}}
, PlotRange -> All];
```



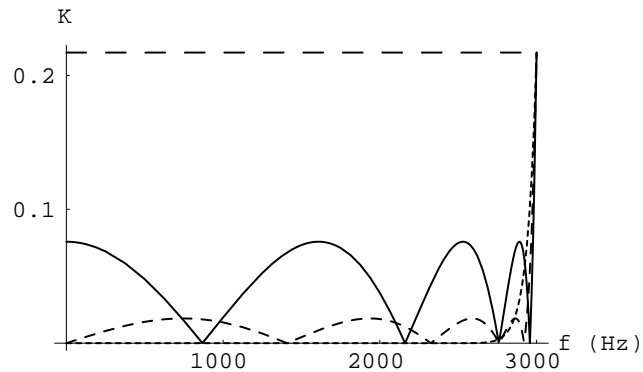
```
LogPlot[Evaluate[{Ks
  , Ke[ 8, amax8,  emax, f/Fp]
  , Ke[ 9, amax9,  emax, f/Fp]
  , Ke[13, amax13, emax, f/Fp]
}], {f, Fp, 2*Fs}
, AxesLabel -> {"f (Hz)", "K"}
, PlotStyle -> {Dashing[{0.04}],
  Dashing[{}],
  Dashing[{0.02}],
  Dashing[{0.01}]
}
, Ticks -> {{3000, 4000, 5000, 6000}, Automatic}
, PlotRange -> {1, 10^8}];
```



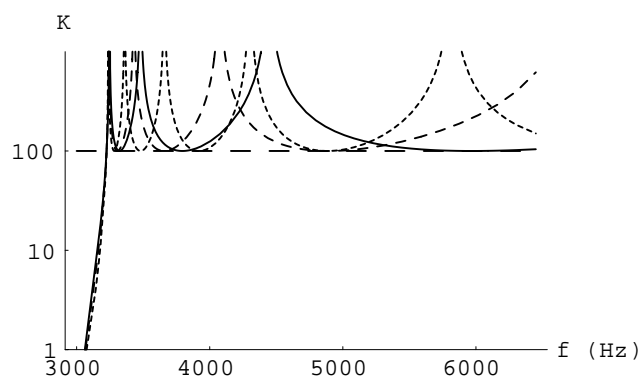
■ A.7.14 Design D4b

```
Plot[Evaluate[{Kp
  , Ke[ 8, amax8,  emin8,  f/(Fs/amax8)]
  , Ke[ 9, amax9,  emin9,  f/(Fs/amax9)]
  , Ke[13, amax13, emin13, f/(Fs/amax13)]
}], {f, 0, Fp}
```

```
, AxesLabel -> {"f (Hz)", "K"}
, PlotStyle -> {Dashing[{0.04}],
               Dashing[{}],
               Dashing[{0.02}],
               Dashing[{0.01}]
              }
, Ticks -> {{1000, 2000, 3000}, {0, 0.1, 0.2}}
, PlotRange -> All];
```



```
LogPlot[Evaluate[{Ks
                 , Ke[ 8, amax8, emin8, f/(Fs/amax8)]
                 , Ke[ 9, amax9, emin9, f/(Fs/amax9)]
                 , Ke[13, amax13, emin13, f/(Fs/amax13)]
                },
        {f, Fp, 2*Fs}
, AxesLabel -> {"f (Hz)", "K"}
, PlotStyle -> {Dashing[{0.04}],
               Dashing[{}],
               Dashing[{0.02}],
               Dashing[{0.01}]
              }
, Ticks -> {{3000, 4000, 5000, 6000}, {1, 10, 100}}
, PlotRange -> {1, 10^3}];
```



A.8 Classical Digital Filters

Transpose Direct Form II

IIR Second-Order Realization

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■ A.8.1 References

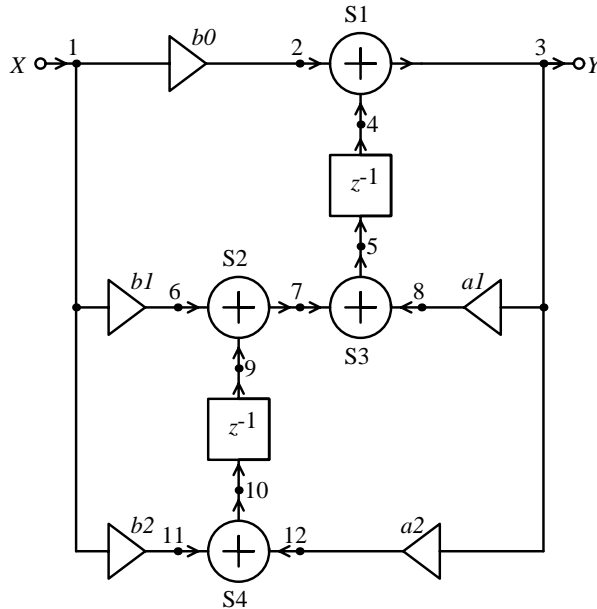
1. Alan Oppenheim, Ronald Schaffer, "Digital Signal Processing,"
Prentice-Hall, Englewood Cliffs, New Jersey, 1975.
2. Sanjit Mitra, James Kaiser, "Handbook for Digital Signal Processing,"
John Wiley, New York, 1993, pp. 127–128.
3. M. D. Lutovac, D. V. Tasic and B. L. Evans,
"Advanced Filter Design for Signal Processing using MATLAB and Mathematica,"
<http://iritel.iritel.bg.ac.yu/~lutovac/www/afdhome.htm>
<http://galeb.etf.bg.ac.yu/~tasic/afdhome.htm>
<http://www.ece.utexas.edu/~bevans/>

■ A.8.2 Initialization

```
SetDirectory[HomeDirectory[]];
<<afd\math\m\clearall.m
<<afd\math\m\drawdfil.m
<<afd\math\m\drawiirf.m
```

■ A.8.3 Block Diagram

`DrawTDF2[0,0,1,1/0.8,10];`



■ A.8.4 Analysis ($v=1/z$)

```

ElementEquations = {
  Y1 == X
, Y2 == b0*Y1 + Xb0
, Y3 == Y2 + Y4
, Y4 == v*Y5
, Y5 == Y7 + Y8
, Y6 == b1*Y1 + Xb1
, Y7 == Y6 + Y9
, Y8 == a1*Y3 + Xa1
, Y9 == v*Y10
, Y10 == Y11 + Y12
, Y11 == b2*Y1 + Xb2
, Y12 == a2*Y3 + Xa2
};
NodeSignals = {Y1,Y2,Y3,Y4,Y5,Y6,Y7,Y8,Y9,Y10,Y11,Y12};
Response = Flatten[Solve[ElementEquations,NodeSignals]];
Y = Together[Y3/.Response];
Hv = Y /. {X -> 1, Xa1 -> 0, Xa2 -> 0, Xb0 -> 0, Xb1 -> 0, Xb2 -> 0};
Hz = Hv /. v -> 1/z //Together;
Print["H(v) = ", Hv]
Print["H(z) = ", Hz]

```

$$H(v) = \frac{-b_0 - b_1 v - b_2 v^2}{-1 + a_1 v + a_2 v^2}$$

$$H(z) = \frac{-b_2 - b_1 z - b_0 z^2}{a_2 + a_1 z - z^2}$$

■ A.8.5 Transfer Function and Noise Transfer Functions

```

Ha1v = Y /. {X -> 0, Xa1 -> 1, Xa2 -> 0, Xb0 -> 0, Xb1 -> 0, Xb2 -> 0};
Ha2v = Y /. {X -> 0, Xa1 -> 0, Xa2 -> 1, Xb0 -> 0, Xb1 -> 0, Xb2 -> 0};
Hb0v = Y /. {X -> 0, Xa1 -> 0, Xa2 -> 0, Xb0 -> 1, Xb1 -> 0, Xb2 -> 0};
Hb1v = Y /. {X -> 0, Xa1 -> 0, Xa2 -> 0, Xb0 -> 0, Xb1 -> 1, Xb2 -> 0};
Hb2v = Y /. {X -> 0, Xa1 -> 0, Xa2 -> 0, Xb0 -> 0, Xb1 -> 0, Xb2 -> 1};
H = Collect[Numerator[Hv],v]/Collect[Denominator[Hv],v];
Ha1 = Collect[Numerator[Ha1v],v]/Collect[Denominator[Ha1v],v];
Ha2 = Collect[Numerator[Ha2v],v]/Collect[Denominator[Ha2v],v];
Hb0 = Collect[Numerator[Hb0v],v]/Collect[Denominator[Hb0v],v];
Hb1 = Collect[Numerator[Hb1v],v]/Collect[Denominator[Hb1v],v];
Hb2 = Collect[Numerator[Hb2v],v]/Collect[Denominator[Hb2v],v];
v2invz = {v->z^"-1", v^2->z^("-2"), v^3->z^("-3"), v^4->z^("-4")};
Print["H(z) = ", H /. v2invz ]
Print["Ha1(z) = ", Ha1 /. v2invz ]
Print["Ha2(z) = ", Ha2 /. v2invz ]
Print["Hb0(z) = ", Hb0 /. v2invz ]
Print["Hb1(z) = ", Hb1 /. v2invz ]
Print["Hb2(z) = ", Hb2 /. v2invz ]

```

$$H(z) = \frac{-b_0 - b_1 z^{-1} - b_2 z^{-2}}{-1 + a_1 z^{-1} + a_2 z^{-2}}$$

$$Ha1(z) = -\left(\frac{-b_0 - b_1 z^{-1} - b_2 z^{-2}}{-1 + a_1 z^{-1} + a_2 z^{-2}}\right)$$

$$Ha2(z) = -\left(\frac{-b_0 - b_1 z^{-1} - b_2 z^{-2}}{-1 + a_1 z^{-1} + a_2 z^{-2}}\right)$$

$$Hb0(z) = -\left(\frac{1}{-1 + a_1 z^{-1} + a_2 z^{-2}}\right)$$

$$Hb1(z) = -\left(\frac{z}{-1 + a_1 z^{-1} + a_2 z^{-2}}\right)$$

$$Hb2(z) = -\left(\frac{z^2}{-1 + a_1 z^{-1} + a_2 z^{-2}}\right)$$

■ A.8.6 VQNR

Variance of Quantization Noise Due to Rounding the Output of the Multiplier

```
VQNR[H_,z_Symbol,a_Symbol,b_Symbol
,r_Symbol,theta_Symbol] := Module[
{ax, bx, d0, d1, d2, denH2, H0, H0inv, H2, numH2
,res0, res1, res2, rtheta2ab, sumres, var=Infinity
,z1, z2},
H0 = Together[H] /. {a -> ax, b -> bx};
H0inv = Together[H0 /. z->1/z];
H2 = Together[Cancel[z^2*H0]];
numH2 = Collect[Numerator[H2],z];
denH2 = Collect[Denominator[H2],z];
If[Exponent[denH2,z] == 2
,{d0,d1,d2} = CoefficientList[denH2,z];
{z1,z2} = Flatten[Solve[denH2==0,z]];
res0 = D[H2*H0inv,{z,2}]/2 /. z->0;
res1 = numH2*H0inv/(d2*(z-(z/.z2))*z^3) /. z1;
res2 = numH2*H0inv/(d2*(z-(z/.z1))*z^3) /. z2;
sumres = Simplify[res0 + res1 + res2];
rtheta2ab = Solve[{d0 == d2*(r^2),
d1 == d2*(-2*r*Cos[theta])}
,{ax,bx}] //Flatten;
var = Together[sumres /. rtheta2ab];
Print["Error in denominator! ", denH2];
];
var]
Haz = Together[Ha2 /. v -> 1/z]
vara = Simplify[VQNR[Haz,z,a1,a2,r,theta], Trig -> True]
```

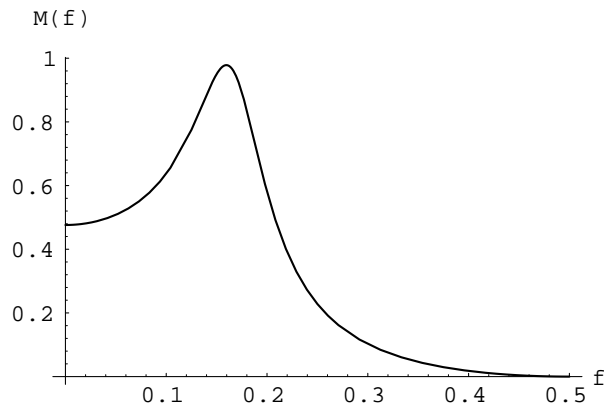

$$-\frac{1}{a_2 + a_1 z - z^2} \frac{-1 - r^2}{(-1 + r^2)(1 + r^4 - 2r^2 \cos[2\theta])}$$

■ A.8.7 Frequency Response ($z=\text{Exp}[I*2*\text{Pi}*f]$)

```

values = {a1 -> 0.8, a2 -> -.64, b0 -> 1/10, b1 -> 2/10, b2 -> 1/10};
Hf = Hz /. z->Exp[I*2*Pi*f] /. values //N;
Plot[Evaluate[Abs[Hf]]
, {f, 0, 1/2}
, AxesLabel -> {"f", "M(f)"}
];

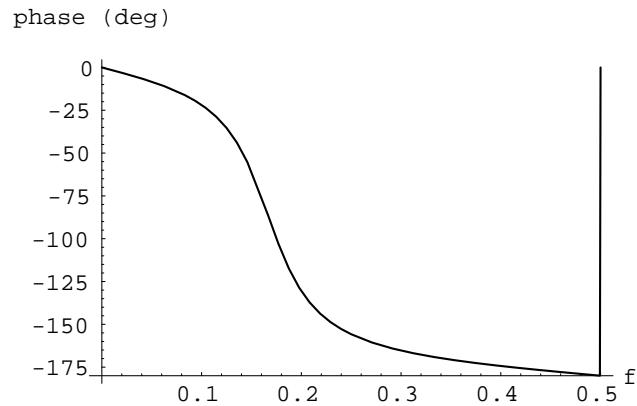
```



```

Plot[(Evaluate[Arg[Hf]])/Degree
, {f, 0, 1/2}
, AxesLabel -> {"f", "phase (deg)"}
, AxesOrigin -> {0, -180}
, PlotRange -> All
];

```



■ A.8.8 Dynamics

Magnitude Response of Partial Transfer Functions

```

excitation = {X -> 1, Xa1 -> 0, Xa2 -> 0, Xb0 -> 0, Xb1 -> 0, Xb2 -> 0};
values = {a1 -> 0.8, a2 -> -.64, b0 -> 1/10, b1 -> 2/10, b2 -> 1/10};
H7 = Together[Y7/.Response] /. excitation;
H7z = H7 /. v -> 1/z //Together;
H7f = H7z /. z->Exp[I*2*Pi*f] /. values //N;

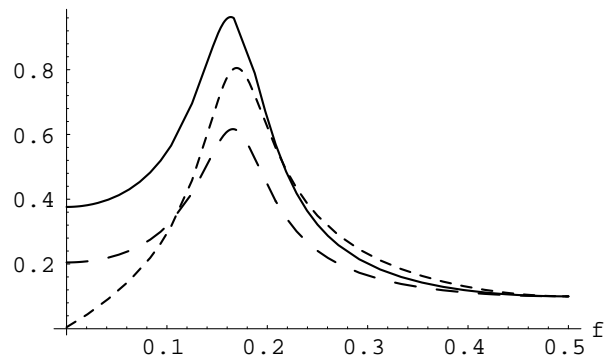
H5 = Together[Y5/.Response] /. excitation;
H5z = H5 /. v -> 1/z //Together;
H5f = H5z /. z->Exp[I*2*Pi*f] /. values //N;

H10 = Together[Y10/.Response] /. excitation;
H10z = H10 /. v -> 1/z //Together;
H10f = H10z /. z->Exp[I*2*Pi*f] /. values //N;

Plot[Evaluate[
  {Abs[H5f], Abs[H7f], Abs[H10f]}]
, {f, 0, 1/2}
, AxesLabel -> {"f", "M5, M7, M10"}
, PlotRange -> All
, PlotStyle -> {Dashing[{}],
  , Dashing[ {.02} ]
  , Dashing[ {.04} ]
}
];

```

M5, M7, M10

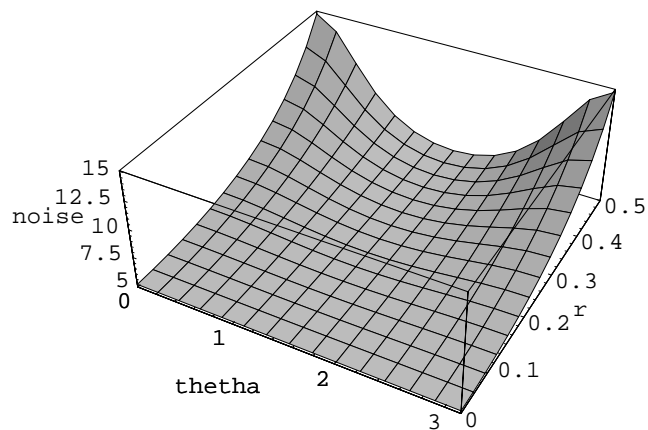


■ A.8.9 Relative Output Noise as a Function of Pole Position Due to All 5 Multipliers

```

valuesb = {b0 -> 1/10, b1 -> 2/10, b2 -> 1/10};
Plot3D[5*(vara /. valuesb)
, {theta, 0, Pi}
, {r, 0, 0.5}
, AxesLabel -> {"theta", "r", "noise"}
, PlotRange -> All];

```



■ Remark

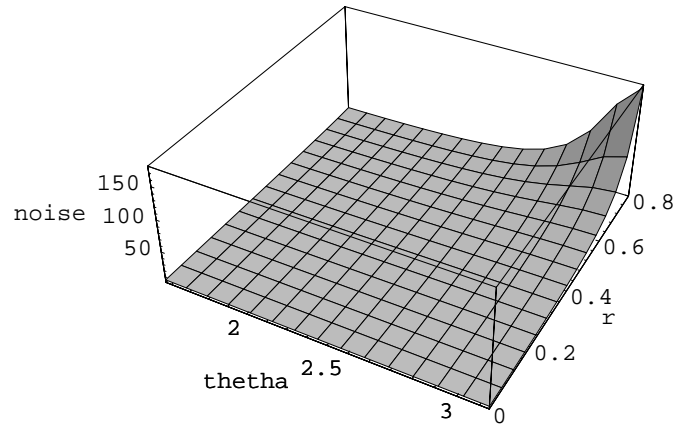
In this example all multipliers have equal noise variance, i.e. sum of variances is equal to $6 \cdot \text{vara}$, where vara represents the noise variance due to the multiplier a_2 .

```

valuesb = {b0 -> 1/10, b1 -> 2/10, b2 -> 1/10};
Plot3D[5*(vara /. valuesb)
, {theta, Pi/2, Pi}
, {r, 0, .8}

```

```
, AxesLabel -> {"theta", "r", "noise"}
, PlotRange -> All
(*, Ticks -> {{Pi/2, Pi}, {0, 1}, {1, 2, 3, 4}}*)
];
```



■ A.8.10 Filtering and Quantization

■ Filtering without Quantization

```
IIRTF2[a1_, a2_, b0_, b1_, b2_, X_, d1_:0, d2_:0] := Module[
{vY5, vY10, Y, Y1, Y2, Y3, Y4, Y5, Y6, Y7, Y8, Y9, Y10, Y11, Y12},
vY5 = d1;
vY10 = d2;
Y1 = X;
Y4 = vY5;
Y9 = vY10;
Y2 = b0*Y1;
Y6 = b1*Y1;
Y11 = b2*Y1;
Y3 = Y2 + Y4;
Y8 = a1*Y3;
Y12 = a2*Y3;
Y7 = Y6 + Y9;
Y5 = Y7 + Y8;
Y10 = Y11 + Y12;
Y = Y3;
{Y, Y5, Y10}];
```

■ Filtering with Quantization

```
Quantize[x_, b_:8, q_:0] := Module[{},
If[q == 0
, y = Round[x*2^b]/2^b;
, y = Floor[2^nm*x]/2^nm ];
y];
```

```

QIIRTF2[a1_, a2_, b0_, b1_, b2_, X_, d1_:0, d2_:0, bits_:8, mode_:0] := Module[
{vY5, vY10, Y, Y1, Y2, Y3, Y4, Y5, Y6, Y7, Y8, Y9, Y10, Y11, Y12},
  vY5 = d1;
  vY10 = d2;
  Y1 = X;
  Y4 = vY5;
  Y9 = vY10;
  Y2 = Quantize[b0*Y1, bits, mode];
  Y6 = Quantize[b1*Y1, bits, mode];
  Y11 = Quantize[b2*Y1, bits, mode];
  Y3 = Y2 + Y4;
  Y8 = Quantize[a1*Y3, bits, mode];
  Y12 = Quantize[a2*Y3, bits, mode];
  Y7 = Y6 + Y9;
  Y5 = Y7 + Y8;
  Y10 = Y11 + Y12;
  Y = Y3;
  {Y, Y5, Y10}];

```

■ Filtering

```

filterQIIRTF2[a1_, a2_, b0_, b1_, b2_, Xdata_List, d1_:0, d2_:0,
  , bits_Integer:8, mode_Integer:0] := Module[
{n, Qd1, Qd2, QYdata, QYn, QY3, Xn},
  Qd1 = d1;
  Qd2 = d2;
  QYdata = {};
  Do[Xn = Xdata[[n]];
    QY3 = QIIRTF2[a1, a2, b0, b1, b2, Xn, Qd1, Qd2, bits, mode];
    {QYn, Qd1, Qd2} = QY3;
    AppendTo[QYdata, QYn];
  , {n, 1, Length[Xdata]}];
  {QYdata, Qd1, Qd2}];

```

■ Example of Filtering

```

values = {a1 -> 0.8, a2 -> -0.64, b0 -> 0.1, b1 -> 0.2, b2 -> 0.1};
Hv /. values
{aD1, aD2, bD0, bD1, bD2} = {a1, a2, b0, b1, b2} /. values
Ndata = 100; (* number of samples *)
Xdata = Table[0, {Ndata}]; Xdata[[1]] = 1;
{QYdata, Qd1, Qd2} = filterQIIRTF2[aD1, aD2, bD0, bD1, bD2, Xdata, 0, 0, 8, 0]/N;
{Ydata, Qd1, Qd2} = filterQIIRTF2[aD1, aD2, bD0, bD1, bD2, Xdata, 0, 0, 12, 0]/N;

```

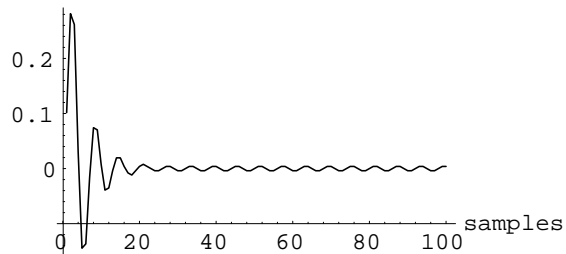
$$\frac{-0.1 - 0.2 v - 0.1 v^2}{-1 + 0.8 v - 0.64 v^2}$$

{0.8, -0.64, 0.1, 0.2, 0.1}

■ Impulse Response

```
ListPlot[QYdata
, PlotJoined->True
, PlotRange -> All
, AxesOrigin -> {.1,-.1}
, AxesLabel -> {"samples", "impulse response"}
];
```

impulse response



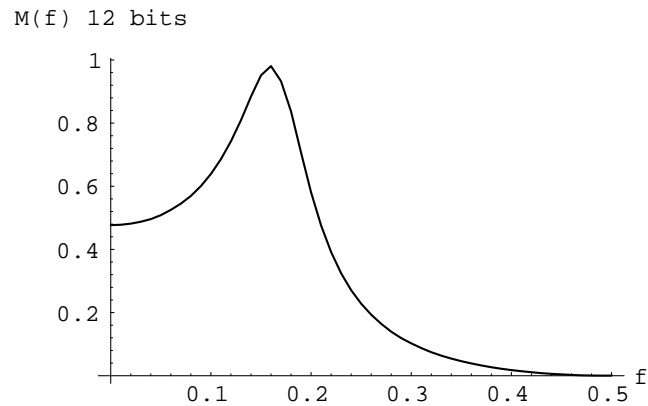
■ Spectrum of Impulse Response Define Electrical Engineering FFT

```
EEfft[data_List] := InverseFourier[data]*Sqrt[Length[data]];
EEifft[data_List] := Fourier[data]/Sqrt[Length[data]];
spectrum = EEfft[Ydata] //Chop;
qspectrum = EEfft[N[QYdata]] //Chop;
```

■ A.8.11 Verify Frequency Response by Spectrum of Impulse Response

■ 12 Bits Filtering Quantization

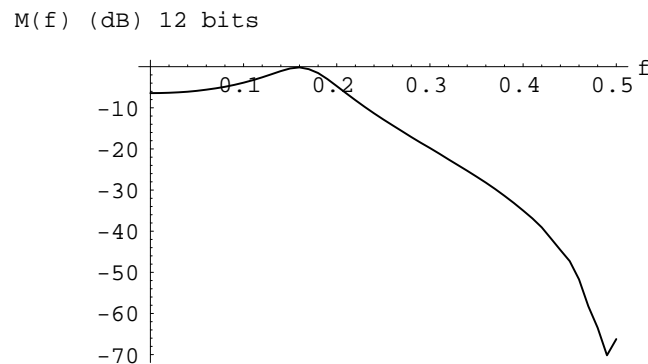
```
pfs = Table[{(k-1)/Length[spectrum], Abs[spectrum[[k]]]}
, {k,1,Length[spectrum]/2+1}];
ListPlot[pfs
, PlotJoined -> True
, PlotRange -> All
, AxesLabel -> {"f", "M(f) 12 bits"}
];
```



```

pfsdB = Table[{(ind-1)/Length[spectrum], 20*Log[10,Abs[spectrum[[ind]]]]}
, {ind,1,Length[spectrum]/2+1}];
ListPlot[pfsdB
, PlotJoined -> True
, PlotRange -> All
, AxesLabel -> {"f", "M(f) (dB) 12 bits"}
];

```



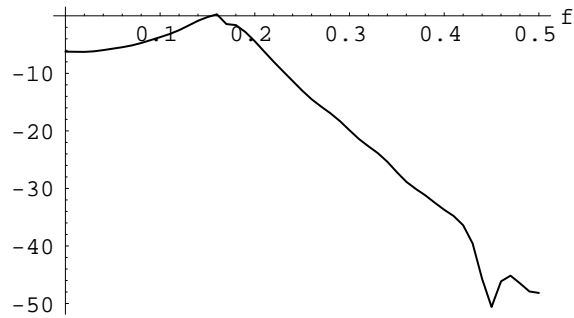
■ 8 Bits Filtering Quantization

```

pfqsdB = Table[{(ind-1)/Length[qspectrum], 20*Log[10,Abs[qspectrum[[ind]]]]}
, {ind,1,Length[qspectrum]/2+1}];
ListPlot[pfqsdB
, PlotJoined -> True
, PlotRange -> All
, AxesLabel -> {"f", "M(f) (dB) 8 bits"}
];

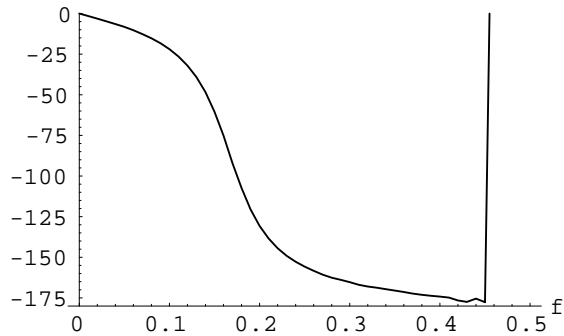
```

M(f) (dB) 8 bits



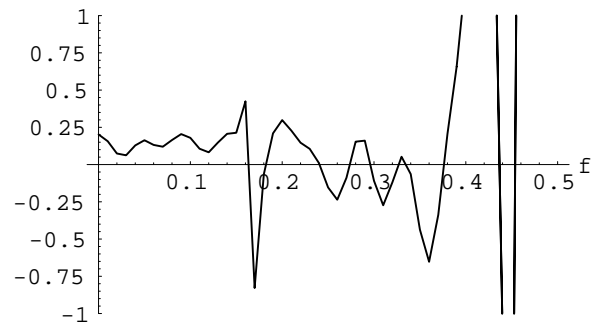
```
pfp = Table[{(ind-1)/Length[spectrum], Arg[spectrum[[ind]]]/Degree}
, {ind,1,Length[qspectrum]/2+1}];
ListPlot[pfp
, PlotJoined -> True
, PlotRange -> {-180,0}
, AxesOrigin -> {0,-180}
, AxesLabel -> {"f", "phase (deg) 8 bits"}
];
```

phase (deg) 8 bits



■ Error in Magnitude Response Due to Quantization

```
deltapfs = Table[{(k-1)/Length[spectrum]
, 20*Log[10,Abs[qspectrum[[k]]]] - 20*Log[10,Abs[spectrum[[k]]]]}
, {k,1,Length[spectrum]/2+1}];
ListPlot[deltapfs
, PlotJoined -> True
, PlotRange -> {-1,1}
, AxesLabel -> {"f", "error in M(f) (dB)"}
];
```


error in $M(f)$ (dB)

A.9 Advanced Digital Filter Design

Case Studies

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 lutovac@iritel.bg.ac.yu tasic@galeb.etf.bg.ac.yu bevans@ece.utexas.edu

■ A.9.1 References

1. D.V. Tasic, M.D.Lutovac, B.L.Evans,
 "Advanced filter design,"
 Proc. IEEE Asilomar Conf. Signal, Systems, Computer,
 Nov. 1997, pp.710–715.
2. D.V.Tasic, M.D.Lutovac, B.L.Evans,
 "Advanced digital filter design,"
 Proc. European Conf. Circuit Theory Design ECCTD'99, Stresa, Italy,
 Sep.1999,vol.2, pp.1323–1326.
3. M. D. Lutovac, D. V. Tasic and B. L. Evans,
 "Advanced Filter Design for Signal Processing using MATLAB and Mathematica,"
<http://iritel.iritel.bg.ac.yu/~lutovac/www/afdhome.htm>
<http://galeb.etf.bg.ac.yu/~tasic/afdhome.htm>
<http://www.ece.utexas.edu/~bevans/>

■ A.9.2 Initialization

```
SetDirectory[HomeDirectory[]];
<<afd\math\m\clearall.m
<<afd\math\m\dfde11ip.m
<<graphics\graphics'
```

■ A.9.3 Notation

```
!!afd\math\m\dfdnotat.m
```

a – selectivity factor
 AdB – attenuation (dB) in terms of digital frequency f
 ap – maximum passband attenuation, (dB), of realized filter
 Ap – maximum passband attenuation, (dB), in specification
 as – minimum stopband attenuation, (dB), of realized filter
 As – minimum stopband attenuation, (dB), in specification
 A2a(n,Ap,As) – minimum selectivity factor from attenuation spec
 A2K(A) – characteristic function in terms of attenuation in dB
 ba – coefficients of the second-order section
 e – ripple factor
 f – digital frequency ($0 < f < 0.5$)
 fp – passband edge ($0 < fp < 0.5$) of realized filter
 Fp – passband edge ($0 < Fp < 0.5$) in specification
 fs – stopband edge ($0 < fs < 0.5$) of realized filter
 Fs – stopband edge ($0 < Fs < 0.5$) in specification

hz – transfer function in z
 Hz(n,a,e,Fp,z) – transfer function in z
 Ke(n,a,e,x) – characteristic function
 L(n,a) – discrimination factor
 n – order
 nbut(Fp,Fs,Ap,As) – minimum Butterworth order from specification
 ncheb(Fp,Fs,Ap,As) – minimum Chebyshev order from specification
 nellip(Fp,Fs,Ap,As) – minimum elliptic order from specification
 nminQ – minimum order of elliptic minimal Q-factor design
 q(k) – modular constant
 Qfactor – Qfactor of the second-order section
 rtan(f1,f2) – $\tan(\pi f_1)/\tan(\pi f_2)$
 R(n,a,x) – elliptic rational function
 S(n,a,e) – list of transfer function poles
 S(n,a,e,i) – transfer function pole
 SA – attenuation-limit specification
 SK – characteristic-function-limit specification
 xtan – $\tan(\pi F_s)/\tan(\pi F_p)$
 X(n,a) – list of zeros of elliptic rational function
 X(n,a,i) – zero of elliptic rational function
 z – complex variable
 Z(n,a,e) – zeta function
 Zbl(Spole,Fp) – bilinear transformation

■ A.9.4 Definitions and Procedures

```

!!afd\math\m\dfde11ip.m
(* DFDELLIP.M
  7:06PM 9/11/98
*)

A2a[n_,Ap_,As_] := Module[
  {m, num, den, terms=9, L, qL},
  L = Sqrt[(-1 + 10^(As/10))/(-1 + 10^(Ap/10))];
  qL = q[1/L]^(1/n);
  num = 1 + 2*Sum[(-1)^m*(qL)^(m^2), {m,1,terms}];
  den = 1 + 2*Sum[(qL)^(m^2), {m,1,terms}];
  1/Sqrt[1 - (num/den)^4]
];

A2K[A_] := Sqrt[1 - 10^(-A/10)]/10^(-A/20);

AdB[hz_,z_,f_] := -20*Log[10,Abs[hz /. z-> N[E^(I*2*Pi*f)]]];

ba[n_,a_,e_,Fp_,z_] := Module[{i,g,t},
  {-2*Re[Zbl[S[n,a,e,1],Fp]], Abs[Zbl[S[n,a,e,1],Fp]]^2}
];

DGDelay[A_,B_,z_,f_] := Module[

```

```

{Ai,Bi,gA,gAd,gAn,nA,tAd,tAn,gB,gBd,gBn,nB,tBd,tBn},
  Ai = CoefficientList[A,z];
  Bi = CoefficientList[B,z];
  nA = Length[Ai];
  tAn = Join[{Sum[(i-1)*Ai[[i]]^2,{i,1,nA}]],
    Table[Cos[(k-1) w]*Sum[(2*i-k-1)*Ai[[i]]*Ai[[i-k+1]],{i,k,nA}]],{k,2,nA}]];
  tAd = Join[{Sum[Ai[[i]]^2,{i,1,nA}]],
    Table[2*Cos[(k-1) w]*Sum[Ai[[i]]*Ai[[i-k+1]],{i,k,nA}]],{k,2,nA}]];
  gAn = Sum[tAn[[i]],{i,1,nA}];
  gAd = Sum[tAd[[i]],{i,1,nA}];
  gA = Simplify[gAn/gAd];
  nB = Length[Bi];
  tBn = Join[{Sum[(i-1)*Bi[[i]]^2,{i,1,nB}]],
    Table[Cos[(k-1) w]*Sum[(2*i-k-1)*Bi[[i]]*Bi[[i-k+1]],{i,k,nB}]],{k,2,nB}]];
  tBd = Join[{Sum[Bi[[i]]^2,{i,1,nB}]],
    Table[2*Cos[(k-1) w]*Sum[Bi[[i]]*Bi[[i-k+1]],{i,k,nB}]],{k,2,nB}]];
  gBn = Sum[tBn[[i]],{i,1,nB}];
  gBd = Sum[tBd[[i]],{i,1,nB}];
  gB = Simplify[gBn/gBd];
  gB - gA + nA - nB /. z -> E^(I*2*N[Pi]*f) /. w -> 2*N[Pi]*f
];

Hz[n_,a_,e_,Fp_,z_] := Module[{i,g,t},
  If[EvenQ[n],
    g = Product[(2-2*Re[Zbl[I*a/X[n,a,i],Fp]])/
      (1-2*Re[Zbl[S[n,a,e,i],Fp]]
      +Abs[Zbl[S[n,a,e,i],Fp]]^2)
      ,{i,1,n/2}]*Sqrt[1+e^2];
    t = (1/g)*Product[(z^2-2*Re[Zbl[I*a/X[n,a,i],Fp]]*z+1)/
      (z^2-2*Re[Zbl[S[n,a,e,i],Fp]]*z
      +Abs[Zbl[S[n,a,e,i],Fp]]^2)
      ,{i,1,n/2}];,
    g = 2*Product[(1-2*Re[Zbl[I*a/X[n,a,i],Fp]]+1)/
      (1-2*Re[Zbl[S[n,a,e,i],Fp]]
      +Abs[Zbl[S[n,a,e,i],Fp]]^2)
      ,{i,1,(n-1)/2}]/(1-Zbl[S[n,a,e,(n+1)/2],Fp]);
    t = (1/g)*(z+1)*Product[(z^2-2*Re[Zbl[I*a/X[n,a,i],Fp]]*z+1)/
      (z^2-2*Re[Zbl[S[n,a,e,i],Fp]]*z
      +Abs[Zbl[S[n,a,e,i],Fp]]^2)
      ,{i,1,(n-1)/2}]/(z-Zbl[S[n,a,e,(n+1)/2],Fp]);
  ];
  t
];

Ke[n_Integer, a_, e_, x_] := e*Abs[R[n,a,x]];

```

```

L[n_Integer,a_] := Module[{i,r},
  If[EvenQ[n],
    r = (1/a^n)*Product[(a^2 - X[n,a,i]^2)^2, {i,n/2}]/
      Product[(1 - X[n,a,i]^2)^2, {i,n/2}];,
    r = (1/a^(n-2))*Product[(a^2 - X[n,a,i]^2)^2, {i,(n-1)/2}]/
      Product[(1 - X[n,a,i]^2)^2, {i,(n-1)/2}];
  ];
r
];

nbut[Fp_,Fs_,Ap_,As_] := Module[
  {L = Sqrt[(-1 + 10^(As/10))/(-1 + 10^(Ap/10))],
  aspec = Tan[Pi*Fs]/Tan[Pi*Fp]},
  Ceiling[Log[10,L]/Log[10,aspec]]/N
];

ncheb[Fp_,Fs_,Ap_,As_] := Module[
  {L = Sqrt[(-1 + 10^(As/10))/(-1 + 10^(Ap/10))],
  aspec = Tan[Pi*Fs]/Tan[Pi*Fp]},
  Ceiling[ArcCosh[L]/ArcCosh[aspec]]/N
];

nellip[Fp_,Fs_,Ap_,As_] := Module[
  {num, den,
  k = Tan[Pi*Fp]/Tan[Pi*Fs],
  L = Sqrt[(-1 + 10^(As/10))/(-1 + 10^(Ap/10))]},
  num = EllipticK[1-L^2]/EllipticK[1/L^2];
  den = EllipticK[1-k^2]/EllipticK[k^2];
  Ceiling[num/den]/N
];

nminQ[Fp_, Fs_, Ap_, As_] := Block[{ai, i, Kp, Ks, x, x1, x2},
  i = nellip[Fp,Fs,Ap,As];
  Kp = A2K[Ap];
  Ks = A2K[As];
  x1 = Tan[Pi*Fs]/Tan[Pi*Fp];
  x2 = x1^2;
  ai = x /. FindRoot[Sqrt[L[i, x]] == Ks, {x,x1,x2}];
  While[Ke[i, ai, 1/Ks, Fp/(Fs/ai)] > Kp,
    i += 1;
    x2 = ai;
    ai = x /. FindRoot[Sqrt[L[i, x]] == Ks, {x,x1,x2}];
  ];
  i
];

```

```

plotstyle012345 := PlotStyle -> {Dashing[{}],
                                Dashing[{0.01}],
                                Dashing[{0.02}],
                                Dashing[{0.03}],
                                Dashing[{0.04}],
                                Dashing[{0.05}]}

plotPTS[MdB_,f_,spec_List] := Block[
{Fpass=spec[[1]], Fstop=spec[[2]],
Apass=spec[[3]], Astop=spec[[4]], c1, c2, g1 ,g2, g3, M},
M = MdB;
c1 = PlotStyle -> {RGBColor[1,0,0],RGBColor[1,0,0],RGBColor[0,0,1]};
c2 = PlotStyle -> {RGBColor[0,1,0],RGBColor[0,1,0],RGBColor[0,0,1]};
g1 = Plot[{Apass,0,M}, {f,0,Fpass}, Evaluate[c1]
, PlotRange -> All
, Ticks -> {{0, Fp},{0,Ap/2,Ap}}
, AxesOrigin->{-0.001,-0.01}];
g2 = Plot[{Apass,Astop,M}, {f,Fpass,Fstop}, Evaluate[c2]
, PlotRange -> All
, Ticks -> {{Fp,Fs},{Ap, As/2, As}}
, AxesOrigin->{Fpass-0.001,0}];
g3 = Plot[{Astop,Astop,M}, {f,Fstop,0.5}, Evaluate[c1]
, PlotRange -> {Astop,Astop+41}
, Ticks -> {{Fs,0.5},{As, As+20, As+40}}
, AxesOrigin->{Fstop-0.001,Astop-2}];
Show[GraphicsArray[{g1,g2,g3}
]];
];

q[k_] := Module[{c,e,r,s,t},
If[k<=1/Sqrt[2.0],
t = (1/2)*(1 - (1-k^2)^(1/4))/(1 + (1-k^2)^(1/4));,
t = (1/2)*(1 - Sqrt[k])/(1 + Sqrt[k]);
];
e = {1,5, 9, 13, 17, 21, 25, 29, 33, 37};
c = {1,2,15,150,1707,20910,268616,3567400,48555069,673458874};
s = Sum[c[[i]]*(t^e[[i]]),{i,Length[e]}];
If[k<=1/Sqrt[2.0],
r = s;,
r = Exp[Pi^2/Log[s]];
];
N[r]
];

Qfactor[a_,b_] := Sqrt[(1+a+b)*(1+a-b)]/(2*(1-a));
rtan[fnum_,fden_,prec_:16] := N[Tan[Pi*fnum]/Tan[Pi*fden],prec];

```

```

R[n_Integer, a_, x_] := Module[{i, r, r0},
  If[EvenQ[n],
    r = Product[x^2 - X[n, a, i]^2, {i, n/2}]/
      Product[x^2 - a^2/X[n, a, i]^2, {i, n/2}];
    r0 = Product[1 - X[n, a, i]^2, {i, n/2}]/
      Product[1 - a^2/X[n, a, i]^2, {i, n/2}];,
    r = x*Product[x^2 - X[n, a, i]^2, {i, (n-1)/2}]/
      Product[x^2 - a^2/X[n, a, i]^2, {i, (n-1)/2}];
    r0 = Product[1 - X[n, a, i]^2, {i, (n-1)/2}]/
      Product[1 - a^2/X[n, a, i]^2, {i, (n-1)/2}];
  ];
  r/r0
];

S[n_Integer, a_, e_] := S[n, a, e, #]& /@ Range[n];
S[n_Integer, a_, e_, i_Integer] := Module[
  {den, num, numim, numre, x, z},
  x = X[n, a, i];
  z = Z[n, a, e];
  numre = -z*Sqrt[1 - z^2]*Sqrt[1 - x^2]*Sqrt[1 - x^2/a^2];
  numim = x*Sqrt[1 - (1-1/a^2)*z^2];
  num = numre + I*numim;
  den = 1 - (1 - x^2/a^2)*z^2;
  num/den
];

X[n_Integer, a_] := X[n, a, #]& /@ Range[n];
X[n_Integer, a_, i_Integer] := -JacobiCD[
  (2*i-1)*EllipticK[1/a^2]/n, 1/a^2
];

Z[n_Integer, a_, i_Integer] := 0 /; And[i==(n+1)/2, OddQ[n]];
Z[n_Integer, a_, e_] := JacobiSN[
  InverseJacobiSN[1/Sqrt[1+e^2], 1-1/(L[n, a])^2]*
  EllipticK[1-1/a^2]/EllipticK[1-1/(L[n, a])^2], 1-1/a^2
];

Zbl[sp_, Fp_] := (1+sp*(Tan[Pi*Fp]))/(1-sp*(Tan[Pi*Fp]));

```

■ A.9.5 Specification

```

SA = {0.2, 0.212, 0.2, 40.};
{Fp, Fs, Ap, As} = SA;
Kp = A2K[Ap];
Ks = A2K[As];
SK = {Fp, Fs, Kp, Ks}

{0.2, 0.212, 0.217091, 99.995}

```

■ A.9.6 Minimum Order

```
{nellip[Fp,Fs,Ap,As], ncheb[Fp,Fs,Ap,As], nbut[Fp,Fs,Ap,As]}
nmin = nellip[Fp,Fs,Ap,As];
nmax = 2*nmin;
nlist = Range[nmin,nmax]
```

```
{8, 18, 79}
{8, 9, 10, 11, 12, 13, 14, 15, 16}
```

■ A.9.7 Range of Selectivity Factor and Ripple Factor

```
xtan = rtan[Fs,Fp];
amin8 = A2a[nmin,Ap,As];
amax8 = a/. FindRoot[R[nmin,a,xtan]==Ks/Kp, {a,xtan,1.095}];
amin9 = A2a[9,Ap,As];
amax9 = a/. FindRoot[R[ 9,a,xtan]==Ks/Kp, {a,amax8,1.1}];
amin10 = A2a[10,Ap,As];
amax10 = a/. FindRoot[R[10,a,xtan]==Ks/Kp, {a,amax9,1.2}];
amin11 = A2a[11,Ap,As];
amax11 = a/. FindRoot[R[11,a,xtan]==Ks/Kp, {a,amax10,1.2}];
amin12 = A2a[12,Ap,As];
amax12 = a/. FindRoot[R[12,a,xtan]==Ks/Kp, {a,amax11,1.2}];
amin13 = A2a[13,Ap,As];
amax13 = a/. FindRoot[R[13,a,xtan]==Ks/Kp, {a,amax12,1.3}];
amin14 = A2a[14,Ap,As];
amax14 = a/. FindRoot[R[14,a,xtan]==Ks/Kp, {a,amax13,1.4}];
amin15 = A2a[15,Ap,As];
amax15 = a/. FindRoot[R[15,a,xtan]==Ks/Kp, {a,amax14,1.6}];
amin16 = A2a[16,Ap,As];
amax16 = a/. FindRoot[R[16,a,xtan]==Ks/Kp, {a,amax15,1.9}];
aminlist = {amin8,amin9,amin10,amin11,amin12,amin13,amin14,amin15,amin16};
amaxlist = {amax8,amax9,amax10,amax11,amax12,amax13,amax14,amax15,amax16};
eminlist = Table[Ks/L[n,amaxlist[[n-8+1]]], {n,nmin,nmax}];
{emin8,emin9,emin10,emin11,emin12,emin13,emin14,emin15,emin16} = eminlist;
emax = Kp;
emaxlist = Table[Kp,{nmax-nmin+1}];
TableForm[Transpose[{nlist,aminlist,amaxlist
                     ,ScientificForm/@eminlist,emaxlist}]
, TableHeadings->{{},{}, {"n","amin","amax","emin","emax"}}]
```

n	amin	amax	emin	emax
			-2	
8	1.04285	1.09245	6.25782 10	0.217091
			-2	
9	1.022	1.11016	1.43176 10	0.217091
			-3	
10	1.01135	1.13668	2.62202 10	0.217091
			-4	
11	1.00587	1.17518	3.65611 10	0.217091

				-5	
12	1.00304	1.23116	3.61027	10	0.217091
				-6	
13	1.00158	1.31502	2.23759	10	0.217091
				-8	
14	1.00082	1.44884	6.96526	10	0.217091
				-10	
15	1.00042	1.69029	6.68381	10	0.217091
				-13	
16	1.00022	2.26906	4.39641	10	0.217091

■ A.9.8 Range of Edge Frequencies

```

fpminlist = ArcTan[Tan[Pi*Fs] / amaxlist]/Pi //N;
{fpmin8,fpmin9,fpmin10,fpmin11,fpmin12,fpmin13,
 fpmin14,fpmin15,fpmin16} = fpminlist;
fpmaxlist = ArcTan[Tan[Pi*Fs] / aminlist]/Pi //N;
{fpmax8,fpmax9,fpmax10,fpmax11,fpmax12,fpmax13,
 fpmax14,fpmax15,fpmax16} = fpmaxlist;
fsminlist = ArcTan[Tan[Pi*Fp] * aminlist]/Pi //N;
{fsmin8,fsmin9,fsmin10,fsmin11,fsmin12,fsmin13,
 fsmin14,fsmin15,fsmin16} = fsminlist;
fsmaxlist = ArcTan[Tan[Pi*Fp] * amaxlist]/Pi //N;
{fsmax8,fsmax9,fsmax10,fsmax11,fsmax12,fsmax13,
 fsmax14,fsmax15,fsmax16} = fsmaxlist;
TableForm[Transpose[{nlist,fpminlist,fpmaxlist,fsminlist,fsmaxlist}]
, TableHeadings->{{},{}, {"n","fpmin","fpmax","fsmin","fsmax"}}]

```

n	fpmin	fpmax	fsmin	fsmax
8	0.198485	0.205546	0.20639	0.213552
9	0.196065	0.208643	0.203305	0.216049
10	0.192535	0.210256	0.201712	0.21973
11	0.187606	0.211095	0.200887	0.224951
12	0.180824	0.21153	0.20046	0.23229
13	0.171447	0.211756	0.200239	0.242744
14	0.158187	0.211874	0.200124	0.258162
15	0.138517	0.211934	0.200064	0.282469
16	0.106119	0.211966	0.200033	0.326442

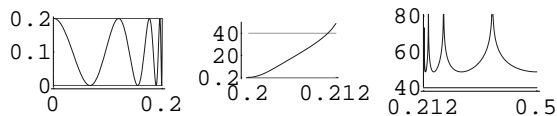
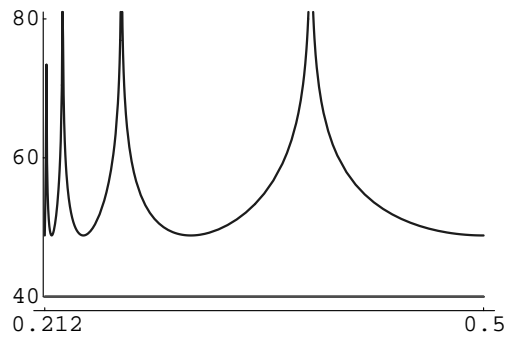
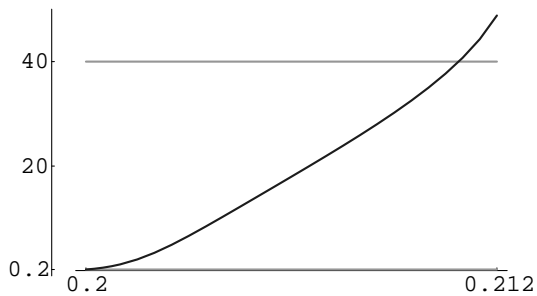
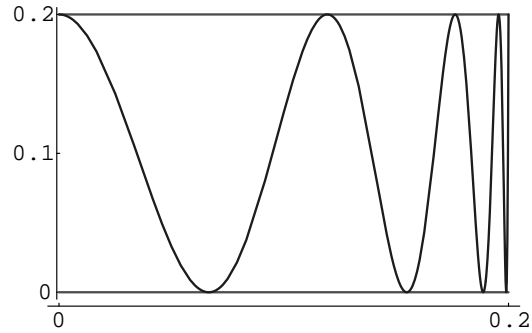
■ A.9.9 Transfer Functions and Plots of Attenuation in dB Versus Frequency

■ A.9.10 Design D1

```

hz8D1 = Hz[8, rtan[Fs,Fp], emax, Fp, z] //N;
AdB8D1 = AdB[hz8D1,z,f];
{b1n8,a1n8} = ba[8, rtan[Fs,Fp], emax, Fp, z] //N;
q1n8 = Qfactor[a1n8,b1n8];
plotPTS[AdB8D1,f,{0.2,0.212,0.2,40}]

```

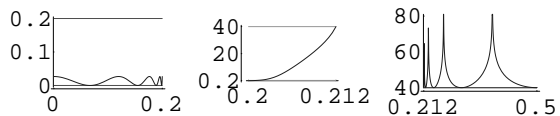
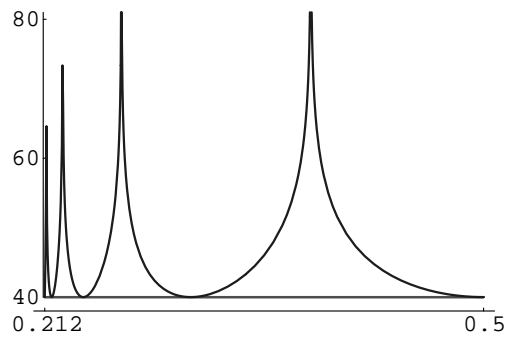
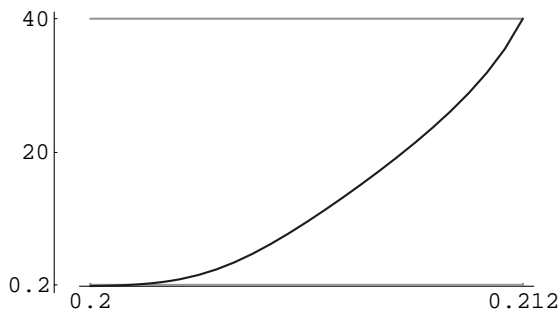
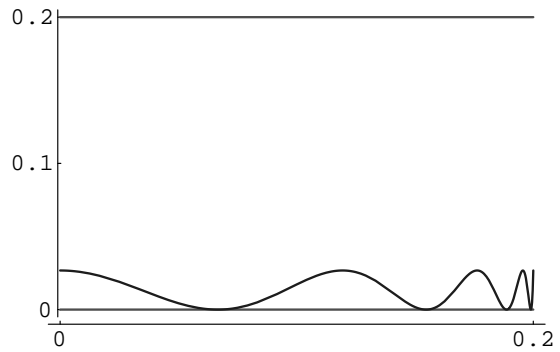


■ A.9.11 Design D2

```

hz8D2 = Hz[8, rtan[Fs,Fp], Ks/L[8,rtan[Fs,Fp]], Fp, z] //N;
AdB8D2 = AdB[hz8D2,z,f];
{b2n8,a2n8} = ba[8, rtan[Fs,Fp], Ks/L[8,rtan[Fs,Fp]], Fp, z] //N;
q2n8 = Qfactor[a2n8,b2n8];
plotPTS[AdB8D2,f,{0.2,0.212,0.2,40}]

```

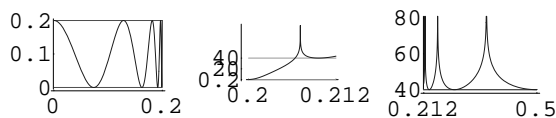
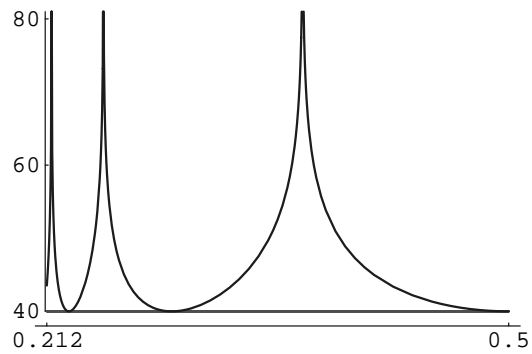
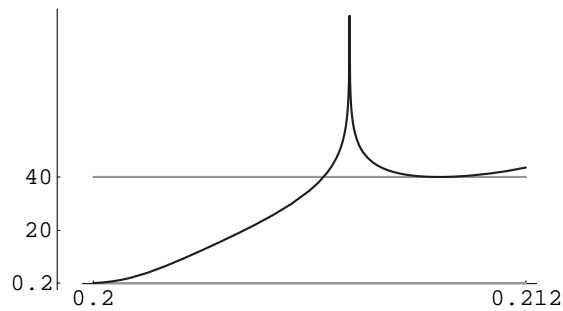
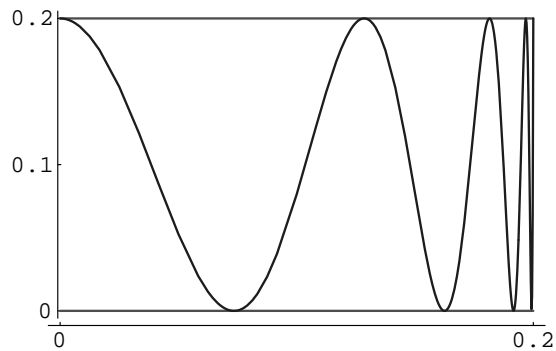


■ A.9.12 Design D3a

```

hz8D3a = Hz[8, amin8, emax, Fp, z] //N;
AdB8D3a = AdB[hz8D3a,z,f];
{b3an8,a3an8} = ba[8, amin8, emax, Fp, z] //N;
q3an8 = Qfactor[a3an8,b3an8];
plotPTS[AdB8D3a,f,{0.2,0.212,0.2,40}]

```

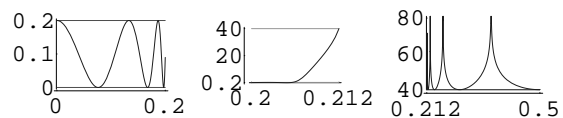
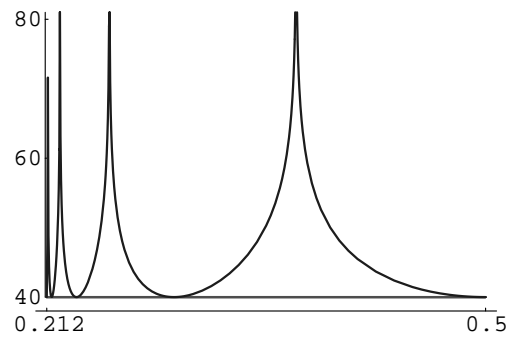
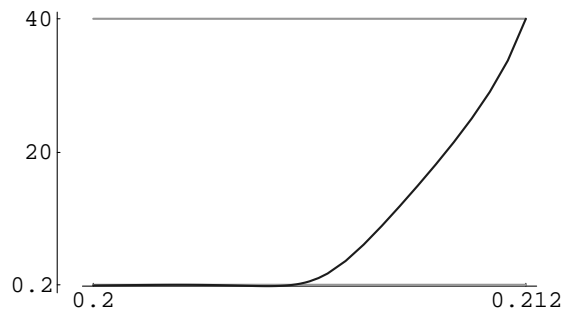
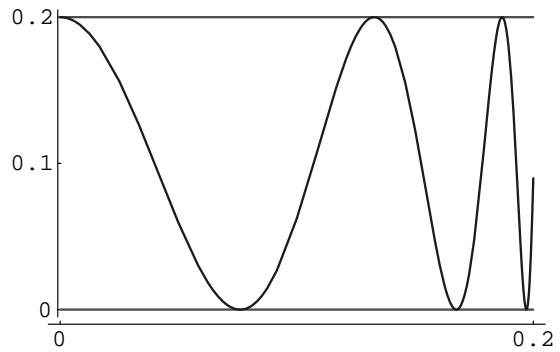


■ A.9.13 Design D3b

```

hz8D3b = Hz[8, amin8, emax, fpmax8, z] //N;
AdB8D3b = AdB[hz8D3b,z,f];
{b3bn8,a3bn8} = ba[8, amin8, emax, fpmax8, z] //N;
q3bn8 = Qfactor[a3bn8,b3bn8];
plotPTS[AdB8D3b,f,{0.2,0.212,0.2,40}]

```

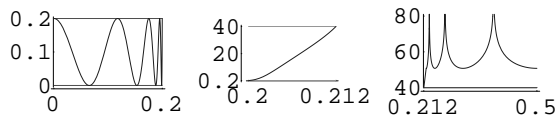
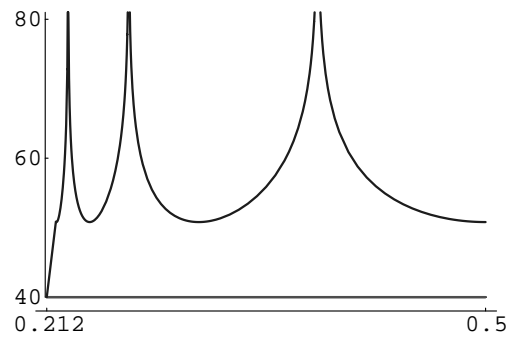
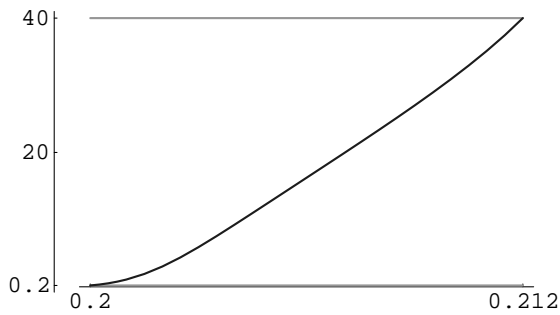
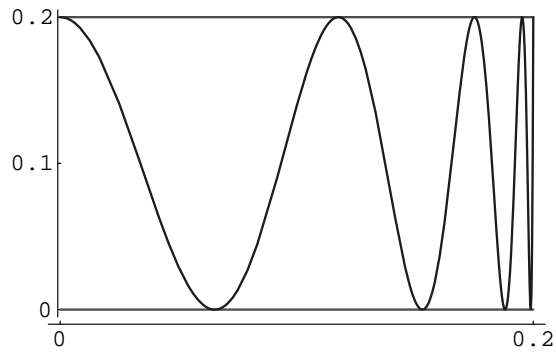


■ A.9.14 Design D4a

```

hz8D4a = Hz[8, amax8, emax, Fp, z] //N;
AdB8D4a = AdB[hz8D4a, z, f];
{b4an8, a4an8} = ba[8, amax8, emax, Fp, z] //N;
q4an8 = Qfactor[a4an8, b4an8];
plotPTS[AdB8D4a, f, {0.2, 0.212, 0.2, 40}]

```

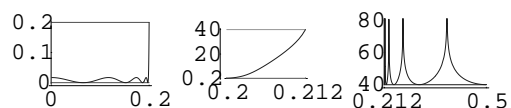
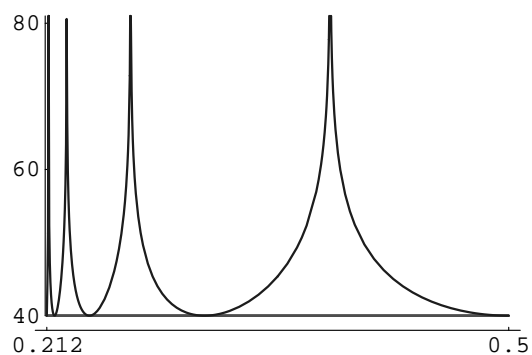
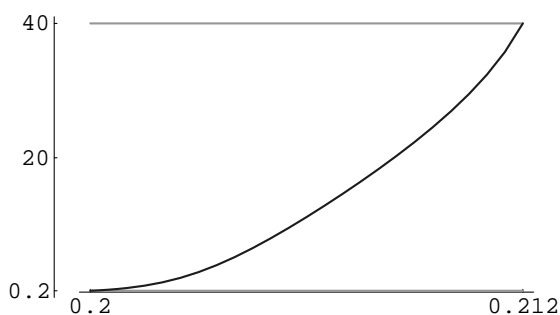
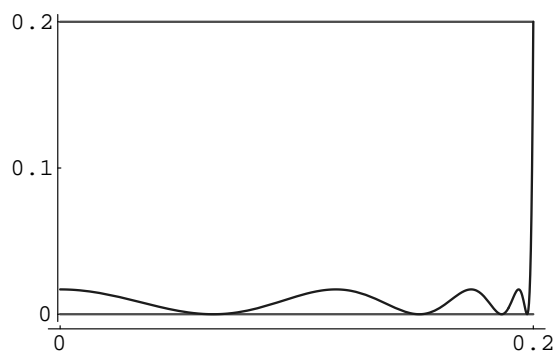


■ A.9.15 Design D4b

```

hz8D4b = Hz[8, amax8, emin8, fpmi8, z] //N;
AdB8D4b = AdB[hz8D4b,z,f];
{b4bn8,a4bn8} = ba[8, amax8, emin8, fpmi8, z] //N;
q4bn8 = Qfactor[a4bn8,b4bn8];
plotPTS[AdB8D4b,f,{0.2,0.212,0.2,40}]

```



■ A.9.16 Coefficients a and b, Q-Factor, and Sensitivity of the Second-Order Section with the Maximal Q-Factor

```
alist = {a1n8, a2n8, a3an8, a3bn8, a4an8, a4bn8};
blist = {b1n8, b2n8, b3an8, b3bn8, b4an8, b4bn8};
qlist = {q1n8, q2n8, q3an8, q3bn8, q4an8, q4bn8};
slist = 1/(1-alist)
```

```
{30.4061, 23.9766, 44.7132, 44.2529, 28.3548, 21.5643}
```

```
TableForm[Transpose[{blist,alist,qlist,slist,slist^3}]
, TableHeadings->{{"D1","D2","D3a","D3b","D4a","D4b"}
, {"b","a","Q-factor","1/(1-a)","1/(1-a)^3"}}]
```

	b	a	Q-factor	$1/(1-a)$	$1/(1-a)^3$
D1	-0.593746	0.967112	28.5112	30.4061	28111.3
D2	-0.565538	0.958293	22.4763	23.9766	13783.6
D3a	-0.600922	0.977635	42.1227	44.7132	89393.7
D3b	-0.534747	0.977403	42.1227	44.2529	86661.3
D4a	-0.592153	0.964733	26.5596	28.3548	22797.2
D4b	-0.572418	0.953627	20.1398	21.5643	10027.8

■ Remark

The expressions $1/(1-a)$ and $1/(1-a)^3$ are used to estimate the magnitude response sensitivity to the transfer function coefficients.

A.10 Jacobi Elliptic Functions

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■ A.10.1 References

1. M.D.Lutovac, D.V. Tasic, I.M.Markoski,
"Symbolic computation of elliptic rational functions,"
5th Int. Workshop, Symbolic Methods, Applications, Circuit Design,
SMACD'98, Kaiserslautern, Germany,
Oct. 1998, pp.177–180.
2. M. Abramowitz and I. Stegun,
"Handbook of Mathematical Functions,"
Dover, New York, 1972.
3. M. D. Lutovac, D. V. Tasic and B. L. Evans,
"Advanced Filter Design for Signal Processing using MATLAB and Mathematica,"
<http://iritel.iritel.bg.ac.yu/~lutovac/www/afdhome.htm>
<http://galeb.etf.bg.ac.yu/~tasic/afdhome.htm>
<http://www.ece.utexas.edu/~bevans/>

■ A.10.2 Initialization

```
SetDirectory[HomeDirectory[]];
<<afd\math\m\cleara11.m
```

■ A.10.3 Notation

cd[u,k] – Jacobi sine shifted
 cn[u,k] – Jacobi cosine
 k – modulus
 K[k] – complete elliptic integral of the first kind
 Kp[k] – complementary complete elliptic integral of the first kind
 sn[u,k] – Jacobi sine

■ A.10.4 Definitions

```
sn[u_,k_] := JacobiSN[u,k^2];
cn[u_,k_] := JacobiCN[u,k^2];
cd[u_,k_] := JacobiCD[u,k^2];
dn[u_,k_] := JacobiDN[u,k^2];
invsn[v_,k_] := InverseJacobiSN[v,k^2];
K[k_] := EllipticK[k^2];
Kp[k_] := K[Sqrt[1-k^2]];
```

■ A.10.5 Basic properties

```
{sn[0,k], sn[K[k],k], cn[0,k], cn[K[k],k]}
{0, 1, 1, 0}
```

```

a = 1.1;
b = 0.9;
{sn[I*a,b],
 I*sn[a,Sqrt[1-b^2]]/cn[a,Sqrt[1-b^2]]}
{cn[I*a,b],
 1/cn[a,Sqrt[1-b^2]]}
{dn[I*a,b],
 dn[a,Sqrt[1-b^2]]/cn[a,Sqrt[1-b^2]]}

{1.81411 I, 1.81411 I}
{2.07147, 2.07147}
{1.9146, 1.9146}

c = sn[a,b]
d = invsn[c,b]
a - d

0.819407
1.1
-16
6.66134 10

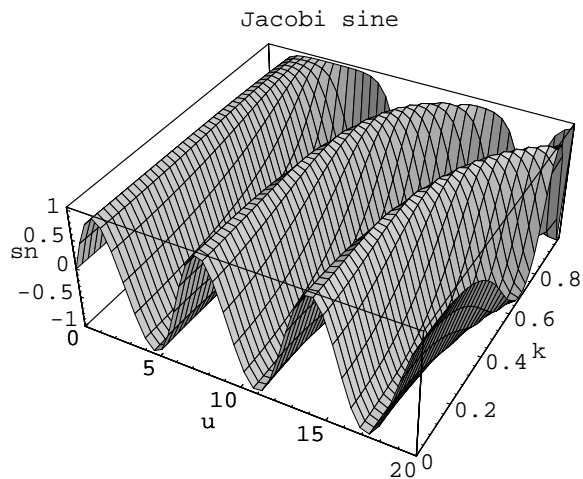
```

■ A.10.6 Plots

```

Plot3D[sn[u,k]
, {u,0,20},{k,0,0.999}
, AxesLabel -> {"u", "k", "sn"}
, PlotLabel -> "Jacobi sine"
, PlotPoints -> 40
];

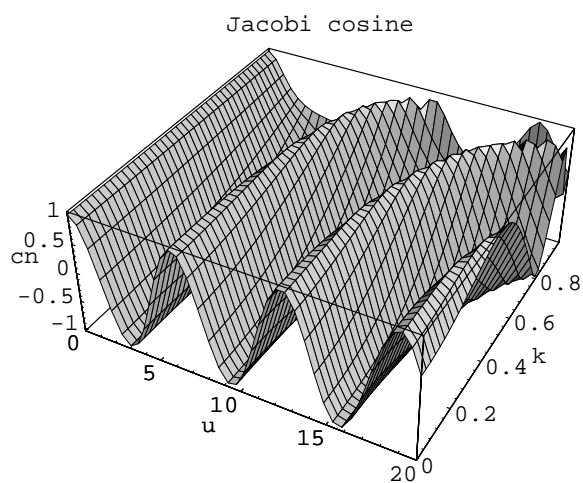
```



```

Plot3D[cn[u,k]
, {u,0,20},{k,0,0.999}
, AxesLabel -> {"u", "k", "cn"}
, PlotLabel -> "Jacobi cosine"
, PlotPoints -> 40
];

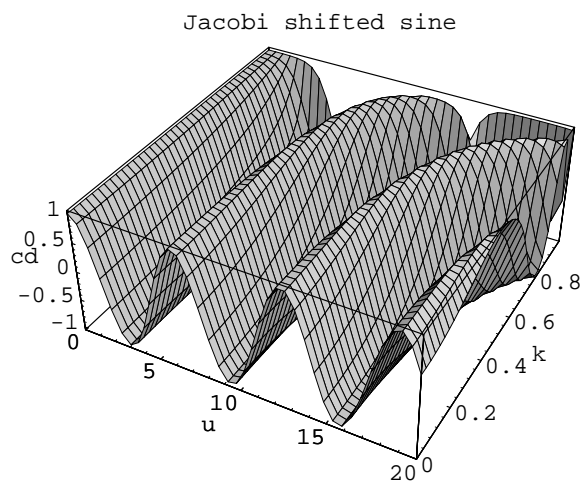
```



```

Plot3D[cd[u,k]
, {u,0,20},{k,0,0.999}
, AxesLabel -> {"u", "k", "cd"}
, PlotLabel -> "Jacobi shifted sine"
, PlotPoints -> 40
];

```



A.11 Elliptic Rational Function

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■ A.11.1 References

1. M. D. Lutovac, D.M.Rabrenovic,
"A simplified design of some Cauer filters without jacobian elliptic functions,"
IEEE Trans. Circ. Systems: Part II, vol. 39, no. 9, pp. 666–671, Sept. 1992.
2. M. D. Lutovac, D.M.Rabrenovic,
"Algebraic design of some lower-order elliptic filters,"
Electronics Letters, vol. 29, no. 2, pp. 192–193, Jan. 1993.
3. D.M.Rabrenovic, M. D. Lutovac,
"Minimum stopband attenuation of the Cauer filters without elliptic functions and integrals,"
IEEE Trans. Circ. Systems: Part I, vol. 40, no. 9, pp. 618–621, Sept. 1993.
4. M. D. Lutovac, D.M.Rabrenovic,
"Exact determination of the natural modes of some Cauer filters by means of a standard analytical procedure,"
IEE Proc. Circuits Devices Syst., vol. 143, no. 3, pp. 134–138, June 1996.
5. M. D. Lutovac, D. V. Tosic and B. L. Evans,
"Advanced Filter Design for Signal Processing using MATLAB and Mathematica,"
<http://iritel.iritel.bg.ac.yu/~lutovac/www/afdhome.htm>
<http://galeb.etf.bg.ac.yu/~tosic/afdhome.htm>
<http://www.ece.utexas.edu/~bevans/>

■ A.11.2 Initialization

```
SetDirectory[HomeDirectory[]];
<<afd\math\m\clearall.m
<<afd\math\m\jelldf.m
<<graphics\graphics'
```

Discrimination Factor $L(n, 1/k)$
 $n = 1, 2, 3, 4, 6, 8, 9, 12, 16, 18$

■ A.11.3 Notation

cd[u,k] – Jacobi sine shifted
 invcd[v,k] – inverse of cd
 k – modulus
 K[k] – complete elliptic integral of the first kind
 Kp[k] – complementary complete elliptic integral of the first kind
 L[n,1/k] – discrimination factor
 n – order, $n > 0$
 R[n,k,p,x] – function from which we generate elliptic rational function

■ A.11.4 Definitions

```
cd[u_,k_] := JacobiCD[u,k^2];
invcd[v_,k_] := InverseJacobiCD[v,k^2];
K[k_] := EllipticK[k^2];
Kp[k_] := K[Sqrt[1-k^2]];
R[n_,k_,p_,x_] := cd[n*K[p]*invcd[x,k]/K[k],p]
x[k_,u_,v_] := cd[u+I*v,k];
Ruv[n_Integer,k_,u_,v_] := cd[n*(u+I*v),1/L[n,1/k]];
```

■ A.11.5 Example

```
{R[2,0.8,0.1,-1.0], R[2,0.8,0.1,0], R[2,0.8,0.1,1.0]}
```

```
{1., -1., 1.}
```

■ Remark

We know that $\text{invcd}(0,k)=K(k)$, so, we add a rule

```
{invcd[0,0.9], K[0.9]}
invcd[0.0,0.9]
```

```
{2.28055, 2.28055}
```

1

Power::infy: Infinite expression -- encountered.

0.

Infinity::indet: Indeterminate expression 0. ComplexInfinity encountered.

```
2.28054913842277 - 2 EllipticLog[{Indeterminate, ComplexInfinity}, {3.62, 0.0361}] -
```

```
2 (1.654616667522527 I Round[0.302185
```

```
Im[0. - 2 EllipticLog[{Indeterminate, ComplexInfinity}, {3.62, 0.0361}]] +
```

```
2.28054913842277 Round[0.219245 Re[0. -
```

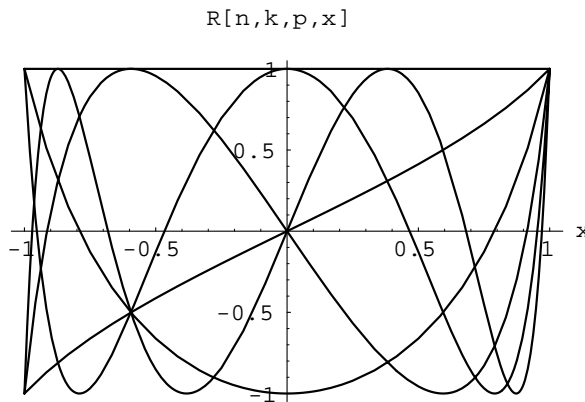
```
2 EllipticLog[{Indeterminate, ComplexInfinity}, {3.62, 0.0361}]]])
```

```
invcd[0.0,k_] := K[k];
{invcd[0,0.9], invcd[0.0,0.9]}
```

```
{2.28054913842277, 2.28055}
```

■ A.11.6 Plots

```
Plot[{R[0, 0.8, 0.1, x],
      R[1, 0.8, 0.1, x],
      R[2, 0.8, 0.1, x],
      R[3, 0.8, 0.1, x],
      R[4, 0.8, 0.1, x],
      R[5, 0.8, 0.1, x]}
, {x,-1,1}
, AxesLabel -> {"x", "R[n,k,p,x]"}];
```



■ **A.11.7 Can we Construct a Rational Function from $R(n, k, p, x)$ by Varying k and p ?**

For $n=1$

$$R(1, k, p, x) = \text{cd}(K(p) \cdot \text{invcd}(x, k) / K(k), p)$$

and simplifies to $R(1, k, p, x) = x$ for $k=p$.

$$R(1, k, p, 0) = 0$$

$$R(1, k, p, 1) = 1$$

$$R(1, k, p, 1/k) = 1/k = 1/p$$

■ **A.11.8 Find p Which Satisfies $R(n, k, p, 1/k) = 1/p$ for $n=1$**

$n1 = 1$

$k1 = N[95/100, 24]$

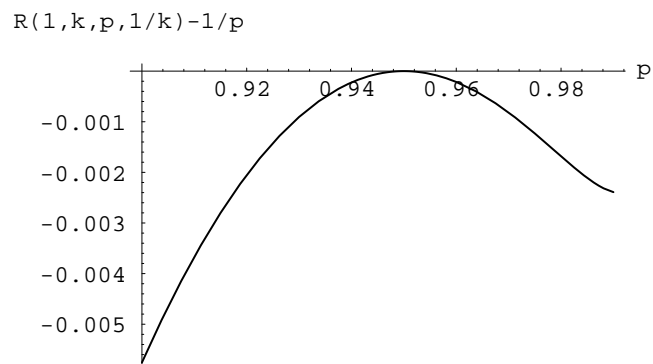
$\text{Plot}[R[n1, k1, p, 1/k1] - 1/p$

, { $p, 0.9, 0.99$ }

, AxesLabel -> {" p ", " $R(1, k, p, 1/k) - 1/p$ "}];

1

0.95



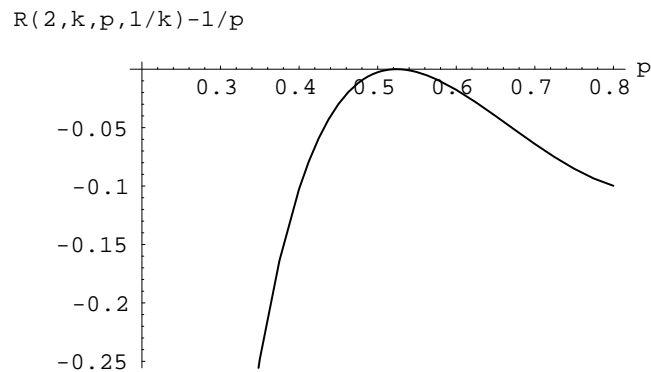
```
sol = FindRoot[
  R[n1,k1,p,1/k1] == 1/p
, {p,0.949,0.951}
, AccuracyGoal -> 16
, MaxIterations -> 64
];
p1 = N[p/.sol,16]
plexact = k1
```

```
0.950000006790337 + 5.259681766043407 10-17 I
0.95
```

■ A.11.9 Find p Which Satisfies $R(n,k,p,1/k)=1/p$ for $n=2$

```
n2 = 2
k2 = N[95/100,24]
Plot[R[n2,k2,p,1/k2] - 1/p
, {p,0.2,0.8}
, AxesLabel -> {"p", "R(2,k,p,1/k)-1/p"}];

2
0.95
```



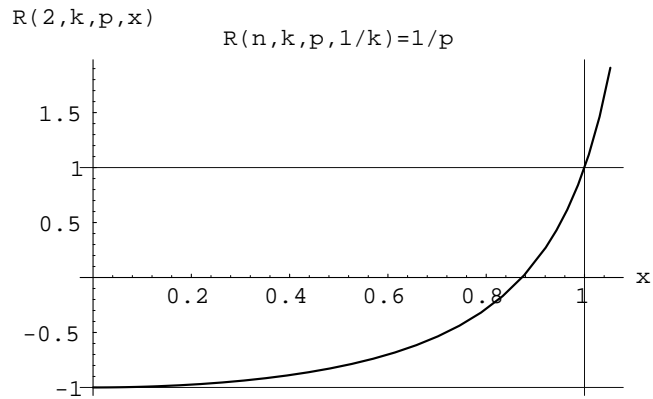
```
sol = FindRoot[
  R[n2,k2,p,1/k2] == 1/p
, {p,0.52,0.53}
, AccuracyGoal -> 16
, MaxIterations -> 64
];
p2 = N[p/.sol,16]
p2exact = 1/(1/k2+Sqrt[1/k2^2-1])^2

0.5240999634235638
0.52409994477580075281475
```

```

Plot[R[n2,k2,p2,x]
, {x,0,1/k2}
, AxesLabel -> {"x", "R(2,k,p,x)"}
, PlotLabel -> "R(n,k,p,1/k)=1/p"
, GridLines -> {{1},{-1,1}}
];

```



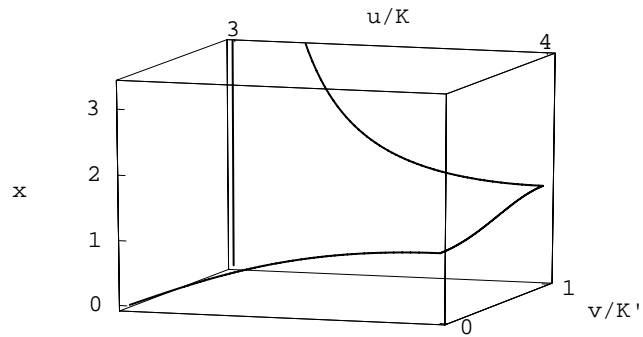
From now on we assume $R(n,k,p,1/k)=1/p$
and calculate $p = 1/L(n,1/k)$
(L can be exactly computed for $n=1,2,3,4,6,8,9,12,16,18$)

■ **A.11.10 Parametric Plot of x in Terms of Complex Parametric Variable $w=u+I*v$**

```

n1 = 3;
k1 = 0.7;
K1 = K[k1];
ParametricPlot3D[
  {{t, 0, x[k1, t*K1, 0]},
   {4, t-3, x[k1, 4*K1, (t-3)*K1]}},
  {t, 1, Re[x[k1, t*K1, K1]]}
, {t,3,4}
, Ticks -> {{3,4},{0,1},{0,1,2,3,4}}
, AspectRatio -> 0.6
, ViewPoint -> {10,-30,20}
, AxesLabel -> {"u/K", "v/K'", "x"}
];

```

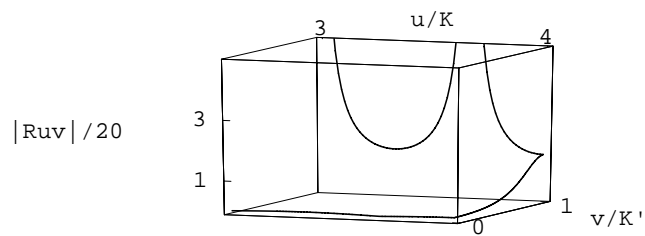



■ A.11.11 Parametric Plot of R in Terms of Complex Parametric Variable $w=u+i*v$

```

n1 = 3;
k1 = 0.7;
Kn1 = K[1/L[n1,1/k1]];
r0 = 1/20;
ParametricPlot3D[
  {{t, 0, r0*Abs[Ruv[n1, k1, t*Kn1, 0]]},
   {4, t-3, r0*Abs[Ruv[n1, k1, 4*Kn1, (t-3)*Kn1]]},
   {t, 1, r0*Abs[Ruv[n1, k1, t*Kn1, Kn1]]}
  }, {t,3,4}
, Ticks -> {{3,4},{0,1},{1,3,5}}
, AspectRatio -> 0.5
, ViewPoint -> {10,-25,20}
, AxesLabel -> {"u/K", "v/K'", "|Ruv|/20"}
];

```

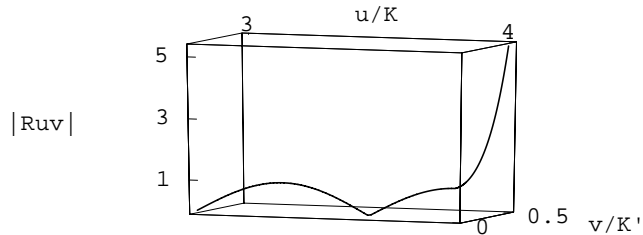


```

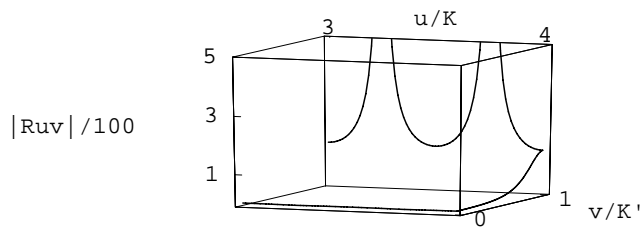
n1 = 3;
k1 = 0.7;
Kn1 = K[1/L[n1,1/k1]];
ParametricPlot3D[
  {{t, 0, Abs[Ruv[n1, k1, t*Kn1, 0]]},
   {4, (t-3)/2, Abs[Ruv[n1, k1, 4*Kn1, (t-3)*Kn1/2]]}
  }, {t,3,4}
, Ticks -> {{3,4},{0,0.5},{1,3,5}}
, AspectRatio -> 0.5
];

```

```
, ViewPoint -> {10,-25,20}
, AxesLabel -> {"u/K", "v/K'", "|Ruv|"}
];
```

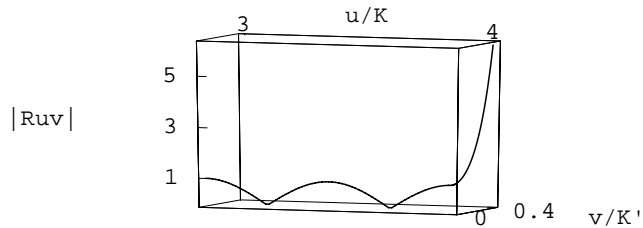


```
n1 = 4;
k1 = 0.7;
Kn1 = K[1/L[n1,1/k1]];
r0 = 1/100;
ParametricPlot3D[
  {{t, 0, r0*Abs[Ruv[n1, k1, t*Kn1, 0]]},
   {4, t-3, r0*Abs[Ruv[n1, k1, 4*Kn1, (t-3)*Kn1]]},
   {t, 1, r0*Abs[Ruv[n1, k1, t*Kn1, Kn1]]}
  ], {t, 3, 4}
, Ticks -> {{3,4},{0,1},{1,3,5}}
, AspectRatio -> 0.5
, ViewPoint -> {10,-25,20}
, AxesLabel -> {"u/K", "v/K'", "|Ruv|/100"}
];
```



```
n1 = 4;
k1 = 0.7;
Kn1 = K[1/L[n1,1/k1]];
ParametricPlot3D[
  {{t, 0, Abs[Ruv[n1, k1, t*Kn1, 0]]},
   {4, (t-3)/2.5, Abs[Ruv[n1, k1, 4*Kn1, (t-3)*Kn1/2.5]]}
  ], {t, 3, 4}
, Ticks -> {{3,4},{0,0.4},{1,3,5}}
, AspectRatio -> 0.5
```

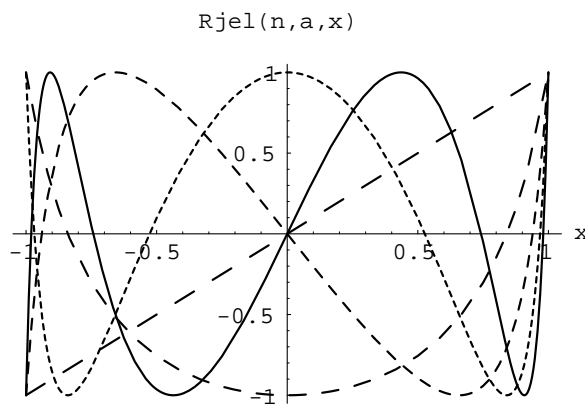
```
, ViewPoint -> {10,-25,20}
, AxesLabel -> {"u/K", "v/K'", "|Ruv|"}
];
```



■ A.11.12 Elliptic Rational Function in Terms of Jacobi Elliptic Functions: $a=1/k$

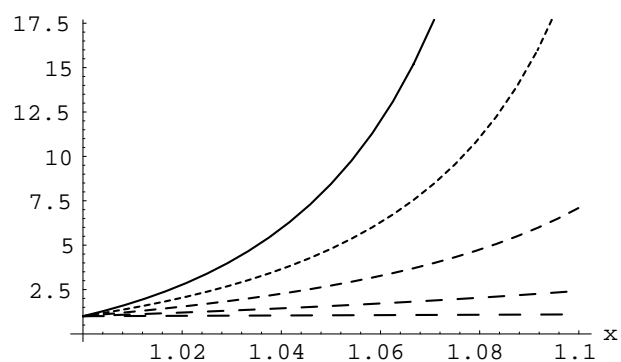
```
Rjel[0, a_, x_] := 1
Rjel[1, a_, x_] := x
Rjel[2, a_, x_] := ((1+Sqrt[1-1/a^2])*x^2-1)/
  ((-1+Sqrt[1-1/a^2])*x^2+1) ;
Rjel[3,a_,x_] := Block[
{k=1/a,z1,z2,b1,b2,b3},
  z1 = JacobiDN[2*EllipticK[k^2]/3,k^2];
  b3 = (1 + z1)^2;
  b1 = -1 - 2*z1;
  b2 = -1 + z1^2;
  (b3*x^3 + b1*x)/(b2*x^2 + 1)
];
Rjel[4, a_, x_] := Block[{t},
  t = Sqrt[1 - 1/a^2];
  ((1+t)*(1+Sqrt[t])^2*x^4-2(1+t)*(1+Sqrt[t])*x^2+1)/
  ((1+t)*(1-Sqrt[t])^2*x^4-2(1+t)*(1-Sqrt[t])*x^2+1)];
Rjel[5,a_,x_] := Block[
{k=1/a,z1,z2,b1,b2,b3,b4,b5},
  z1 = JacobiDN[2*EllipticK[k^2]/5,k^2];
  z2 = JacobiDN[4*EllipticK[k^2]/5,k^2];
  b5 = (1+z1)^2*(1+z2)^2;
  b3 = -(1-z1^2*z2^2+(1+z1)^2*(1+z2)^2+2*(z1+z2));
  b1 = (1+2*(z1+z2));
  b4 = (1-z1^2)*(1-z2^2);
  b2 = -(2-z1^2-z2^2);
  (b5*x^5 + b3*x^3 + b1*x)/(b4*x^4 + b2*x^2 + 1)
];
a1 = 1.1;
Plot[{Rjel[1, a1, x],
  Rjel[2, a1, x],
  Rjel[3, a1, x],
  Rjel[4, a1, x],
  Rjel[5, a1, x]}
```

```
, {x, -1, 1}
, AxesLabel -> {"x", "Rjel(n,a,x)"}
, PlotStyle -> {Dashing[{0.04}],
                Dashing[{0.03}],
                Dashing[{0.02}],
                Dashing[{0.01}],
                Dashing[{}}]}
];
```



```
a1 = 1.1;
Plot[{Rjel[1, a1, x],
      Rjel[2, a1, x],
      Rjel[3, a1, x],
      Rjel[4, a1, x],
      Rjel[5, a1, x]}
, {x, -1, 1}
, AxesLabel -> {"x", "Rjel(n,a,x)"}
, PlotStyle -> {Dashing[{0.04}],
                Dashing[{0.03}],
                Dashing[{0.02}],
                Dashing[{0.01}],
                Dashing[{}}]}
];
```

Rjel(n,a,x)

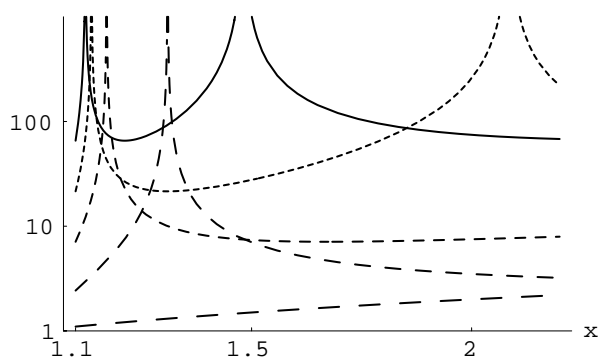


```

a1 = 1.1;
LogPlot[{Abs[Rjel[1, a1, x]],
          Abs[Rjel[2, a1, x]],
          Abs[Rjel[3, a1, x]],
          Abs[Rjel[4, a1, x]],
          Abs[Rjel[5, a1, x]]},
  {x, a1, 2*a1},
  AxesLabel -> {"x", "Rjel(n,a,x)"},
  PlotRange -> {1, 1000},
  Ticks -> {{a1, 1.5, 2}, {1, 10, 100}},
  PlotStyle -> {Dashing[{0.04}],
                 Dashing[{0.03}],
                 Dashing[{0.02}],
                 Dashing[{0.01}],
                 Dashing[{}}]
];

```

Rjel(n,a,x)

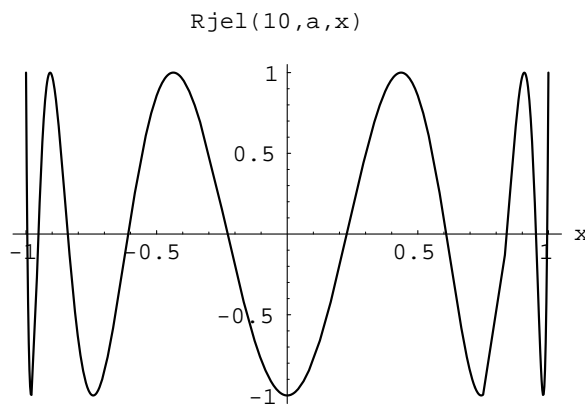


■ A.11.13 Nesting Property

```

Rjel[10, a_, x_] := Rjel[2, Rjel[5, a, a], Rjel[5, a, x]];
a1 = 1.1;
{Rjel[10, a1, -1.0], Rjel[10, a1, 0.0], Rjel[10, a1, 1.0]}
{1., -1., 1.}
a1 = 1.1;
Plot[Evaluate[Rjel[10, a1, x]]
, {x, -1, 1}
, AxesLabel -> {"x", "Rjel(10, a, x)"}
];

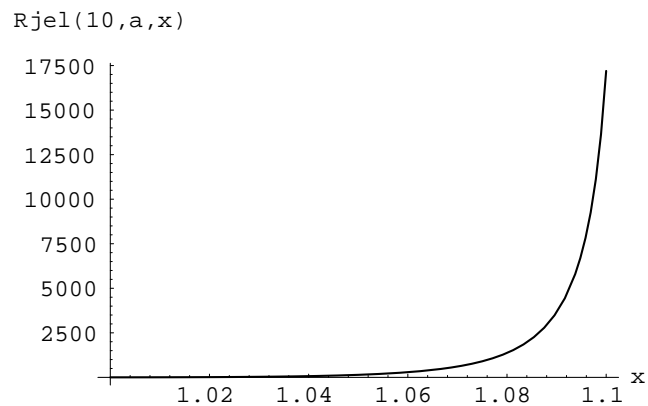
```



```

a1 = 1.1;
Plot[Evaluate[Rjel[10, a1, x]]
, {x, 1, a1}
, AxesLabel -> {"x", "Rjel(10, a, x)"}
];

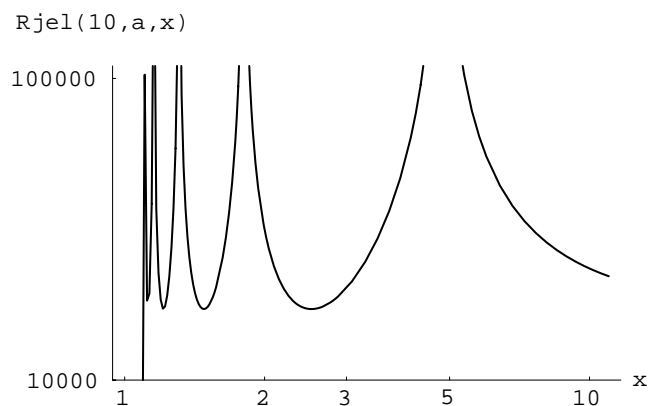
```



```

a1 = 1.1;
LogLogPlot[Evaluate[Abs[Rjel[10, a1, x]]], {x, 1, 10*a1},
, AxesLabel -> {"x", "Rjel(10,a,x)"}
, PlotRange -> {10^4, 1.1*10^5}
, Ticks -> {{1, 2, 3, 5, 10}, {10^4, 10^5}}
];

```



Rjel[10,a1,x] //Together

$$\begin{array}{r}
 211.208 - 5361.41 x^2 + 28064.4 x^4 - 57774.2 x^6 + 52056. x^8 - 17198.7 x^{10} \\
 \hline
 -211.208 + 528.325 x^2 - 484.596 x^4 + 194.543 x^6 - 30.7152 x^8 + 1. x^{10}
 \end{array}$$

A.12 Transfer Function with Minimal Q-Factors

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■ A.12.1 References

1. D.M.Rabrenovic, M. D. Lutovac,
"Elliptic filters with minimal Q-factors,"
Electronics Letters, vol. 30, no. 3, pp. 206–207, Feb. 1994.
2. M. D. Lutovac, D. V. Tasic and B. L. Evans,
"Advanced Filter Design for Signal Processing using MATLAB and Mathematica,"
<http://iritel.iritel.bg.ac.yu/~lutovac/www/afdhome.htm>
<http://galeb.etf.bg.ac.yu/~tasic/afdhome.htm>
<http://www.ece.utexas.edu/~bevans/>

■ A.12.2 Initialization

```
SetDirectory[HomeDirectory[]];
<<afd\math\m\clearall.m
<<afd\math\m\erfdf.m
<<afd\math\m\erfez.m
<<afd\math\m\erfzh.m
<<afd\math\m\erf235.m
<<graphics\graphics'
```

```
Discrimination Factor L(n,a)
n = 1,2,3,4,6,8,9,12,16,18
Exact Formulas for Zeros: n=2^i 3^j
X(n,a)={list-of-zeros}, X(n,a,i)
n = 1,2,3,4,6,8,9,12,16,18
Zeta Function Z(n,a,e)
Elliptic rational functions: n=2^i 3^j
Reliptic(n,a,x)
n = 1,2,3,4,5,6,8,9,10,12,15,16,18
```

■ A.12.3 Notation

a – selectivity factor
e – ripple factor
n – order
L(n,a) – discrimination factor
H - normalized lowpass transfer function
HminQ - minQ normalized lowpass transfer function
Q(s) – Q-factor of a pole s
QminQ(n,a) – list of pole Q-factors of minQ transfer function
QminQ(n,a,i) – pole Q-factor of minQ transfer function
Reliptic(n,a,x) – elliptic rational function
s - normalized complex frequency
S(n,a,e) – list of transfer function poles

$S(n,a,e,i)$ – transfer function pole
 $SminQ(n,a,e)$ – list of minQ transfer function poles
 $SminQ(n,a,e,i)$ – minQ transfer function pole
 $X(n,a)$ – list of zeros of elliptic rational function
 $X(n,a,i)$ – zero of elliptic rational function
 w – normalized angular frequency
 $Z(n,a,e)$ – zeta function

■ A.12.4 Definition and Procedures

```

Q[s_] := - Abs[s]/(2*Re[s]);
QminQ[n_Integer, a_, i_Integer] :=
  (a + X[n,a,i]^2)/
  (2 Sqrt[1- X[n,a,i]^2]*Sqrt[a^2 - X[n,a,i]^2]);
QminQ[n_Integer, a_] :=
  (a + X[n,a]^2)/
  (2 Sqrt[1- X[n,a]^2]*Sqrt[a^2 - X[n,a]^2]);
SminQ[n_Integer, a_, i_Integer] := Block[
{den,num,numim,numre,x,z},
  numre = -Sqrt[1- X[n,a,i]^2]*Sqrt[a^2- X[n,a,i]^2];
  numim = X[n,a,i]*(a + 1);
  num = numre + I*numim;
  den = a + X[n,a,i]^2;
  Sqrt[a]*num/den
];
SminQ[n_Integer, a_] := Block[
{den,num,numim,numre,x,z},
  numre = -Sqrt[1- X[n,a]^2]*Sqrt[a^2- X[n,a]^2];
  numim = X[n,a]*(a + 1);
  num = numre + I*numim;
  den = a + X[n,a]^2;
  Sqrt[a]*num/den
];
S[n_Integer, a_, e_, i_Integer] := Block[
{den,num,numim,numre,x,z},
  x = X[n,a,i];
  z = Z[n,a,e];
  numre = -z*Sqrt[1 - z^2]*Sqrt[1 - x^2]*Sqrt[1 - x^2/a^2];
  numim = x*Sqrt[1 - (1-1/a^2)*z^2];
  num = numre + I*numim;
  den = 1 - (1 - x^2/a^2)*z^2;
  num/den
];
S[n_Integer, a_, e_] := S[n,a,e,#]& /@ Range[n]

```

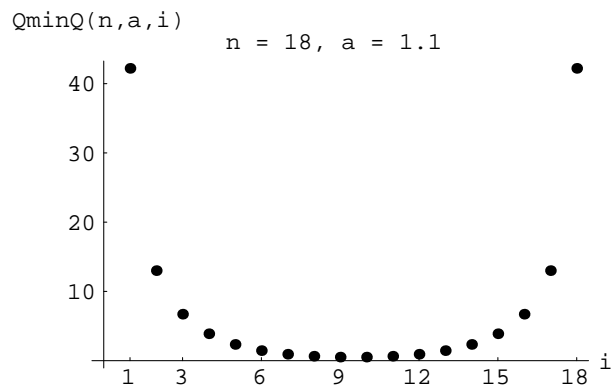
■ A.12.5 Example

```

n1 = 18;
a1 = 1.1;
e1 = 0.1;
i1 = 1;
S[n1,a1,e1,i1]
SminQ[n1,a1,i1]
Q[S[n1,a1,e1,i1]]
Q[SminQ[n1,a1,i1]]
QminQ[n1,a1,i1]
QminQ[n1,a1]
ListPlot[%
,PlotStyle->{AbsolutePointSize[4]}
,AxesLabel -> {"i", "QminQ(n,a,i)"}
,Ticks->{{1,3,6,9,12,15,18},{10,20,30,40}}
,PlotLabel -> "n = 18, a = 1.1"
];

-0.0053858 - 1.00386 I
-0.0124234 - 1.04874 I
93.1967
42.2111
42.2111
{42.2111, 13.0034, 6.71209, 3.88702, 2.34521, 1.45084, 0.930164, 0.642843, 0.515239,
 0.515239, 0.642843, 0.930164, 1.45084, 2.34521, 3.88702, 6.71209, 13.0034, 42.2111}

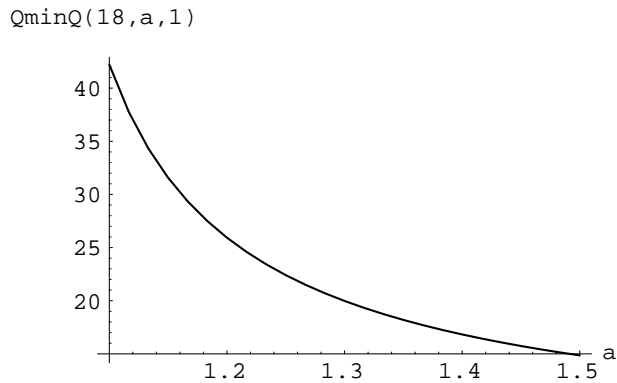
```



```

Plot[QminQ[18,a,1]
,{a,1.1,1.5}
,AxesLabel -> {"a", "QminQ(18,a,1)"}
];

```

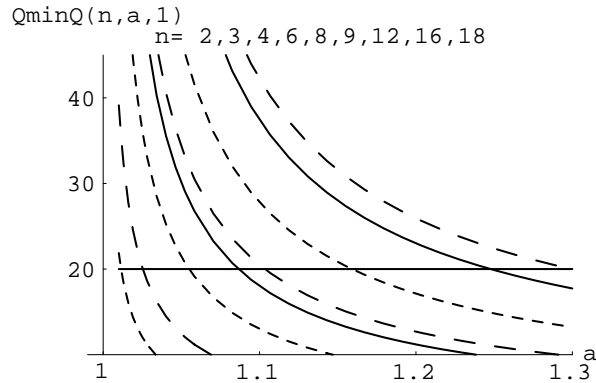


■ A.12.6 Find Selectivity Factor for Given Maximal Q-Factor

```
FindRoot[QminQ[18,a,1]==20, {a,1.1,1.5}]
```

```
{a -> 1.29948}
```

```
Plot[{QminQ[18,a,1],
      QminQ[16,a,1],
      QminQ[12,a,1],
      QminQ[9, a,1],
      QminQ[8, a,1],
      QminQ[6, a,1],
      QminQ[4, a,1],
      20,
      QminQ[3,a,1],
      QminQ[2,a,1]},
  {a,1.01,1.3},
  ,PlotRange -> {{0.99,1.3},{10,45}}
  ,PlotStyle -> {Dashing[.04]},
               Dashing[{}],
               Dashing[.02]}
  ,AxesLabel -> {"a", "QminQ(n,a,1)"}
  ,Ticks->{{1.0,1.1,1.2,1.3},{10,20,30,40}}
  ,PlotLabel -> "n= 2,3,4,6,8,9,12,16,18"
  ];
```



```
nminQ = {18,16,12,9,8,6,4,3,2};
```

```
aminQ = a /.
```

```
{FindRoot[QminQ[18,a,1]==20, {a,1.1, 1.3}],
```

```
FindRoot[QminQ[16,a,1]==20, {a,1.1, 1.3}],
```

```
FindRoot[QminQ[12,a,1]==20, {a,1.1, 1.3}],
```

```
FindRoot[QminQ[9, a,1]==20, {a,1.1, 1.3}],
```

```
FindRoot[QminQ[8, a,1]==20, {a,1.05, 1.1}],
```

```
FindRoot[QminQ[6, a,1]==20, {a,1.04, 1.1}],
```

```
FindRoot[QminQ[4, a,1]==20, {a,1.03, 1.1}],
```

```
FindRoot[QminQ[3, a,1]==20, {a,1.02, 1.1}],
```

```
FindRoot[QminQ[2, a,1]==20, {a,1.001,1.1}]
```

```
};
```

```
t = Transpose[{nminQ,aminQ}]
```

```
ListPlot[t
```

```
, PlotJoined -> True
```

```
, PlotRange -> {1,1.3}
```

```
, AxesLabel -> {"n", "a"}
```

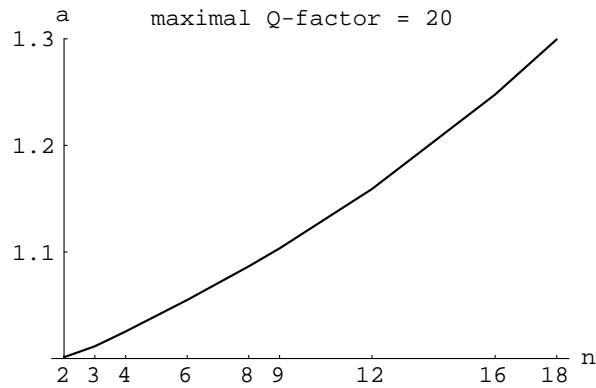
```
, Ticks->{{2,3,4,6,8,9,12,16,18},{1,1.1,1.2,1.3}}
```

```
, PlotLabel -> "maximal Q-factor = 20"
```

```
];
```

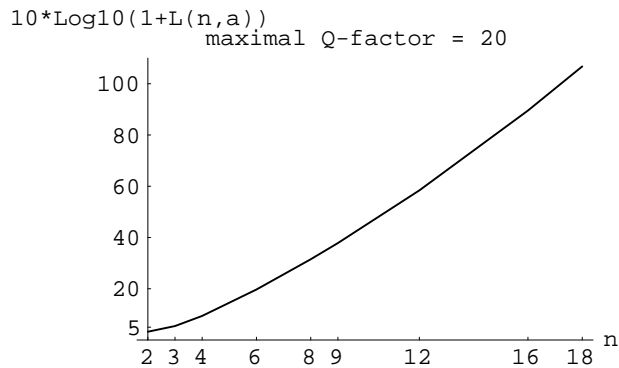
```
{{18, 1.29948}, {16, 1.24774}, {12, 1.15906}, {9, 1.10338}, {8, 1.08654}, {6, 1.05495},
```

```
{4, 1.02538}, {3, 1.01146}, {2, 1.00125}}
```



```
tL = Table[{nminQ[[i]],
            10*Log[10,1+L[nminQ[[i]],aminQ[[i]]]}],
           {i,1,9}
];
ListPlot[tL
, PlotJoined -> True
, PlotRange -> {0,110}
, AxesLabel -> {"n", "10*Log10(1+L(n,a))"}
, Ticks->{{2,3,4,6,8,9,12,16,18},{0,5,20,40,60,80,100}}
, PlotLabel -> "maximal Q-factor = 20"
];
```

```
{{18, 106.763}, {16, 89.5348}, {12, 58.386}, {9, 37.8317}, {8, 31.5114}, {6, 19.7043},
{4, 9.39632}, {3, 5.4385}, {2, 3.23299}}
```



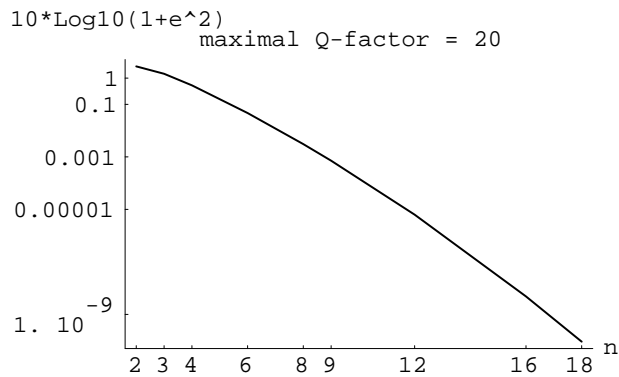
■ A.12.7 Ripple Factor $e_{\min Q} = 1 / \text{Sqrt}[L(n, a)]$

```
te = Table[{nminQ[[i]],
            10*Log[10,1+1/L[nminQ[[i]],aminQ[[i]]]}],
           {i,1,9}
];
LogListPlot[te
```

```
, PlotJoined -> True
, AxesLabel -> {"n", "10*Log10(1+e^2)"}
, Ticks->{{2,3,4,6,8,9,12,16,18},{1,.1,.001,.00001,.000000001}}
, PlotLabel -> "maximal Q-factor = 20"
];
```

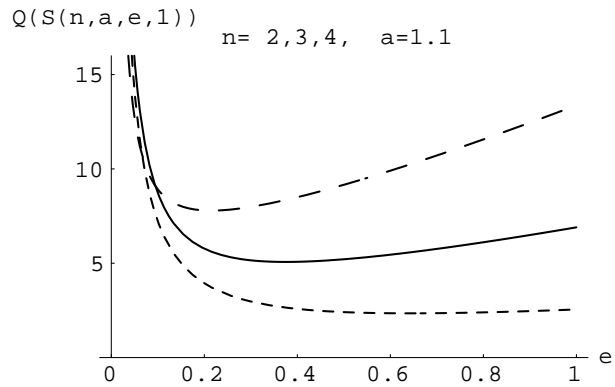
```

-11          -9          -6
{{18, 9.15089 10  }, {16, 4.83395 10  }, {12, 6.29779 10  }, {9, 0.00071557},
 {8, 0.0030676}, {6, 0.0467403}, {4, 0.530139}, {3, 1.46215}, {2, 2.79847}}
```



■ A.12.8 Pole Q-Factor in Terms of Ripple Factor

```
a1 = 1.1;
Plot[{Q[S[4,a1,e,1]],
      Q[S[3,a1,e,1]],
      Q[S[2,a1,e,1]]},
{e,0.01,1}
,PlotRange -> {0,16}
,PlotStyle -> {Dashing[.04]},
              Dashing[{}],
              Dashing[.02]}
,AxesLabel -> {"e", "Q(S(n,a,e,1))"}
,Ticks->{{0,0.2,0.4,0.6,0.8,1},{0,5,10,15}}
,PlotLabel -> "n= 2,3,4, a=1.1"
];
```



■ A.12.9 Minimal Q-Factor Transfer Function

```
HminQ[n_,a_,s_] := Block[{e,i,K,t},
  e = 1/Sqrt[L[n,a]];
  If[EvenQ[n],
    K = Product[X[n,a,i]^2,{i,1,n/2}]/(Sqrt[1+e^2]*Sqrt[a^n]);
    t = K*Product[(s^2+(a^2/X[n,a,i]^2))/(s^2+(Sqrt[a]/QminQ[n,a,i])*s+a),
      {i,1,n/2}];,
    K = Product[X[n,a,i]^2,{i,1,(n-1)/2}]/((s+Sqrt[a])*Sqrt[a^(n-2)]);
    t = K*Product[(s^2+(a^2/X[n,a,i]^2))/(s^2+(Sqrt[a]/QminQ[n,a,i])*s+a),
      {i,1,(n-1)/2}];
  ];
  t
];
a1 = 1.1;
H1 = HminQ[1,a1,s]
H2 = HminQ[2,a1,s]
H3 = HminQ[3,a1,s]
H4 = HminQ[4,a1,s]
H6 = HminQ[6,a1,s]
H8 = HminQ[8,a1,s];
H9 = HminQ[9,a1,s];

1.04881
-----
1.04881 + s
2
0.540094 (1.71408 + s )
-----
2
1.1 + 0.447214 s + s
```

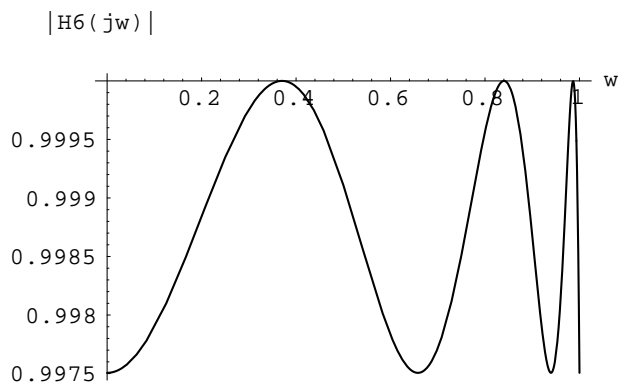
$$\frac{0.841917 (1.37031 + s)^2}{(1.04881 + s) (1.1 + 0.206892 s + s^2) \cdot 0.210643 (1.29093 + s)^2 (4.34993 + s^2)}$$

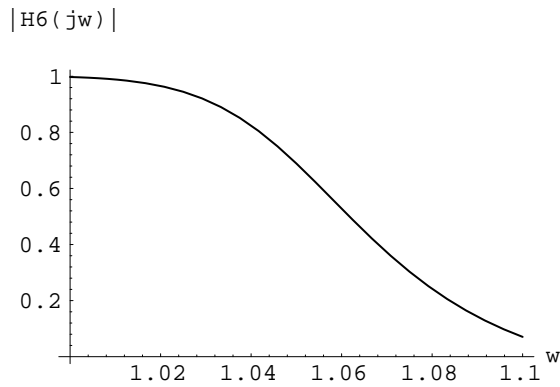
$$\frac{(1.1 + 0.134615 s + s^2) (1.1 + 1.24828 s + s^2) \cdot 0.0705794 (1.24336 + s)^2 (1.71408 + s)^2 (8.82646 + s^2)}{(1.1 + 0.0806565 s + s^2) (1.1 + 0.447214 s + s^2) (1.1 + 1.63152 s + s^2)}$$

```
H6w = H6 /. s->I*w;
```

■ A.12.10 Magnitude Response in Terms of Angular Frequency

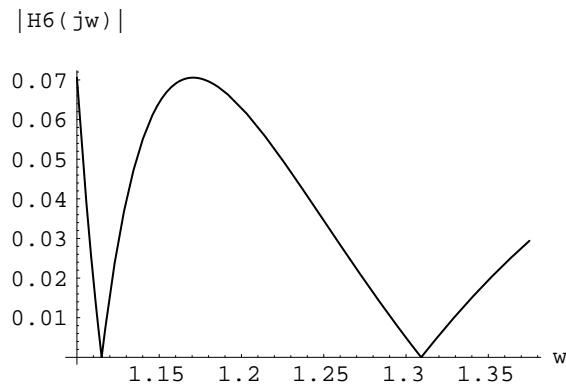
```
Plot[Abs[H6w]
,{w,0,1}
,AxesLabel -> {"w", "|H6(jw)|"}
];
Plot[Abs[H6w]
,{w,1,a1}
,AxesLabel -> {"w", "|H6(jw)|"}
];
```

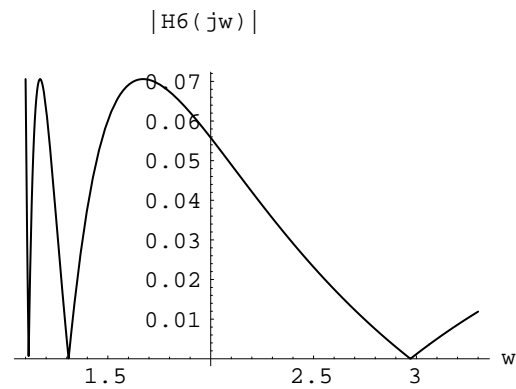




```
Plot[Abs[H6w]
,{w,a1,5*a1/4}
,AxesLabel -> {"w", "|H6(jw)|"}
];

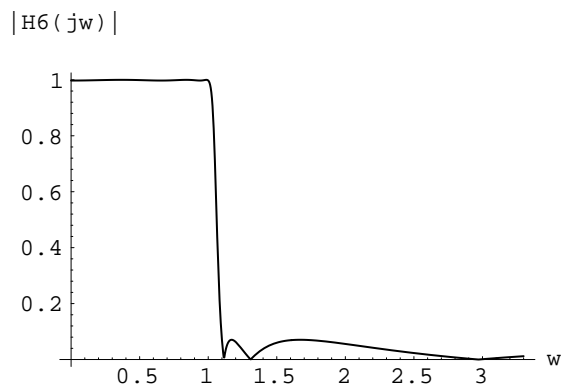
Plot[Abs[H6w]
,{w,a1,3*a1}
,AxesLabel -> {"w", "|H6(jw)|"}
,PlotPoints->75
];
```

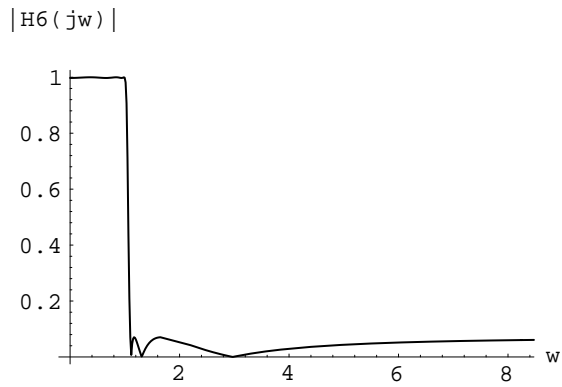




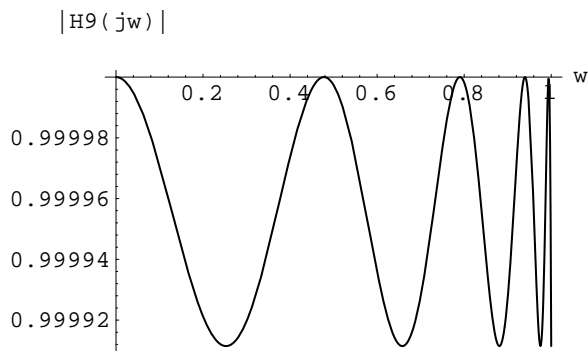
```
Plot[Abs[H6w]
,{w,0,3*a1}
,AxesLabel -> {"w", "|H6(jw)|"}
];
```

```
Plot[Abs[H6w]
,{w,0,12*a1}
,AxesLabel -> {"w", "|H6(jw)|"}
];
```

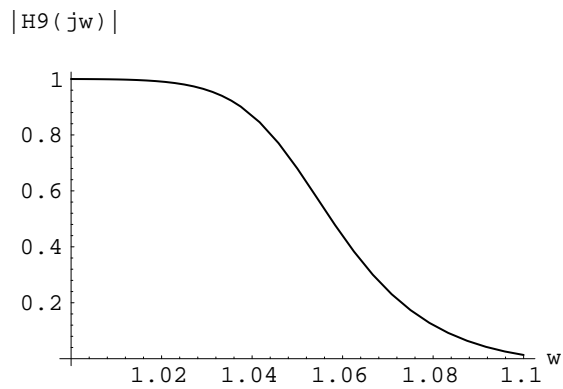




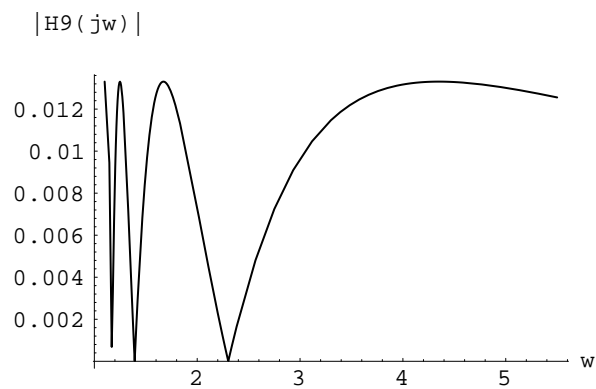
```
H9w = H9 /. s->I*w;
Plot[Abs[H9w]
,{w,0,1}
,AxesLabel -> {"w", "|H9(jw)|"}
];
```



```
Plot[Abs[H9w]
,{w,1,a1}
,AxesLabel -> {"w", "|H9(jw)|"}
];
```



```
Plot[Abs[H9w]
,{w,a1,5*a1}
,AxesLabel -> {"w", "|H9(jw)|"}
];
```



```
Plot[Abs[H9w]
,{w,0,5*a1}
,AxesLabel -> {"w", "|H9(jw)|"}
];
```

$|H_9(j\omega)|$ 